## Chapter 7: Plane-Wave Propagation

## Lesson \#43

Chapter - Section: 7-1
Topics: Time-harmonic fields

## Highlights:

- Phasors
- Complex permittivity
- Wave equations


## Special Illustrations:

## Lesson \#44

Chapter - Section: 7-2
Topics: Waves in lossless media

## Highlights:

- Uniform plane waves
- Intrinsic impedance
- Wave properties


## Special Illustrations:

- Example 7-1
- CD-ROM Modules 7.3 and 7.4


## Module 7.2: Wave Properties

Given: The electric field of a plane electromagnetic wave traveling in air exhibits the pattern shown in the figure. In order to be able to visually observe the time variation, the rate has been slowed down by a factor $M$.


Start Animati 0:03
Q1. What is the wavelength of the wave?

$$
\lambda=\square \mathrm{cm} \quad \text { check answer } \quad \text { I give up }
$$

Q2. What is the apparent frquency of the wave?

$$
f=\square \mathrm{Hz} \quad \text { check answer } \quad 1 \text { give up }
$$

Q3. What is the slow-down factor, $\mathrm{M}=$ (True frequency / Apparent frequency)?

$$
M=\square \times 10^{9} \quad \text { Check answer } \quad \text { I give up }
$$

## Lesson \#45 and 46

## Chapter - Section: 7-3

Topics: Wave polarization

## Highlights:

- Definition of polarization
- Linear, circular, elliptical


## Special Illustrations:

- CD-ROM Demos 7.1-7.5
- Liquid Crystal Display


## Liquid Crystal Display (LCD)

LCDs are used in digital clocks, cellular phones, desktop and laptop computers, and some televisions and other electronic systems. They offer a decided advantage over other display technologies, such as cathode ray tubes, in that they are much lighter and thinner and consume a lot less power to operate. LCD technology relies on special electrical and optical properties of a class of materials known as liquid crystals, first discovered in the 1880s by botanist Friedrich Reinitzer.

## Physical Principle

Liquid crystals are neither a pure solid nor a pure liquid, but rather a hybrid of both. One particular variety of interest is the twisted nematic liquid crystal whose molecules have a natural tendency to assume a twisted spiral structure when the material is sandwiched between finely grooved glass substrates with orthogonal orientations (A). Note that the molecules in contact with the grooved surfaces align themselves in parallel along
 the grooves. The molecular spiral causes the crystal to behave like a wave polarizer; unpolarized light incident upon the entrance substrate follows the orientation of the spiral, emerging through the exit substrate with its polarization (direction of electric field) parallel to the groove's direction.

## Lesson \#47

Chapter - Section: 7-4
Topics: Waves in lossy media

## Highlights:

- Attenuation and skin depth
- Low loss medium
- Good conductor


## Special Illustrations:

- CD-ROM Demos 7.6-7.8


## Demo 7.7: Moderately Lossy

Given: A $10-\mathrm{MHz}$ EM plane wave propagating in a moderately lossy medium characterized by:

$$
\varepsilon_{\mathrm{r}}=9 \text { and } \sigma=10^{-2} \mathrm{~S} / \mathrm{m} .
$$

Assuming that $\mathbf{E}$ has a magnitude of $10 \mathrm{~V} / \mathrm{m}$ at $z=0$, solve for and display the following profiles:
(a) $E(z, t)$.
(b) $\left|\eta_{C}\right| H(z, t)$.
(c) The power density $S(z, t)$.


## Lesson \#48

Chapter - Section: 7-5
Topics: Current flow in conductors

## Highlights:

- Skin depth dependence on frequency
- Surface impedance


## Special Illustrations:

## Lesson \#49

Chapter - Section: 7-6
Topics: EM power density

## Highlights:

- Power density in a lossless medium
- Power density in a lossy medium
- Time-average power


## Special Illustrations:

- CD-ROM Module 7.5


## Module 7.5: UHF Antenna Reception

Given: A television receiver with a loop antenna.

Q. If the electric field of the wave radiated by the TV station antenna is along the vertical ( $z$-axis), what plane should the loop antenna be placed into in order to maximize the received signal? In (a) the loop is in the $\mathbf{E}-\hat{k}$ plane, and in (b) it is in the $\mathbf{E}-\mathbf{H}$ plane.

## Chapter 7

## Section 7-2: Propagation in Lossless Media

Problem 7.1 The magnetic field of a wave propagating through a certain nonmagnetic material is given by

$$
\mathbf{H}=\hat{\mathbf{z}} 30 \cos \left(10^{8} t-0.5 y\right) \quad(\mathrm{mA} / \mathrm{m}) .
$$

Find (a) the direction of wave propagation, (b) the phase velocity, (c) the wavelength in the material, (d) the relative permittivity of the material, and (e) the electric field phasor.

## Solution:

(a) Positive $y$-direction.
(b) $\omega=10^{8} \mathrm{rad} / \mathrm{s}, k=0.5 \mathrm{rad} / \mathrm{m}$.

$$
u_{\mathrm{p}}=\frac{\omega}{k}=\frac{10^{8}}{0.5}=2 \times 10^{8} \mathrm{~m} / \mathrm{s} .
$$

(c) $\lambda=2 \pi / k=2 \pi / 0.5=12.6 \mathrm{~m}$.
(d) $\varepsilon_{\mathrm{r}}=\left(\frac{c}{u_{\mathrm{p}}}\right)^{2}=\left(\frac{3 \times 10^{8}}{2 \times 10^{8}}\right)^{2}=2.25$.
(e) From Eq. (7.39b),

$$
\begin{aligned}
& \widetilde{\mathbf{E}}=-\eta \hat{\mathbf{k}} \times \widetilde{\mathbf{H}}, \\
& \eta=\sqrt{\frac{\mu}{\varepsilon}}=\frac{120 \pi}{\sqrt{\varepsilon_{\mathrm{r}}}}=\frac{120 \pi}{1.5}=251.33 \quad(\Omega), \\
& \hat{\mathbf{k}}=\hat{\mathbf{y}}, \quad \text { and } \quad \widetilde{\mathbf{H}}=\hat{\mathbf{z}} 30 e^{-j 0.5 y} \times 10^{-3} \quad(\mathrm{~A} / \mathrm{m}) .
\end{aligned}
$$

Hence,

$$
\widetilde{\mathbf{E}}=-251.33 \hat{\mathbf{y}} \times \hat{\mathbf{z}} 30 e^{-j 0.5 y} \times 10^{-3}=-\hat{\mathbf{x}} 7.54 e^{-j 0.5 y} \quad(\mathrm{~V} / \mathrm{m}),
$$

and

$$
\mathbf{E}(y, t)=\mathfrak{R e}\left(\widetilde{\mathbf{E}} e^{j \omega t}\right)=-\hat{\mathbf{x}} 7.54 \cos \left(10^{8} t-0.5 y\right) \quad(\mathrm{V} / \mathrm{m}) .
$$

Problem 7.2 Write general expressions for the electric and magnetic fields of a $1-\mathrm{GHz}$ sinusoidal plane wave traveling in the $+y$-direction in a lossless nonmagnetic medium with relative permittivity $\varepsilon_{\mathrm{r}}=9$. The electric field is polarized along the $x$-direction, its peak value is $6 \mathrm{~V} / \mathrm{m}$ and its intensity is $4 \mathrm{~V} / \mathrm{m}$ at $t=0$ and $y=2 \mathrm{~cm}$.

Solution: For $f=1 \mathrm{GHz}, \mu_{\mathrm{r}}=1$, and $\varepsilon_{\mathrm{r}}=9$,

$$
\begin{aligned}
\omega & =2 \pi f=2 \pi \times 10^{9} \mathrm{rad} / \mathrm{s} \\
k & =\frac{2 \pi}{\lambda}=\frac{2 \pi}{\lambda_{0}} \sqrt{\varepsilon_{\mathrm{r}}}=\frac{2 \pi f}{c} \sqrt{\varepsilon_{\mathrm{r}}}=\frac{2 \pi \times 10^{9}}{3 \times 10^{8}} \sqrt{9}=20 \pi \mathrm{rad} / \mathrm{m} \\
\mathbf{E}(y, t) & =\hat{\mathbf{x}} 6 \cos \left(2 \pi \times 10^{9} t-20 \pi y+\phi_{0}\right) \quad(\mathrm{V} / \mathrm{m}) .
\end{aligned}
$$

At $t=0$ and $y=2 \mathrm{~cm}, E=4 \mathrm{~V} / \mathrm{m}:$

$$
4=6 \cos \left(-20 \pi \times 2 \times 10^{-2}+\phi_{0}\right)=6 \cos \left(-0.4 \pi+\phi_{0}\right)
$$

Hence,

$$
\phi_{0}-0.4 \pi=\cos ^{-1}\left(\frac{4}{6}\right)=0.84 \mathrm{rad}
$$

which gives

$$
\phi_{0}=2.1 \mathrm{rad}=120.19^{\circ}
$$

and

$$
\mathbf{E}(y, t)=\hat{\mathbf{x}} 6 \cos \left(2 \pi \times 10^{9} t-20 \pi y+120.19^{\circ}\right) \quad(\mathrm{V} / \mathrm{m})
$$

Problem 7.3 The electric field phasor of a uniform plane wave is given by $\widetilde{\mathbf{E}}=\hat{\mathbf{y}} 10 e^{j 0.2 z}(\mathrm{~V} / \mathrm{m})$. If the phase velocity of the wave is $1.5 \times 10^{8} \mathrm{~m} / \mathrm{s}$ and the relative permeability of the medium is $\mu_{\mathrm{r}}=2.4$, find (a) the wavelength, (b) the frequency $f$ of the wave, (c) the relative permittivity of the medium, and (d) the magnetic field $\mathbf{H}(z, t)$.

