

**AMERICAN UNIVERSITY OF BEIRUT**  
**ELECTRICAL AND COMPUTER ENGINEERING DEPARTMENT**

**EECE 340 – Signals and Systems**  
**QUIZ 1 - Solution**

**Problem 1: (9 pts)**

Categorize each of the following signals as an energy signal or a power signal. State the reason for your answer.

(a) The continuous-time signal  $x(t)$ , defined by

$$x(t) = \begin{cases} 3e^{-2t}, & t \geq 0, \\ 0, & \text{otherwise.} \end{cases}$$

$x(t)$  is an energy signal. (1 pt)

$$\int_{-\infty}^{+\infty} x^2(t) dt = \int_0^{\infty} 9e^{-4t} dt = \frac{9}{4} \quad (2 \text{ pts})$$

Or

Most of the energy of the signal is within finite period of time (2 pts)

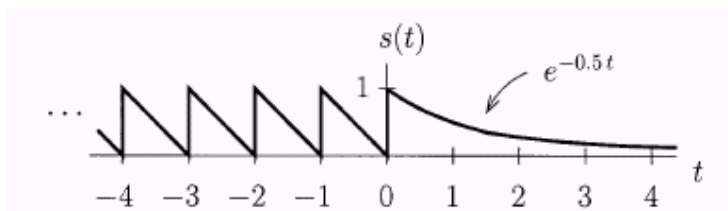
(b) The continuous-time signal  $z(t)$ , defined for  $-\infty < t < \infty$  by

$$z(t) = 3 \sin(\pi t) + 2 \cos(3\pi t)$$

$z(t)$  is a power signal (1 pt)

$z(t)$  is periodic with a period of 2 seconds (2 pts)

(c)



$s(t)$  is not an energy signal nor a power signal (1 pt)

Area under the square of the curve is not finite and the signal is not periodic (2 pts)

**Problem 2: (8 pts)**

A system is defined by the input-output relationship given by:

$$y(t) = x(t) \sin(2t) + 1$$

a. Is this system linear? Justify your answer

Not a linear system (1 pt)

$$x_1(t) \rightarrow y_1(t) = x_1(t) \sin(2t) + 1$$

$$x_2(t) \rightarrow y_2(t) = x_2(t) \sin(2t) + 1$$

(1 pt)

$$x_1(t) + x_2(t) \rightarrow [x_1(t) + x_2(t)] \sin(2t) + 1 \neq y_1(t) + y_2(t)$$

b. Is this system time-invariant? Justify your answer

Not a time invariant system. (1 pt)

$$x(t) \rightarrow y(t) = x(t) \sin(2t) + 1$$

$$x(t - t_0) \rightarrow x(t - t_0) \sin(2t) + 1 \neq y(t - t_0) = x(t - t_0) \sin(2t - 2t_0) + 1$$

(1 pt)

c. Is this system stable? Justify your answer

Yes, it is a stable system (1 pt)

$$\text{If } x(t) \text{ is bounded } |x(t)| \leq M, \text{ then } |y(t)| \leq |x(t)| + 1 \leq M + 1 < M$$

(bounded) (1 pt)

d. Is this system causal? Justify your answer.

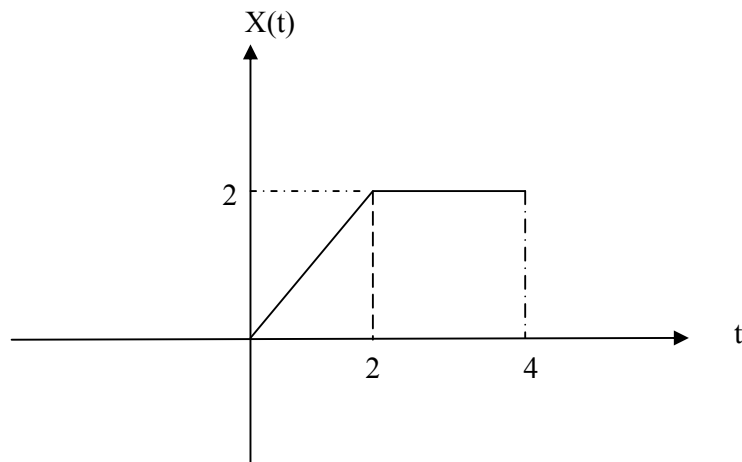
Yes it is a causal system (1 pt)