## Solution:

(a) From $\widetilde{\mathbf{E}}=\hat{\mathbf{y}} 10 e^{j 0.2 z}(\mathrm{~V} / \mathrm{m})$, we deduce that $k=0.2 \mathrm{rad} / \mathrm{m}$. Hence,

$$
\lambda=\frac{2 \pi}{k}=\frac{2 \pi}{0.2}=10 \pi=31.42 \mathrm{~m} .
$$

(b)

$$
f=\frac{u_{\mathrm{p}}}{\lambda}=\frac{1.5 \times 10^{8}}{31.42}=4.77 \times 10^{6} \mathrm{~Hz}=4.77 \mathrm{MHz} .
$$

(c) From

$$
u_{\mathrm{p}}=\frac{c}{\sqrt{\mu_{\mathrm{r}} \varepsilon_{\mathrm{r}}}}, \quad \varepsilon_{\mathrm{r}}=\frac{1}{\mu_{\mathrm{r}}}\left(\frac{c}{u_{\mathrm{p}}}\right)^{2}=\frac{1}{2.4}\left(\frac{3}{1.5}\right)^{2}=1.67 .
$$

(d)

$$
\begin{aligned}
\eta & =\sqrt{\frac{\mu}{\varepsilon}} \simeq 120 \pi \sqrt{\frac{\mu_{\mathrm{r}}}{\varepsilon_{\mathrm{r}}}}=120 \pi \sqrt{\frac{2.4}{1.67}}=451.94 \quad(\Omega), \\
\widetilde{\mathbf{H}} & =\frac{1}{\eta}(-\hat{\mathbf{z}}) \times \widetilde{\mathbf{E}}=\frac{1}{\eta}(-\hat{\mathbf{z}}) \times \hat{\mathbf{y}} 10 e^{j 0.2 z}=\hat{\mathbf{x}} 22.13 e^{j 0.2 z \quad(\mathrm{~mA} / \mathrm{m}),} \\
\mathbf{H}(z, t) & =\hat{\mathbf{x}} 22.13 \cos (\omega t+0.2 z) \quad(\mathrm{mA} / \mathrm{m}),
\end{aligned}
$$

with $\omega=2 \pi f=9.54 \pi \times 10^{6} \mathrm{rad} / \mathrm{s}$.

Problem 7.4 The electric field of a plane wave propagating in a nonmagnetic material is given by

$$
\mathbf{E}=\left[\hat{\mathbf{y}} 3 \sin \left(\pi \times 10^{7} t-0.2 \pi x\right)+\hat{\mathbf{z}} 4 \cos \left(\pi \times 10^{7} t-0.2 \pi x\right)\right] \quad(\mathrm{V} / \mathrm{m}) .
$$

Determine (a) the wavelength, (b) $\varepsilon_{\mathrm{r}}$, and (c) $\mathbf{H}$.

## Solution:

(a) Since $k=0.2 \pi$,

$$
\lambda=\frac{2 \pi}{k}=\frac{2 \pi}{0.2 \pi}=10 \mathrm{~m} .
$$

(b)

$$
u_{\mathrm{p}}=\frac{\omega}{k}=\frac{\pi \times 10^{7}}{0.2 \pi}=5 \times 10^{7} \mathrm{~m} / \mathrm{s} .
$$

But

$$
u_{\mathrm{p}}=\frac{c}{\sqrt{\varepsilon_{\mathrm{r}}}} .
$$

Hence,

$$
\varepsilon_{\mathrm{r}}=\left(\frac{c}{u_{\mathrm{p}}}\right)^{2}=\left(\frac{3 \times 10^{8}}{5 \times 10^{7}}\right)^{2}=36 .
$$

(c)

$$
\begin{aligned}
\mathbf{H}=\frac{1}{\eta} \hat{\mathbf{k}} \times \mathbf{E} & =\frac{1}{\eta} \hat{\mathbf{x}} \times\left[\hat{\mathbf{y}} 3 \sin \left(\pi \times 10^{7} t-0.2 \pi x\right)+\hat{\mathbf{z}} 4 \cos \left(\pi \times 10^{7} t-0.2 \pi x\right)\right] \\
& =\hat{\mathbf{z}} \frac{3}{\eta} \sin \left(\pi \times 10^{7} t-0.2 \pi x\right)-\hat{\mathbf{y}} \frac{4}{\eta} \cos \left(\pi \times 10^{7} t-0.2 \pi x\right) \quad(\mathrm{A} / \mathrm{m}),
\end{aligned}
$$

with

$$
\eta=\frac{\eta_{0}}{\sqrt{\varepsilon_{\mathrm{r}}}} \simeq \frac{120 \pi}{6}=20 \pi=62.83
$$

Problem 7.5 A wave radiated by a source in air is incident upon a soil surface, whereupon a part of the wave is transmitted into the soil medium. If the wavelength of the wave is 60 cm in air and 20 cm in the soil medium, what is the soil's relative permittivity? Assume the soil to be a very low loss medium.
Solution: From $\lambda=\lambda_{0} / \sqrt{\varepsilon_{\mathrm{r}}}$,

$$
\varepsilon_{\mathrm{r}}=\left(\frac{\lambda_{0}}{\lambda}\right)^{2}=\left(\frac{60}{20}\right)^{2}=9
$$

Problem 7.6 The electric field of a plane wave propagating in a lossless, nonmagnetic, dielectric material with $\varepsilon_{\mathrm{r}}=2.56$ is given by

$$
\mathbf{E}=\hat{\mathbf{y}} 20 \cos \left(6 \pi \times 10^{9} t-k z\right) \quad(\mathrm{V} / \mathrm{m})
$$

Determine:
(a) $f, u_{\mathrm{p}}, \lambda, k$, and $\eta$, and
(b) the magnetic field $\mathbf{H}$.

## Solution:

(a)

$$
\begin{aligned}
\omega & =2 \pi f=6 \pi \times 10^{9} \mathrm{rad} / \mathrm{s}, \\
f & =3 \times 10^{9} \mathrm{~Hz}=3 \mathrm{GHz} \\
u_{\mathrm{p}} & =\frac{c}{\sqrt{\varepsilon_{\mathrm{r}}}}=\frac{3 \times 10^{8}}{\sqrt{2.56}}=1.875 \times 10^{8} \mathrm{~m} / \mathrm{s}, \\
\lambda & =\frac{u_{\mathrm{p}}}{f}=\frac{1.875 \times 10^{8}}{6 \times 10^{9}}=3.12 \mathrm{~cm} \\
k & =\frac{2 \pi}{\lambda}=\frac{2 \pi}{3.12 \times 10^{-2}}=201.4 \mathrm{rad} / \mathrm{m} \\
\eta & =\frac{\eta_{0}}{\sqrt{\varepsilon_{\mathrm{r}}}}=\frac{377}{\sqrt{2.56}}=\frac{377}{1.6}=235.62 \Omega .
\end{aligned}
$$

(b)

$$
\begin{aligned}
\mathbf{H} & =-\hat{\mathbf{x}} \frac{20}{\eta} \cos \left(6 \pi \times 10^{9} t-k z\right) \\
& =-\hat{\mathbf{x}} \frac{20}{235.62} \cos \left(6 \pi \times 10^{9} t-201.4 z\right) \\
& =-\hat{\mathbf{x}} 8.49 \times 10^{-2} \cos \left(6 \pi \times 10^{9} t-201.4 z\right) \quad(\mathrm{A} / \mathrm{m}) .
\end{aligned}
$$

## Section 7-3: Wave Polarization

Problem 7.7 An RHC-polarized wave with a modulus of $2(\mathrm{~V} / \mathrm{m})$ is traveling in free space in the negative $z$-direction. Write down the expression for the wave's electric field vector, given that the wavelength is 6 cm .


Figure P7.7: Locus of $\mathbf{E}$ versus time.

Solution: For an RHC wave traveling in $-\hat{\mathbf{z}}$, let us try the following:

$$
\mathbf{E}=\hat{\mathbf{x}} a \cos (\omega t+k z)+\hat{\mathbf{y}} a \sin (\omega t+k z) .
$$

Modulus $|E|=\sqrt{a^{2}+a^{2}}=a \sqrt{2}=2(\mathrm{~V} / \mathrm{m})$. Hence,

$$
a=\frac{2}{\sqrt{2}}=\sqrt{2} .
$$

Next, we need to check the sign of the $\hat{\mathbf{y}}$-component relative to that of the $\hat{\mathbf{x}}$-component. We do this by examining the locus of $\mathbf{E}$ versus $t$ at $z=0$ : Since the wave is traveling along $-\hat{\mathbf{z}}$, when the thumb of the right hand is along $-\hat{\mathbf{z}}$ (into the page), the other four fingers point in the direction shown (clockwise as seen from above). Hence, we should reverse the sign of the $\hat{\mathbf{y}}$-component:

$$
\mathbf{E}=\hat{\mathbf{x}} \sqrt{2} \cos (\omega t+k z)-\hat{\mathbf{y}} \sqrt{2} \sin (\omega t+k z) \quad(\mathrm{V} / \mathrm{m})
$$

with

$$
k=\frac{2 \pi}{\lambda}=\frac{2 \pi}{6 \times 10^{-2}}=104.72 \quad(\mathrm{rad} / \mathrm{m})
$$

and

$$
\omega=k c=\frac{2 \pi}{\lambda} \times 3 \times 10^{8}=\pi \times 10^{10} \quad(\mathrm{rad} / \mathrm{s})
$$

Problem 7.8 For a wave characterized by the electric field

$$
\mathbf{E}(z, t)=\hat{\mathbf{x}} a_{x} \cos (\omega t-k z)+\hat{\mathbf{y}} a_{y} \cos (\omega t-k z+\delta),
$$

identify the polarization state, determine the polarization angles $(\gamma, \chi)$, and sketch the locus of $\mathbf{E}(0, t)$ for each of the following cases:
(a) $a_{x}=3 \mathrm{~V} / \mathrm{m}, a_{y}=4 \mathrm{~V} / \mathrm{m}$, and $\delta=0$,
(b) $a_{x}=3 \mathrm{~V} / \mathrm{m}, a_{y}=4 \mathrm{~V} / \mathrm{m}$, and $\delta=180^{\circ}$,
(c) $a_{x}=3 \mathrm{~V} / \mathrm{m}, a_{y}=3 \mathrm{~V} / \mathrm{m}$, and $\delta=45^{\circ}$,
(d) $a_{x}=3 \mathrm{~V} / \mathrm{m}, a_{y}=4 \mathrm{~V} / \mathrm{m}$, and $\delta=-135^{\circ}$.

## Solution:

$$
\begin{aligned}
\psi_{0} & =\tan ^{-1}\left(a_{y} / a_{x}\right), \quad[\text { Eq. }(7.60)], \\
\tan 2 \gamma & =\left(\tan 2 \psi_{0}\right) \cos \delta \quad[\text { Eq. }(7.59 \mathrm{a})], \\
\sin 2 \chi & =\left(\sin 2 \psi_{0}\right) \sin \delta \quad[\text { Eq. }(7.59 \mathrm{~b})] .
\end{aligned}
$$

| Case | $a_{x}$ | $a_{y}$ | $\delta$ | $\psi_{0}$ | $\gamma$ | $\chi$ | Polarization State |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| (a) | 3 | 4 | 0 | $53.13^{\circ}$ | $53.13^{\circ}$ | 0 | Linear |
| (b) | 3 | 4 | $180^{\circ}$ | $53.13^{\circ}$ | $-53.13^{\circ}$ | 0 | Linear |
| (c) | 3 | 3 | $45^{\circ}$ | $45^{\circ}$ | $45^{\circ}$ | $22.5^{\circ}$ | Left elliptical |
| (d) | 3 | 4 | $-135^{\circ}$ | $53.13^{\circ}$ | $-56.2^{\circ}$ | $-21.37^{\circ}$ | Right elliptical |

(a) $\mathbf{E}(z, t)=\hat{\mathbf{x}} 3 \cos (\omega t-k z)+\hat{\mathbf{y}} 4 \cos (\omega t-k z)$.
(b) $\mathbf{E}(z, t)=\hat{\mathbf{x}} 3 \cos (\omega t-k z)-\hat{\mathbf{y}} 4 \cos (\omega t-k z)$.
(c) $\mathbf{E}(z, t)=\hat{\mathbf{x}} 3 \cos (\omega t-k z)+\hat{\mathbf{y}} 3 \cos \left(\omega t-k z+45^{\circ}\right)$.
(d) $\mathbf{E}(z, t)=\hat{\mathbf{x}} 3 \cos (\omega t-k z)+\hat{\mathbf{y}} 4 \cos \left(\omega t-k z-135^{\circ}\right)$.


Figure P7.8: Plots of the locus of $\mathbf{E}(0, t)$.

Problem 7.9 The electric field of a uniform plane wave propagating in free space is given by $\widetilde{\mathbf{E}}=(\hat{\mathbf{x}}+j \hat{\mathbf{y}}) 20 e^{-j \pi z / 6}(\mathrm{~V} / \mathrm{m})$. Specify the modulus and direction of the electric field intensity at the $z=0$ plane at $t=0,5$ and 10 ns .

## Solution:

$$
\begin{aligned}
\mathbf{E}(z, t) & =\mathfrak{R e}\left[\widetilde{\mathbf{E}} e^{j \omega t}\right] \\
& =\mathfrak{R e}\left[(\hat{\mathbf{x}}+j \hat{\mathbf{y}}) 20 e^{-j \pi z / 6} e^{j \omega t}\right] \\
& =\mathfrak{R e}\left[\left(\hat{\mathbf{x}}+\hat{\mathbf{y}} e^{j \pi / 2}\right) 20 e^{-j \pi z / 6} e^{j \omega t}\right] \\
& =\hat{\mathbf{x}} 20 \cos (\omega t-\pi z / 6)+\hat{\mathbf{y}} 20 \cos (\omega t-\pi z / 6+\pi / 2) \\
& =\hat{\mathbf{x}} 20 \cos (\omega t-\pi z / 6)-\hat{\mathbf{y}} 20 \sin (\omega t-\pi z / 6) \quad(\mathrm{V} / \mathrm{m}), \\
|\mathbf{E}| & =\left[E_{x}^{2}+E_{y}^{2}\right]^{1 / 2}=20 \quad(\mathrm{~V} / \mathrm{m}), \\
\psi & =\tan ^{-1}\left(\frac{E_{y}}{E_{x}}\right)=-(\omega t-\pi z / 6) .
\end{aligned}
$$

From

$$
\begin{aligned}
& f=\frac{c}{\lambda}=\frac{k c}{2 \pi}=\frac{\pi / 6 \times 3 \times 10^{8}}{2 \pi}=2.5 \times 10^{7} \mathrm{~Hz}, \\
& \omega=2 \pi f=5 \pi \times 10^{7} \mathrm{rad} / \mathrm{s} .
\end{aligned}
$$

At $z=0$,

$$
\psi=-\omega t=-5 \pi \times 10^{7} t= \begin{cases}0 & \text { at } t=0, \\ -0.25 \pi=-45^{\circ} & \text { at } t=5 \mathrm{~ns}, \\ -0.5 \pi=-90^{\circ} & \text { at } t=10 \mathrm{~ns} .\end{cases}
$$

Therefore, the wave is LHC polarized.
Problem 7.10 A linearly polarized plane wave of the form $\widetilde{\mathbf{E}}=\hat{\mathbf{x}} a_{x} e^{-j k z}$ can be expressed as the sum of an RHC polarized wave with magnitude $a_{\mathrm{R}}$ and an LHC polarized wave with magnitude $a_{\mathrm{L}}$. Prove this statement by finding expressions for $a_{\mathrm{R}}$ and $a_{\mathrm{L}}$ in terms of $a_{x}$.

## Solution:

$$
\begin{aligned}
\widetilde{\mathbf{E}} & =\hat{\mathbf{x}} a_{x} e^{-j k z}, \\
\text { RHC wave: } \quad \widetilde{\mathbf{E}}_{\mathrm{R}} & =a_{\mathrm{R}}\left(\hat{\mathbf{x}}+\hat{\mathbf{y}} e^{-j \pi / 2}\right) e^{-j k z}=a_{\mathrm{R}}(\hat{\mathbf{x}}-j \hat{\mathbf{y}}) e^{-j k z}, \\
\text { LHC wave: } \quad \widetilde{\mathbf{E}}_{\mathrm{L}} & =a_{\mathrm{L}}\left(\hat{\mathbf{x}}+\hat{\mathbf{y}} e^{j \pi / 2}\right) e^{-j k z}=a_{\mathrm{L}}(\hat{\mathbf{x}}+j \hat{\mathbf{y}}) e^{-j k z}, \\
\widetilde{\mathbf{E}} & =\widetilde{\mathbf{E}}_{\mathrm{R}}+\widetilde{\mathbf{E}}_{\mathrm{L}}, \\
\hat{\mathbf{x}} a_{x} & =a_{\mathrm{R}}(\hat{\mathbf{x}}-j \hat{\mathbf{y}})+a_{\mathrm{L}}(\hat{\mathbf{x}}+j \hat{\mathbf{y}}) .
\end{aligned}
$$

By equating real and imaginary parts, $a_{x}=a_{\mathrm{R}}+a_{\mathrm{L}}, 0=-a_{\mathrm{R}}+a_{\mathrm{L}}$, or $a_{\mathrm{L}}=a_{x} / 2$, $a_{\mathrm{R}}=a_{x} / 2$.

Problem 7.11 The electric field of an elliptically polarized plane wave is given by

$$
\mathbf{E}(z, t)=\left[-\hat{\mathbf{x}} 10 \sin \left(\omega t-k z-60^{\circ}\right)+\hat{\mathbf{y}} 30 \cos (\omega t-k z)\right] \quad(\mathrm{V} / \mathrm{m}) .
$$

Determine (a) the polarization angles $(\gamma, \chi)$ and (b) the direction of rotation.

## Solution:

(a)

$$
\begin{aligned}
\mathbf{E}(z, t) & =\left[-\hat{\mathbf{x}} 10 \sin \left(\omega t-k z-60^{\circ}\right)+\hat{\mathbf{y}} 30 \cos (\omega t-k z)\right] \\
& =\left[\hat{\mathbf{x}} 10 \cos \left(\omega t-k z+30^{\circ}\right)+\hat{\mathbf{y}} 30 \cos (\omega t-k z)\right](\mathrm{V} / \mathrm{m}) .
\end{aligned}
$$

Phasor form:

$$
\widetilde{\mathbf{E}}=\left(\hat{\mathbf{x}} 10 e^{j 30^{\circ}}+\hat{\mathbf{y}} 30\right) e^{-j k z}
$$

Since $\delta$ is defined as the phase of $E_{y}$ relative to that of $E_{x}$,

$$
\begin{aligned}
\delta & =-30^{\circ}, \\
\psi_{0} & =\tan ^{-1}\left(\frac{30}{10}\right)=71.56^{\circ}, \\
\tan 2 \gamma & =\left(\tan 2 \psi_{0}\right) \cos \delta=-0.65 \quad \text { or } \gamma=73.5^{\circ}, \\
\sin 2 \chi & =\left(\sin 2 \psi_{0}\right) \sin \delta=-0.40 \quad \text { or } \chi=-8.73^{\circ} .
\end{aligned}
$$

(b) Since $\chi<0$, the wave is right-hand elliptically polarized.

Problem 7.12 Compare the polarization states of each of the following pairs of plane waves:
(a) wave 1: $\mathbf{E}_{1}=\hat{\mathbf{x}} 2 \cos (\omega t-k z)+\hat{\mathbf{y}} 2 \sin (\omega t-k z)$,
wave 2: $\mathbf{E}_{2}=\hat{\mathbf{x}} 2 \cos (\omega t+k z)+\hat{\mathbf{y}} 2 \sin (\omega t+k z)$,
(b) wave 1: $\mathbf{E}_{1}=\hat{\mathbf{x}} 2 \cos (\omega t-k z)-\hat{\mathbf{y}} 2 \sin (\omega t-k z)$, wave 2: $\mathbf{E}_{2}=\hat{\mathbf{x}} 2 \cos (\omega t+k z)-\hat{\mathbf{y}} 2 \sin (\omega t+k z)$.

## Solution:

(a)

$$
\begin{aligned}
\mathbf{E}_{1} & =\hat{\mathbf{x}} 2 \cos (\omega t-k z)+\hat{\mathbf{y}} 2 \sin (\omega t-k z) \\
& =\hat{\mathbf{x}} 2 \cos (\omega t-k z)+\hat{\mathbf{y}} 2 \cos (\omega t-k z-\pi / 2), \\
\widetilde{\mathbf{E}}_{1} & =\hat{\mathbf{x}} 2 e^{-j k z}+\hat{\mathbf{y}} 2 e^{-j k z} e^{-j \pi / 2},
\end{aligned}
$$

$$
\begin{aligned}
\psi_{0} & =\tan ^{-1}\left(\frac{a y}{a x}\right)=\tan ^{-1} 1=45^{\circ}, \\
\delta & =-\pi / 2 .
\end{aligned}
$$

Hence, wave 1 is RHC.
Similarly,

$$
\widetilde{\mathbf{E}}_{2}=\hat{\mathbf{x}} 2 e^{j k z}+\hat{\mathbf{y}} 2 e^{j k z} e^{-j \pi / 2}
$$

Wave 2 has the same magnitude and phases as wave 1 except that its direction is along $-\hat{\mathbf{z}}$ instead of $+\hat{\mathbf{z}}$. Hence, the locus of rotation of $\mathbf{E}$ will match the left hand instead of the right hand. Thus, wave 2 is LHC.
(b)

$$
\begin{aligned}
& \mathbf{E}_{1}=\hat{\mathbf{x}} 2 \cos (\omega t-k z)-\hat{\mathbf{y}} 2 \sin (\omega t-k z), \\
& \widetilde{\mathbf{E}}_{1}=\hat{\mathbf{x}} 2 e^{-j k z}+\hat{\mathbf{y}} 2 e^{-j k z} e^{j \pi / 2}
\end{aligned}
$$

Wave 1 is LHC.

$$
\widetilde{\mathbf{E}}_{2}=\hat{\mathbf{x}} 2 e^{j k z}+\hat{\mathbf{y}} 2 e^{j k z} e^{j \pi / 2}
$$

Reversal of direction of propagation (relative to wave 1) makes wave 2 RHC.

Problem 7.13 Plot the locus of $\mathbf{E}(0, t)$ for a plane wave with

$$
\mathbf{E}(z, t)=\hat{\mathbf{x}} \sin (\omega t+k z)+\hat{\mathbf{y}} 2 \cos (\omega t+k z)
$$

Determine the polarization state from your plot.

## Solution:

$$
\mathbf{E}=\hat{\mathbf{x}} \sin (\omega t+k z)+\hat{\mathbf{y}} 2 \cos (\omega t+k z) .
$$

Wave direction is $-\hat{\mathbf{z}}$. At $z=0$,

$$
\mathbf{E}=\hat{\mathbf{x}} \sin \omega t+\hat{\mathbf{y}} 2 \cos \omega t .
$$

Tip of $\mathbf{E}$ rotates in accordance with right hand (with thumb pointing along $-\hat{\mathbf{z}}$ ). Hence, wave state is RHE.


Figure P7.13: Locus of $\mathbf{E}$ versus time.

## Sections 7-4: Propagation in a Lossy Medium

Problem 7.14 For each of the following combination of parameters, determine if the material is a low-loss dielectric, a quasi-conductor, or a good conductor, and then calculate $\alpha, \beta, \lambda$, $u_{\mathrm{p}}$, and $\eta_{\mathrm{c}}$ :
(a) glass with $\mu_{\mathrm{r}}=1, \varepsilon_{\mathrm{r}}=5$, and $\sigma=10^{-12} \mathrm{~S} / \mathrm{m}$ at 10 GHz ,
(b) animal tissue with $\mu_{\mathrm{r}}=1, \varepsilon_{\mathrm{r}}=12$, and $\sigma=0.3 \mathrm{~S} / \mathrm{m}$ at 100 MHz ,
(c) wood with $\mu_{\mathrm{r}}=1, \varepsilon_{\mathrm{r}}=3$, and $\sigma=10^{-4} \mathrm{~S} / \mathrm{m}$ at 1 kHz .

Solution: Using equations given in Table 7-1:

|  | Case (a) | Case (b) | Case (c) |
| :---: | :---: | :---: | :---: |
| $\sigma / \omega \varepsilon$ | $3.6 \times 10^{-13}$ | 4.5 | 600 |
| Type | low-loss dielectric | quasi-conductor | good conductor |
| $\alpha$ | $8.42 \times 10^{-11} \mathrm{~Np} / \mathrm{m}$ | $9.75 \mathrm{~Np} / \mathrm{m}$ | $6.3 \times 10^{-4} \mathrm{~Np} / \mathrm{m}$ |
| $\beta$ | $468.3 \mathrm{rad} / \mathrm{m}$ | $12.16 \mathrm{rad} / \mathrm{m}$ | $6.3 \times 10^{-4} \mathrm{rad} / \mathrm{m}$ |
| $\lambda$ | 1.34 cm | 51.69 cm | 10 km |
| $u_{\mathrm{p}}$ | $1.34 \times 10^{8} \mathrm{~m} / \mathrm{s}$ | $0.52 \times 10^{8} \mathrm{~m} / \mathrm{s}$ | $0.1 \times 10^{8} \mathrm{~m} / \mathrm{s}$ |
| $\eta_{\mathrm{c}}$ | $\simeq 168.5 \Omega$ | $39.54+j 31.72 \Omega$ | $6.28(1+j) \Omega$ |

Problem 7.15 Dry soil is characterized by $\varepsilon_{\mathrm{r}}=2.5, \mu_{\mathrm{r}}=1$, and $\sigma=10^{-4}(\mathrm{~S} / \mathrm{m})$. At each of the following frequencies, determine if dry soil may be considered a good conductor, a quasi-conductor, or a low-loss dielectric, and then calculate $\alpha, \beta, \lambda, \mu_{\mathrm{p}}$, and $\eta_{\mathrm{c}}$ :
(a) 60 Hz ,
(b) 1 kHz ,
(c) 1 MHz ,
(d) 1 GHz .

Solution: $\varepsilon_{\mathrm{r}}=2.5, \mu_{\mathrm{r}}=1, \sigma=10^{-4} \mathrm{~S} / \mathrm{m}$.

| $f \rightarrow$ | 60 Hz | 1 kHz | 1 MHz | 1 GHz |
| :---: | :---: | :---: | :---: | :---: |
| $\frac{\varepsilon^{\prime \prime}}{\varepsilon^{\prime}}=\frac{\sigma}{\omega \varepsilon}$ <br> $=\frac{\sigma}{2 \pi f \varepsilon_{\mathrm{r}} \varepsilon_{0}}$ | $1.2 \times 10^{4}$ | 720 | 0.72 | $7.2 \times 10^{-4}$ |
| Type of medium | Good conductor | Good conductor | Quasi-conductor | Low-loss dielectric |
| $\alpha(\mathrm{Np} / \mathrm{m})$ | $1.54 \times 10^{-4}$ | $6.28 \times 10^{-4}$ | $1.13 \times 10^{-2}$ | $1.19 \times 10^{-2}$ |
| $\beta(\mathrm{rad} / \mathrm{m})$ | $1.54 \times 10^{-4}$ | $6.28 \times 10^{-4}$ | $3.49 \times 10^{-2}$ | 33.14 |
| $\lambda(\mathrm{~m})$ | $4.08 \times 10^{4}$ | $10^{4}$ | 180 | 0.19 |
| $u_{\mathrm{p}}(\mathrm{m} / \mathrm{s})$ | $2.45 \times 10^{6}$ | $10^{7}$ | $1.8 \times 10^{8}$ | $1.9 \times 10^{8}$ |
| $\eta_{\mathrm{c}}(\Omega)$ | $1.54(1+j)$ | $6.28(1+j)$ | $204.28+j 65.89$ | 238.27 |

Problem 7.16 In a medium characterized by $\varepsilon_{\mathrm{r}}=9, \mu_{\mathrm{r}}=1$, and $\sigma=0.1 \mathrm{~S} / \mathrm{m}$, determine the phase angle by which the magnetic field leads the electric field at 100 MHz .

Solution: The phase angle by which the magnetic field leads the electric field is $-\theta_{\eta}$ where $\theta_{\eta}$ is the phase angle of $\eta_{c}$.

$$
\frac{\sigma}{\omega \varepsilon}=\frac{0.1 \times 36 \pi}{2 \pi \times 10^{8} \times 10^{-9} \times 9}=2 .
$$

Hence, quasi-conductor.

$$
\begin{aligned}
\eta_{\mathrm{c}}=\sqrt{\frac{\mu}{\varepsilon^{\prime}}}\left(1-j \frac{\varepsilon^{\prime \prime}}{\varepsilon^{\prime}}\right)^{-1 / 2} & =\frac{120 \pi}{\sqrt{\varepsilon_{\mathrm{r}}}}\left(1-j \frac{\sigma}{\omega \varepsilon_{0} \varepsilon_{\mathrm{r}}}\right)^{-1 / 2} \\
& =125.67(1-j 2)^{-1 / 2}=71.49+j 44.18=84.04 \angle 311.72^{\circ}
\end{aligned}
$$

Therefore $\theta_{\eta}=31.72^{\circ}$.
Since $\mathbf{H}=\left(1 / \eta_{\mathrm{c}}\right) \hat{\mathbf{k}} \times \mathbf{E}, \mathbf{H}$ leads $\mathbf{E}$ by $-\theta_{\eta}$, or by $-31.72^{\circ}$. In other words, $\mathbf{H}$ lags E by $31.72^{\circ}$.

Problem 7.17 Generate a plot for the skin depth $\delta_{\mathrm{s}}$ versus frequency for seawater for the range from 1 kHz to 10 GHz (use log-log scales). The constitutive parameters of seawater are $\mu_{\mathrm{r}}=1, \varepsilon_{\mathrm{r}}=80$ and $\sigma=4 \mathrm{~S} / \mathrm{m}$.

## Solution:

$$
\begin{aligned}
\delta_{\mathrm{s}} & =\frac{1}{\alpha}=\frac{1}{\omega}\left[\frac{\mu \varepsilon^{\prime}}{2}\left[\sqrt{1+\left(\frac{\varepsilon^{\prime \prime}}{\varepsilon^{\prime}}\right)^{2}}-1\right]\right]^{-1 / 2}, \\
\omega & =2 \pi f, \\
\mu \varepsilon^{\prime} & =\mu_{0} \varepsilon_{0} \varepsilon_{\mathrm{r}}=\frac{\varepsilon_{\mathrm{r}}}{c^{2}}=\frac{80}{c^{2}}=\frac{80}{\left(3 \times 10^{8}\right)^{2}}, \\
\frac{\varepsilon^{\prime \prime}}{\varepsilon^{\prime}} & =\frac{\sigma}{\omega \varepsilon}=\frac{\sigma}{\omega \varepsilon_{0} \varepsilon_{\mathrm{r}}}=\frac{4 \times 36 \pi}{2 \pi f \times 10^{-9} \times 80}=\frac{72}{80 f} \times 10^{9} .
\end{aligned}
$$

See Fig. P7.17 for plot of $\delta_{\mathrm{s}}$ versus frequency.


Figure P7.17: Skin depth versus frequency for seawater.

Problem 7.18 Ignoring reflection at the air-soil boundary, if the amplitude of a $3-\mathrm{GHz}$ incident wave is $10 \mathrm{~V} / \mathrm{m}$ at the surface of a wet soil medium, at what depth will it be down to $1 \mathrm{mV} / \mathrm{m}$ ? Wet soil is characterized by $\mu_{\mathrm{r}}=1, \varepsilon_{\mathrm{r}}=9$, and $\sigma=5 \times 10^{-4}$ S/m.

## Solution:

$$
\begin{gathered}
E(z)=E_{0} e^{-\alpha z}=10 e^{-\alpha z} \\
\frac{\sigma}{\omega \varepsilon}=\frac{5 \times 10^{-4} \times 36 \pi}{2 \pi \times 3 \times 10^{9} \times 10^{-9} \times 9}=3.32 \times 10^{-4} .
\end{gathered}
$$

Hence, medium is a low-loss dielectric.

$$
\begin{aligned}
\alpha & =\frac{\sigma}{2} \sqrt{\frac{\mu}{\varepsilon}}=\frac{\sigma}{2} \cdot \frac{120 \pi}{\sqrt{\varepsilon_{\mathrm{r}}}}=\frac{5 \times 10^{-4} \times 120 \pi}{2 \times \sqrt{9}}=0.032 \quad(\mathrm{~Np} / \mathrm{m}), \\
10^{-3} & =10 e^{-0.032 z}, \quad \ln 10^{-4}=-0.032 z, \\
z & =287.82 \mathrm{~m} .
\end{aligned}
$$

Problem 7.19 The skin depth of a certain nonmagnetic conducting material is $3 \mu \mathrm{~m}$ at 5 GHz . Determine the phase velocity in the material.
Solution: For a good conductor, $\alpha=\beta$, and for any material $\delta_{s}=1 / \alpha$. Hence,

$$
u_{\mathrm{p}}=\frac{\omega}{\beta}=\frac{2 \pi f}{\beta}=2 \pi f \delta_{\mathrm{s}}=2 \pi \times 5 \times 10^{9} \times 3 \times 10^{-6}=9.42 \times 10^{4} \quad(\mathrm{~m} / \mathrm{s}) .
$$

Problem 7.20 Based on wave attenuation and reflection measurements conducted at 1 MHz , it was determined that the intrinsic impedance of a certain medium is $28.1 \angle 45^{\circ}(\Omega)$ and the skin depth is 2 m . Determine (a) the conductivity of the material, (b) the wavelength in the medium, and (c) the phase velocity.

## Solution:

(a) Since the phase angle of $\eta_{c}$ is $45^{\circ}$, the material is a good conductor. Hence,

$$
\eta_{\mathrm{c}}=(1+j) \frac{\alpha}{\sigma}=28.1 e^{j 45^{\circ}}=28.1 \cos 45^{\circ}+j 28.1 \sin 45^{\circ},
$$

or

$$
\frac{\alpha}{\sigma}=28.1 \cos 45^{\circ}=19.87 .
$$

Since $\alpha=1 / \delta_{s}=1 / 2=0.5 \mathrm{~Np} / \mathrm{m}$,

$$
\sigma=\frac{\alpha}{19.87}=\frac{0.5}{19.87}=2.52 \times 10^{-2} \mathrm{~S} / \mathrm{m}
$$

(b) Since $\alpha=\beta$ for a good conductor, and $\alpha=0.5$, it follows that $\beta=0.5$. Therefore,

$$
\lambda=\frac{2 \pi}{\beta}=\frac{2 \pi}{0.5}=4 \pi=12.57 \mathrm{~m} .
$$

(c) $u_{\mathrm{p}}=f \lambda=10^{6} \times 12.57=1.26 \times 10^{7} \mathrm{~m} / \mathrm{s}$.

Problem 7.21 The electric field of a plane wave propagating in a nonmagnetic medium is given by

$$
\mathbf{E}=\hat{\mathbf{z}} 25 e^{-30 x} \cos \left(2 \pi \times 10^{9} t-40 x\right) \quad(\mathrm{V} / \mathrm{m})
$$

Obtain the corresponding expression for $\mathbf{H}$.
Solution: From the given expression for $\mathbf{E}$,

$$
\begin{aligned}
& \omega=2 \pi \times 10^{9} \quad(\mathrm{rad} / \mathrm{s}), \\
& \alpha=30 \quad(\mathrm{~Np} / \mathrm{m}), \\
& \beta=40 \quad(\mathrm{rad} / \mathrm{m}) .
\end{aligned}
$$

From (7.65a) and (7.65b),

$$
\begin{aligned}
\alpha^{2}-\beta^{2} & =-\omega^{2} \mu \varepsilon^{\prime}=-\omega^{2} \mu_{0} \varepsilon_{0} \varepsilon_{\mathrm{r}}^{\prime}=-\frac{\omega^{2}}{c^{2}} \varepsilon_{\mathrm{r}}^{\prime}, \\
2 \alpha \beta & =\omega^{2} \mu \varepsilon^{\prime \prime}=\frac{\omega^{2}}{c^{2}} \varepsilon_{\mathrm{r}}^{\prime \prime} .
\end{aligned}
$$

Using the above values for $\omega, \alpha$, and $\beta$, we obtain the following:

$$
\begin{align*}
& \varepsilon_{\mathrm{r}}^{\prime}=1.6, \\
& \varepsilon_{\mathrm{r}}^{\prime \prime}=5.47 \\
& \eta_{\mathrm{c}}=\sqrt{\frac{\mu}{\varepsilon^{\prime}}}\left(1-j \frac{\varepsilon^{\prime \prime}}{\varepsilon^{\prime}}\right)^{-1 / 2} \\
&= \eta_{0} \\
& \sqrt{\varepsilon_{\mathrm{r}}^{\prime}} \\
&\left(1-j \frac{\varepsilon_{\mathrm{r}}^{\prime \prime}}{\varepsilon_{\mathrm{r}}^{\prime}}\right)^{-1 / 2}=\frac{377}{\sqrt{1.6}}\left(1-j \frac{5.47}{1.6}\right)^{-1 / 2}=157.9 e^{j 36.85^{\circ}}
\end{align*}
$$

$$
\begin{aligned}
& \widetilde{\mathbf{E}}=\hat{\mathbf{z}} 25 e^{-30 x} e^{-j 40 x}, \\
& \widetilde{\mathbf{H}}=\frac{1}{\eta_{\mathrm{c}}} \hat{\mathbf{k}} \times \widetilde{\mathbf{E}}=\frac{1}{157.9 e^{j 36.855^{\circ}} \hat{\mathbf{x}} \times \hat{\mathbf{z}} 25 e^{-30 x} e^{-j 40 x}=-\hat{\mathbf{y}} 0.16 e^{-30 x} e^{-40 x} e^{-j 36.85^{\circ}},} \\
& \mathbf{H}=\mathfrak{R e}\left\{\widetilde{\mathbf{H}} e^{j \omega t}\right\}=-\hat{\mathbf{y}} 0.16 e^{-30 x} \cos \left(2 \pi \times 10^{9} t-40 x-36.85^{\circ}\right) \quad(\mathrm{A} / \mathrm{m}) .
\end{aligned}
$$

## Section 7-5: Current Flow in Conductors

Problem 7.22 In a nonmagnetic, lossy, dielectric medium, a $300-\mathrm{MHz}$ plane wave is characterized by the magnetic field phasor

$$
\tilde{\mathbf{H}}=(\hat{\mathbf{x}}-j 4 \hat{\mathbf{z}}) e^{-2 y} e^{-j 9 y} \quad(\mathrm{~A} / \mathrm{m}) .
$$

Obtain time-domain expressions for the electric and magnetic field vectors.

## Solution:

$$
\widetilde{\mathbf{E}}=-\eta_{\mathrm{c}} \hat{\mathbf{k}} \times \widetilde{\mathbf{H}} .
$$

To find $\eta_{\mathrm{c}}$, we need $\varepsilon^{\prime}$ and $\varepsilon^{\prime \prime}$. From the given expression for $\widetilde{\mathbf{H}}$,

$$
\begin{array}{ll}
\alpha=2 & (\mathrm{~Np} / \mathrm{m}) \\
\beta=9 & (\mathrm{rad} / \mathrm{m}) .
\end{array}
$$

Also, we are given than $f=300 \mathrm{MHz}=3 \times 10^{8} \mathrm{~Hz}$. From (7.65a),

$$
\begin{aligned}
\alpha^{2}-\beta^{2} & =-\omega^{2} \mu \varepsilon^{\prime}, \\
4-81 & =-\left(2 \pi \times 3 \times 10^{8}\right)^{2} \times 4 \pi \times 10^{-7} \times \varepsilon_{\mathrm{r}}^{\prime} \times \frac{10^{-9}}{36 \pi},
\end{aligned}
$$

whose solution gives

$$
\varepsilon_{\mathrm{r}}^{\prime}=1.95
$$

Similarly, from (7.65b),

$$
\begin{aligned}
2 \alpha \beta & =\omega^{2} \mu \varepsilon^{\prime \prime}, \\
2 \times 2 \times 9 & =\left(2 \pi \times 3 \times 10^{8}\right)^{2} \times 4 \pi \times 10^{-7} \times \varepsilon_{\mathrm{r}}^{\prime \prime} \times \frac{10^{-9}}{36 \pi}
\end{aligned}
$$

which gives

$$
\varepsilon_{\mathrm{r}}^{\prime \prime}=0.91
$$

$$
\begin{aligned}
\eta_{\mathrm{c}} & =\sqrt{\frac{\mu}{\varepsilon^{\prime}}}\left(1-j \frac{\varepsilon^{\prime \prime}}{\varepsilon^{\prime}}\right)^{-1 / 2} \\
& =\frac{\eta_{0}}{\sqrt{\varepsilon_{\mathrm{r}}^{\prime}}}\left(1-j \frac{0.91}{1.95}\right)^{-1 / 2}=\frac{377}{\sqrt{1.95}}(0.93+j 0.21)=256.9 e^{j 12.6^{\circ}}
\end{aligned}
$$

Hence,

$$
\begin{aligned}
\widetilde{\mathbf{E}}= & -256.9 e^{j 12.6^{\circ}} \hat{\mathbf{y}} \times(\hat{\mathbf{x}}-j 4 \hat{\mathbf{z}}) e^{-2 y} e^{-j 9 y} \\
= & (\hat{\mathbf{x}} j 4+\hat{\mathbf{z}}) 256.9 e^{-2 y} e^{-j 9 y} e^{j 12.6^{\circ}} \\
= & \left(\hat{\mathbf{x}} 4 e^{j \pi / 2}+\hat{\mathbf{z}}\right) 256.9 e^{-2 y} e^{-j 9 y} e^{j 12.6^{\circ}}, \\
\mathbf{E}= & \mathfrak{R e}\left\{\widetilde{\mathbf{E}} e^{j \omega t}\right\} \\
= & \hat{\mathbf{x}} 1.03 \times 10^{3} e^{-2 y} \cos \left(\omega t-9 y+102.6^{\circ}\right) \\
& +\hat{\mathbf{z}} 256.9 e^{-2 y} \cos \left(\omega t-9 y+12.6^{\circ}\right) \quad(\mathrm{V} / \mathrm{m}), \\
\mathbf{H}= & \mathfrak{R e}\left\{\widetilde{\mathbf{H}} e^{j \omega t}\right\} \\
= & \mathfrak{R e}\left\{(\hat{\mathbf{x}}+j 4 \hat{\mathbf{z}}) e^{-2 y} e^{-j 9 y} e^{j \omega t}\right\} \\
= & \hat{\mathbf{x}} e^{-2 y} \cos (\omega t-9 y)+\hat{\mathbf{z}} 4 e^{-2 y} \sin (\omega t-9 y) \quad(\mathrm{A} / \mathrm{m}) .
\end{aligned}
$$

Problem 7.23 A rectangular copper block is 30 cm in height (along $z$ ). In response to a wave incident upon the block from above, a current is induced in the block in the positive $x$-direction. Determine the ratio of the a-c resistance of the block to its d-c resistance at 1 kHz . The relevant properties of copper are given in Appendix B.


Figure P7.23: Copper block of Problem 7.23.

## Solution:

$$
\begin{gathered}
\mathrm{d}-\mathrm{c} \text { resistance } R_{\mathrm{dc}}=\frac{l}{\sigma A}=\frac{l}{0.3 \sigma w}, \\
\mathrm{a}-\mathrm{c} \text { resistance } R_{\mathrm{ac}}=\frac{l}{\sigma w \delta_{\mathrm{s}}} . \\
\frac{R_{\mathrm{ac}}}{R_{\mathrm{dc}}}=\frac{0.3}{\delta_{\mathrm{s}}}=0.3 \sqrt{\pi f \mu \sigma}=0.3\left[\pi \times 10^{3} \times 4 \pi \times 10^{-7} \times 5.8 \times 10^{7}\right]^{1 / 2}=143.55 .
\end{gathered}
$$

Problem 7.24 The inner and outer conductors of a coaxial cable have radii of 0.5 cm and 1 cm , respectively. The conductors are made of copper with $\varepsilon_{\mathrm{r}}=1$, $\mu_{\mathrm{r}}=1$ and $\sigma=5.8 \times 10^{7} \mathrm{~S} / \mathrm{m}$, and the outer conductor is 0.5 mm thick. At 10 MHz :
(a) Are the conductors thick enough to be considered infinitely thick so far as the flow of current through them is concerned?
(b) Determine the surface resistance $R_{\mathrm{s}}$.
(c) Determine the a-c resistance per unit length of the cable.

Solution:
(a) From Eqs. (7.72) and (7.77b),

$$
\delta_{\mathrm{s}}=[\pi f \mu \sigma]^{-1 / 2}=\left[\pi \times 10^{7} \times 4 \pi \times 10^{-7} \times 5.8 \times 10^{7}\right]^{-1 / 2}=0.021 \mathrm{~mm}
$$

Hence,

$$
\frac{d}{\delta_{\mathrm{s}}}=\frac{0.5 \mathrm{~mm}}{0.021 \mathrm{~mm}} \approx 25 .
$$

Hence, conductor is plenty thick.
(b) From Eq. (7.92a),

$$
R_{\mathrm{s}}=\frac{1}{\sigma \delta_{\mathrm{s}}}=\frac{1}{5.8 \times 10^{7} \times 2.1 \times 10^{-5}}=8.2 \times 10^{-4} \Omega
$$

(c) From Eq. (7.96),

$$
R^{\prime}=\frac{R_{\mathrm{S}}}{2 \pi}\left(\frac{1}{a}+\frac{1}{b}\right)=\frac{8.2 \times 10^{-4}}{2 \pi}\left(\frac{1}{5 \times 10^{-3}}+\frac{1}{10^{-2}}\right)=0.039 \quad(\Omega / \mathrm{m})
$$

## Section 7-6: EM Power Density

Problem 7.25 The magnetic field of a plane wave traveling in air is given by $\mathbf{H}=\hat{\mathbf{x}} 50 \sin \left(2 \pi \times 10^{7} t-k y\right)(\mathrm{mA} / \mathrm{m})$. Determine the average power density carried by the wave.

## Solution:

$$
\begin{aligned}
\mathbf{H} & =\hat{\mathbf{x}} 50 \sin \left(2 \pi \times 10^{7} t-k y\right) \quad(\mathrm{mA} / \mathrm{m}), \\
\mathbf{E} & =-\eta_{0} \hat{\mathbf{y}} \times \mathbf{H}=\hat{\mathbf{z}} \eta_{0} 50 \sin \left(2 \pi \times 10^{7} t-k y\right) \quad(\mathrm{mV} / \mathrm{m}), \\
\mathbf{S}_{\mathrm{av}} & =(\hat{\mathbf{z}} \times \hat{\mathbf{x}}) \frac{\eta_{0}(50)^{2}}{2} \times 10^{-6}=\hat{\mathbf{y}} \frac{120 \pi}{2}(50)^{2} \times 10^{-6}=\hat{\mathbf{y}} 0.48 \quad\left(\mathrm{~W} / \mathrm{m}^{2}\right) .
\end{aligned}
$$

Problem 7.26 A wave traveling in a nonmagnetic medium with $\varepsilon_{\mathrm{r}}=9$ is characterized by an electric field given by

$$
\mathbf{E}=\left[\hat{\mathbf{y}} 3 \cos \left(\pi \times 10^{7} t+k x\right)-\hat{\mathbf{z}} 2 \cos \left(\pi \times 10^{7} t+k x\right)\right] \quad(\mathrm{V} / \mathrm{m}) .
$$

Determine the direction of wave travel and the average power density carried by the wave.

## Solution:

$$
\eta \simeq \frac{\eta_{0}}{\sqrt{\varepsilon_{\mathrm{r}}}}=\frac{120 \pi}{\sqrt{9}}=40 \pi
$$

The wave is traveling in the negative $x$-direction.

$$
\mathbf{S}_{\mathrm{av}}=-\hat{\mathbf{x}} \frac{\left[3^{2}+2^{2}\right]}{2 \eta}=-\hat{\mathbf{x}} \frac{13}{2 \times 40 \pi}=-\hat{\mathbf{x}} 0.05 \quad\left(\mathrm{~W} / \mathrm{m}^{2}\right)
$$

Problem 7.27 The electric-field phasor of a uniform plane wave traveling downward in water is given by

$$
\widetilde{\mathbf{E}}=\hat{\mathbf{x}} 5 e^{-0.2 z} e^{-j 0.2 z} \quad(\mathrm{~V} / \mathrm{m}),
$$

where $\hat{\mathbf{z}}$ is the downward direction and $z=0$ is the water surface. If $\sigma=4 \mathrm{~S} / \mathrm{m}$,
(a) obtain an expression for the average power density,
(b) determine the attenuation rate, and
(c) determine the depth at which the power density has been reduced by 40 dB .

## Solution:

(a) Since $\alpha=\beta=0.2$, the medium is a good conductor.

$$
\eta_{\mathrm{c}}=(1+j) \frac{\alpha}{\sigma}=(1+j) \frac{0.2}{4}=(1+j) 0.05=0.0707 e^{j 45^{\circ}}
$$

From Eq. (7.109),

$$
\mathbf{S}_{\mathrm{av}}=\hat{\mathbf{z}} \frac{\left|E_{0}\right|^{2}}{2\left|\eta_{\mathrm{c}}\right|^{-2 \alpha z} \cos \theta_{\eta}=\hat{\mathbf{z}}} \frac{25}{2 \times 0.0707} e^{-0.4 z} \cos 45^{\circ}=\hat{\mathbf{z}} 125 e^{-0.4 z} \quad\left(\mathrm{~W} / \mathrm{m}^{2}\right)
$$

(b) $A=-8.68 \alpha z=-8.68 \times 0.2 z=-1.74 z(\mathrm{~dB})$.
(c) 40 dB is equivalent to $10^{-4}$. Hence,

$$
10^{-4}=e^{-2 \alpha z}=e^{-0.4 z}, \quad \ln \left(10^{-4}\right)=-0.4 z,
$$

or $z=23.03 \mathrm{~m}$.

Problem 7.28 The amplitudes of an elliptically polarized plane wave traveling in a lossless, nonmagnetic medium with $\varepsilon_{\mathrm{r}}=4$ are $H_{y 0}=3(\mathrm{~mA} / \mathrm{m})$ and $H_{z 0}=4(\mathrm{~mA} / \mathrm{m})$. Determine the average power flowing through an aperture in the $y-z$ plane if its area is $20 \mathrm{~m}^{2}$.

## Solution:

$$
\begin{aligned}
\eta & =\frac{\eta_{0}}{\sqrt{\varepsilon_{\mathrm{r}}}}=\frac{120 \pi}{\sqrt{4}}=60 \pi=188.5 \Omega, \\
\mathbf{S}_{\mathrm{av}} & =\hat{\mathbf{x}} \frac{\eta}{2}\left[H_{y 0}^{2}+H_{x 0}^{2}\right]=\hat{\mathbf{x}} \frac{188.5}{2}[9+16] \times 10^{-6}=2.36 \quad\left(\mathrm{~mW} / \mathrm{m}^{2}\right), \\
P & =S_{\mathrm{av}} A=2.36 \times 10^{-3} \times 20=47.13 \quad(\mathrm{~mW}) .
\end{aligned}
$$

Problem 7.29 A wave traveling in a lossless, nonmagnetic medium has an electric field amplitude of $24.56 \mathrm{~V} / \mathrm{m}$ and an average power density of $2.4 \mathrm{~W} / \mathrm{m}^{2}$. Determine the phase velocity of the wave.

## Solution:

$$
S_{\mathrm{av}}=\frac{\left|E_{0}\right|^{2}}{2 \eta}, \quad \eta=\frac{\left|E_{0}\right|^{2}}{2 S_{\mathrm{av}}},
$$

or

$$
\eta=\frac{(24.56)^{2}}{2 \times 2.4}=125.67 \Omega .
$$

But

$$
\eta=\frac{\eta_{0}}{\sqrt{\varepsilon_{\mathrm{r}}}}=\frac{377}{\sqrt{\varepsilon_{\mathrm{r}}}}, \quad \varepsilon_{\mathrm{r}}=\left(\frac{377}{125.67}\right)^{2}=9 .
$$

Hence,

$$
u_{\mathrm{p}}=\frac{c}{\sqrt{\varepsilon_{\mathrm{r}}}}=\frac{3 \times 10^{8}}{3}=1 \times 10^{8} \mathrm{~m} / \mathrm{s}
$$

Problem 7.30 At microwave frequencies, the power density considered safe for human exposure is $1\left(\mathrm{~mW} / \mathrm{cm}^{2}\right)$. A radar radiates a wave with an electric field amplitude $E$ that decays with distance as $E(R)=(3,000 / R)(\mathrm{V} / \mathrm{m})$, where $R$ is the distance in meters. What is the radius of the unsafe region?

## Solution:

$$
\begin{aligned}
S_{\mathrm{av}} & =\frac{|E(R)|^{2}}{2 \eta_{0}}, \quad 1\left(\mathrm{~mW} / \mathrm{cm}^{2}\right)=10^{-3} \mathrm{~W} / \mathrm{cm}^{2}=10 \mathrm{~W} / \mathrm{m}^{2} \\
10 & =\left(\frac{3 \times 10^{3}}{R}\right)^{2} \times \frac{1}{2 \times 120 \pi}=\frac{1.2 \times 10^{4}}{R^{2}} \\
R & =\left(\frac{1.2 \times 10^{4}}{10}\right)^{1 / 2}=34.64 \mathrm{~m}
\end{aligned}
$$

Problem 7.31 Consider the imaginary rectangular box shown in Fig. 7-19 (P7.31).
(a) Determine the net power flux $P(t)$ entering the box due to a plane wave in air given by

$$
\mathbf{E}=\hat{\mathbf{x}} E_{0} \cos (\omega t-k y) \quad(\mathrm{V} / \mathrm{m})
$$

(b) Determine the net time-average power entering the box.

## Solution:

(a)

$$
\begin{gathered}
\mathbf{E}=\hat{\mathbf{x}} E_{0} \cos (\omega t-k y), \\
\mathbf{H}=-\hat{\mathbf{z}} \frac{E_{0}}{\eta_{0}} \cos (\omega t-k y) . \\
\mathbf{S}(t)=\mathbf{E} \times \mathbf{H}=\hat{\mathbf{y}} \frac{E_{0}^{2}}{\eta_{0}} \cos ^{2}(\omega t-k y), \\
P(t)=\left.S(t) A\right|_{y=0}-\left.S(t) A\right|_{y=b}=\frac{E_{0}^{2}}{\eta_{0}} a c\left[\cos ^{2} \omega t-\cos ^{2}(\omega t-k b)\right] .
\end{gathered}
$$



Figure P7.31: Imaginary rectangular box of Problems 7.31 and 7.32 .
(b)

$$
P_{\mathrm{av}}=\frac{1}{T} \int_{0}^{T} P(t) d t .
$$

where $T=2 \pi / \omega$.

$$
P_{\mathrm{av}}=\frac{E_{0}^{2} a c}{\eta_{0}}\left\{\frac{\omega}{2 \pi} \int_{0}^{2 \pi / \omega}\left[\cos ^{2} \omega t-\cos ^{2}(\omega t-k b)\right] d t\right\}=0 .
$$

Net average energy entering the box is zero, which is as expected since the box is in a lossless medium (air).

Problem 7.32 Repeat Problem 7.31 for a wave traveling in a lossy medium in which

$$
\begin{aligned}
& \mathbf{E}=\hat{\mathbf{x}} 100 e^{-20 y} \cos \left(2 \pi \times 10^{9} t-40 y\right) \quad(\mathrm{V} / \mathrm{m}) \\
& \mathbf{H}=-\hat{\mathbf{z}} 0.64 e^{-20 y} \cos \left(2 \pi \times 10^{9} t-40 y-36.85^{\circ}\right) \quad(\mathrm{A} / \mathrm{m})
\end{aligned}
$$

The box has dimensions $A=1 \mathrm{~cm}, b=2 \mathrm{~cm}$, and $c=0.5 \mathrm{~cm}$.

## Solution:

(a)

$$
\begin{aligned}
\mathbf{S}(t)= & \mathbf{E} \times \mathbf{H} \\
= & \hat{\mathbf{x}} 100 e^{-20 y} \cos \left(2 \pi \times 10^{9} t-40 y\right) \\
& \times(-\hat{\mathbf{z}} 0.64) e^{-20 y} \cos \left(2 \pi \times 10^{9} t-40 y-36.85^{\circ}\right) \\
= & \hat{\mathbf{y}} 64 e^{-40 y} \cos \left(2 \pi \times 10^{9} t-40 y\right) \cos \left(2 \pi \times 10^{9} t-40 y-36.85^{\circ}\right) .
\end{aligned}
$$

Using the identity $\cos \theta \cos \phi=\frac{1}{2}[\cos (\theta+\phi)+\cos (\theta-\phi)]$,

$$
\begin{aligned}
S(t)= & \frac{64}{2} e^{-40 y}\left[\cos \left(4 \pi \times 10^{9} t-80 y-36.85^{\circ}\right)+\cos 36.85^{\circ}\right], \\
P(t)= & \left.S(t) A\right|_{y=0}-\left.S(t) A\right|_{y=b} \\
= & 32 a c\left\{\left[\cos \left(4 \pi \times 10^{9} t-36.85^{\circ}\right)+\cos 36.85^{\circ}\right]\right. \\
& \left.-e^{-40 b}\left[\cos \left(4 \pi \times 10^{9} t-80 y-36.85^{\circ}\right)+\cos 36.85^{\circ}\right]\right\} .
\end{aligned}
$$

(b)

$$
P_{\mathrm{av}}=\frac{1}{T} \int_{0}^{T} P(t) d t=\frac{\omega}{2 \pi} \int_{0}^{2 \pi / \omega} P(t) d t .
$$

The average of $\cos (\omega t+\theta)$ over a period $T$ is equal to zero, regardless of the value of $\theta$. Hence,

$$
P_{\mathrm{av}}=32 a c\left(1-e^{-40 b}\right) \cos 36.85^{\circ} .
$$

With $a=1 \mathrm{~cm}, b=2 \mathrm{~cm}$, and $c=0.5 \mathrm{~cm}$,

$$
P_{\mathrm{av}}=7.05 \times 10^{-4} \quad(\mathrm{~W})
$$

This is the average power absorbed by the lossy material in the box.

Problem 7.33 Given a wave with

$$
\mathbf{E}=\hat{\mathbf{x}} E_{0} \cos (\omega t-k z),
$$

calculate:
(a) the time-average electric energy density

$$
\left(w_{\mathrm{e}}\right)_{\mathrm{av}}=\frac{1}{T} \int_{0}^{T} w_{\mathrm{e}} d t=\frac{1}{2 T} \int_{0}^{T} \varepsilon E^{2} d t
$$

(b) the time-average magnetic energy density

$$
\left(w_{\mathrm{m}}\right)_{\mathrm{av}}=\frac{1}{T} \int_{0}^{T} w_{\mathrm{m}} d t=\frac{1}{2 T} \int_{0}^{T} \mu H^{2} d t
$$

and
(c) show that $\left(w_{\mathrm{e}}\right)_{\mathrm{av}}=\left(w_{\mathrm{m}}\right)_{\mathrm{av}}$.

## Solution:

(a)

$$
\left(w_{\mathrm{e}}\right)_{\mathrm{av}}=\frac{1}{2 T} \int_{0}^{T} \varepsilon E_{0}^{2} \cos ^{2}(\omega t-k z) d t .
$$

With $T=\frac{2 \pi}{\omega}$,

$$
\begin{aligned}
\left(w_{\mathrm{e}}\right)_{\mathrm{av}} & =\frac{\omega \varepsilon E_{0}^{2}}{4 \pi} \int_{0}^{2 \pi / \omega} \cos ^{2}(\omega t-k z) d t \\
& =\frac{\varepsilon E_{0}^{2}}{4 \pi} \int_{0}^{2 \pi} \cos ^{2}(\omega t-k z) d(\omega t) \\
& =\frac{\varepsilon E_{0}^{2}}{4} .
\end{aligned}
$$

(b)

$$
\begin{aligned}
& \mathbf{H}=\hat{\mathbf{y}} \frac{E_{0}}{\eta} \cos (\omega t-k z) . \\
\left(w_{\mathrm{m}}\right)_{\mathrm{av}} & =\frac{1}{2 T} \int_{0}^{T} \mu H^{2} d t \\
& =\frac{1}{2 T} \int_{0}^{T} \mu \frac{E_{0}^{2}}{\eta^{2}} \cos ^{2}(\omega t-k z) d t \\
& =\frac{\mu E_{0}^{2}}{4 \eta^{2}} .
\end{aligned}
$$

(c)

$$
\left(w_{\mathrm{m}}\right)_{\mathrm{av}}=\frac{\mu E_{0}^{2}}{4 \eta^{2}}=\frac{\mu E_{0}^{2}}{4\left(\frac{\mu}{\varepsilon}\right)}=\frac{\varepsilon E_{0}^{2}}{4}=\left(w_{\mathrm{e}}\right)_{\mathrm{av}} .
$$

Problem 7.34 A $60-\mathrm{MHz}$ plane wave traveling in the $-x$-direction in dry soil with relative permittivity $\varepsilon_{\mathrm{r}}=4$ has an electric field polarized along the $z$-direction. Assuming dry soil to be approximately lossless, and given that the magnetic field has a peak value of $10(\mathrm{~mA} / \mathrm{m})$ and that its value was measured to be $7(\mathrm{~mA} / \mathrm{m})$ at $t=0$ and $x=-0.75 \mathrm{~m}$, develop complete expressions for the wave's electric and magnetic fields.
Solution: For $f=60 \mathrm{MHz}=6 \times 10^{7} \mathrm{~Hz}, \varepsilon_{\mathrm{r}}=4, \mu_{\mathrm{r}}=1$,

$$
k=\frac{\omega}{c} \sqrt{\varepsilon_{\mathrm{r}}}=\frac{2 \pi \times 6 \times 10^{7}}{3 \times 10^{8}} \sqrt{4}=0.8 \pi \quad(\mathrm{rad} / \mathrm{m})
$$

Given that $\mathbf{E}$ points along $\hat{\mathbf{z}}$ and wave travel is along $-\hat{\mathbf{x}}$, we can write

$$
\mathbf{E}(x, t)=\hat{\mathbf{z}} E_{0} \cos \left(2 \pi \times 60 \times 10^{6} t+0.8 \pi x+\phi_{0}\right) \quad(\mathrm{V} / \mathrm{m})
$$

where $E_{0}$ and $\phi_{0}$ are unknown constants at this time. The intrinsic impedance of the medium is

$$
\eta=\frac{\eta_{0}}{\sqrt{\varepsilon_{\mathrm{r}}}}=\frac{120 \pi}{2}=60 \pi
$$

With $\mathbf{E}$ along $\hat{\mathbf{z}}$ and $\hat{\mathbf{k}}$ along $-\hat{\mathbf{x}}$, (7.39) gives

$$
\mathbf{H}=\frac{1}{\eta} \hat{\mathbf{k}} \times \mathbf{E}
$$

or

$$
\mathbf{H}(x, t)=\hat{\mathbf{y}} \frac{E_{0}}{\eta} \cos \left(1.2 \pi \times 10^{8} t+0.8 \pi x+\phi_{0}\right) \quad(\mathrm{A} / \mathrm{m}) .
$$

Hence,

$$
\begin{aligned}
& \frac{E_{0}}{\eta}=10 \quad(\mathrm{~mA} / \mathrm{m}) \\
& E_{0}=10 \times 60 \pi \times 10^{-3}=0.6 \pi \quad(\mathrm{~V} / \mathrm{m})
\end{aligned}
$$

Also,

$$
H(-0.75 \mathrm{~m}, 0)=7 \times 10^{-3}=10 \cos \left(-0.8 \pi \times 0.75+\phi_{0}\right) \times 10^{-3}
$$

which leads to $\phi_{0}=153.6^{\circ}$.
Hence,

$$
\begin{aligned}
& \mathbf{E}(x, t)=\hat{\mathbf{z}} 0.6 \pi \cos \left(1.2 \pi \times 10^{8} t+0.8 \pi x+153.6^{\circ}\right) \quad(\mathrm{V} / \mathrm{m}) . \\
& \mathbf{H}(x, t)=\hat{\mathbf{y}} 10 \cos \left(1.2 \pi \times 10^{8} t+0.8 \pi x+153.6^{\circ}\right) \quad(\mathrm{mA} / \mathrm{m}) .
\end{aligned}
$$

Problem 7.35 At 2 GHz , the conductivity of meat is on the order of $1(\mathrm{~S} / \mathrm{m})$. When a material is placed inside a microwave oven and the field is activated, the presence of the electromagnetic fields in the conducting material causes energy dissipation in the material in the form of heat.
(a) Develop an expression for the time-average power per $\mathrm{mm}^{3}$ dissipated in a material of conductivity $\sigma$ if the peak electric field in the material is $E_{0}$.
(b) Evaluate the result for meat with $E_{0}=4 \times 10^{4}(\mathrm{~V} / \mathrm{m})$.

## Solution:

(a) Let us consider a small volume of the material in the shape of a box of length $d$ and cross sectional area $A$. Let us assume the microwave oven creates a wave traveling along the $z$ direction with $\mathbf{E}$ along $y$, as shown.


Along $y$, the $\mathbf{E}$ field will create a voltage difference across the length of the box $V$, where

$$
V=E d
$$

Conduction current through the cross sectional area $A$ is

$$
I=J A=\sigma E A
$$

Hence, the instantaneous power is

$$
\begin{aligned}
P=I V & =\sigma E^{2}(A d) \\
& =\sigma E^{2} \mathcal{V}
\end{aligned}
$$

where $\mathcal{v}=A d$ is the small volume under consideration. The power per $\mathrm{mm}^{3}$ is obtained by setting $\mathcal{V}=\left(10^{-3}\right)^{3}$,

$$
P^{\prime}=\frac{P}{10^{-9}}=\sigma E^{2} \times 10^{-9} \quad\left(\mathrm{~W} / \mathrm{mm}^{3}\right)
$$

As a time harmonic signal, $E=E_{0} \cos \omega t$. The time average dissipated power is

$$
\begin{aligned}
P_{\mathrm{av}}^{\prime} & =\left[\frac{1}{T} \int_{0}^{T} \sigma E_{0}^{2} \cos ^{2} \omega t d t\right] \times 10^{-9} \\
& =\frac{1}{2} \sigma E_{0}^{2} \times 10^{-9} \quad\left(\mathrm{~W} / \mathrm{mm}^{3}\right)
\end{aligned}
$$

(b)

$$
P_{\mathrm{av}}^{\prime}=\frac{1}{2} \times 1 \times\left(4 \times 10^{4}\right) 2 \times 10^{-9}=0.8 \quad\left(\mathrm{~W} / \mathrm{mm}^{3}\right)
$$

Problem 7.36 A team of scientists is designing a radar as a probe for measuring the depth of the ice layer over the antarctic land mass. In order to measure a detectable echo due to the reflection by the ice-rock boundary, the thickness of the ice sheet should not exceed three skin depths. If $\varepsilon_{\mathrm{r}}^{\prime}=3$ and $\varepsilon_{\mathrm{r}}^{\prime \prime}=10^{-2}$ for ice and if the maximum anticipated ice thickness in the area under exploration is 1.2 km , what frequency range is useable with the radar?

## Solution:

$$
\begin{aligned}
3 \delta_{\mathrm{s}} & =1.2 \mathrm{~km}=1200 \mathrm{~m} \\
\delta_{\mathrm{s}} & =400 \mathrm{~m} .
\end{aligned}
$$

Hence,

$$
\alpha=\frac{1}{\delta_{\mathrm{s}}}=\frac{1}{400}=2.5 \times 10^{-3} \quad(\mathrm{~Np} / \mathrm{m})
$$

Since $\varepsilon^{\prime \prime} / \varepsilon^{\prime} \ll 1$, we can use (7.75a) for $\alpha$ :

$$
\alpha=\frac{\omega \varepsilon^{\prime \prime}}{2} \sqrt{\frac{\mu}{\varepsilon^{\prime}}}=\frac{2 \pi f \varepsilon_{\mathrm{r}}^{\prime \prime} \varepsilon_{0}}{2 \sqrt{\varepsilon_{\mathrm{r}}^{\prime}} \sqrt{\varepsilon_{0}}} \sqrt{\mu_{0}}=\frac{\pi f \varepsilon_{\mathrm{r}}^{\prime \prime}}{c \sqrt{\varepsilon_{\mathrm{r}}}}=\frac{\pi f \times 10^{-2}}{3 \times 10^{8} \sqrt{3}}=6 f \times 10^{-11} \mathrm{~Np} / \mathrm{m} .
$$

For $\alpha=2.5 \times 10^{-3}=6 f \times 10^{-11}$,

$$
f=41.6 \mathrm{MHz} .
$$

Since $\alpha$ increases with increasing frequency, the useable frequency range is

$$
f \leq 41.6 \mathrm{MHz}
$$