Output does not depend on future values of either the input or the output

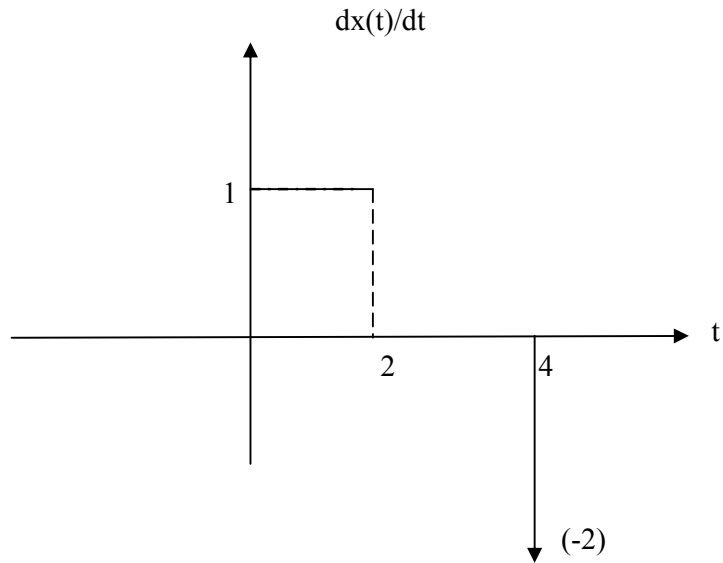
(1 pts)

**Problem 3 (6 pts)**

Consider the signal shown below.



- a. Draw the derivate of  $x(t)$ .



**-2 if the impulse at  $t=4s$  is missing**

- b. Determine the Laplace transform of  $x(t)$ .

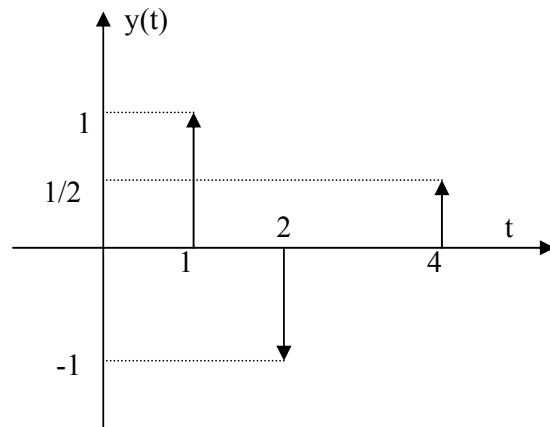
$$\text{LT}\left(\frac{dx(t)}{dt}\right) = \frac{1}{s} - \frac{e^{-2s}}{s} - 2e^{-4s} = sX(s) \quad (2 \text{ pts})$$

$$X(s) = \frac{1}{s^2} - \frac{e^{-2s}}{s^2} - \frac{2e^{-4s}}{s} \quad (1 \text{ pt})$$

**Problem 4 (6 pts)**

Consider the signal  $y(t)$  shown below

- a. Write  $y(t)$  in time domain



$$y(t) = \delta(t-1) - \delta(t-2) + \frac{1}{2}\delta(t-4) \quad (3 \text{ pts})$$

b. Determine the Laplace transform of  $y(t)$

$$Y(s) = e^{-s} - e^{-2s} + \frac{1}{2}e^{-4s} \quad (3 \text{ pts})$$

**Problem 5 (8 pts)**

For problems 3 and 4, determine the convolution signal  $z(t)=x(t)*y(t)$

$$Z(s) = X(s)Y(s) = \left[ \frac{1}{s^2} - \frac{e^{-2s}}{s^2} - \frac{2e^{-4s}}{s} \right] \left[ e^{-s} - e^{-2s} + 0.5e^{-4s} \right]$$

$$Z(s) = \left[ \frac{e^{-s}}{s^2} - \frac{e^{-2s}}{s^2} + 0.5 \frac{e^{-4s}}{s^2} - \frac{e^{-3s}}{s^2} + \frac{e^{-4s}}{s^2} - 0.5 \frac{e^{-6s}}{s^2} - 2 \frac{e^{-5s}}{s} + 2 \frac{e^{-6s}}{s} - \frac{e^{-8s}}{s} \right]$$

$$Z(s) = \left[ \frac{e^{-s}}{s^2} - \frac{e^{-2s}}{s^2} - \frac{e^{-3s}}{s^2} + 1.5 \frac{e^{-4s}}{s^2} - 2 \frac{e^{-5s}}{s} - 0.5 \frac{e^{-6s}}{s^2} + 2 \frac{e^{-6s}}{s} - \frac{e^{-8s}}{s} \right]$$

(4 pts)

$$z(t) = (t-1)u(t-1) - (t-2)u(t-2) - (t-3)u(t-3) + 1.5(t-4)u(t-4) - 2u(t-5) - 0.5(t-6)u(t-6) + 2u(t-6) - u(t-8)$$

(4 pts)

**Problem 6 (8 pts)**

Given

$$f(t) = \int_0^t e^{-3\tau} (t-\tau) e^{-2(t-\tau)} d\tau, \quad t \geq 0$$

a. Find the Laplace transform of  $f(t)$

$$f(t) = e^{-3t}u(t) * te^{-2t}u(t) \quad (2 \text{ pts})$$

$$F(s) = \frac{1}{(s+3)(s+2)^2} \quad (2 \text{ pts})$$

b. Using  $F(s)$  and the final value theorem. Can we use the Final value theorem? Justify your answer.

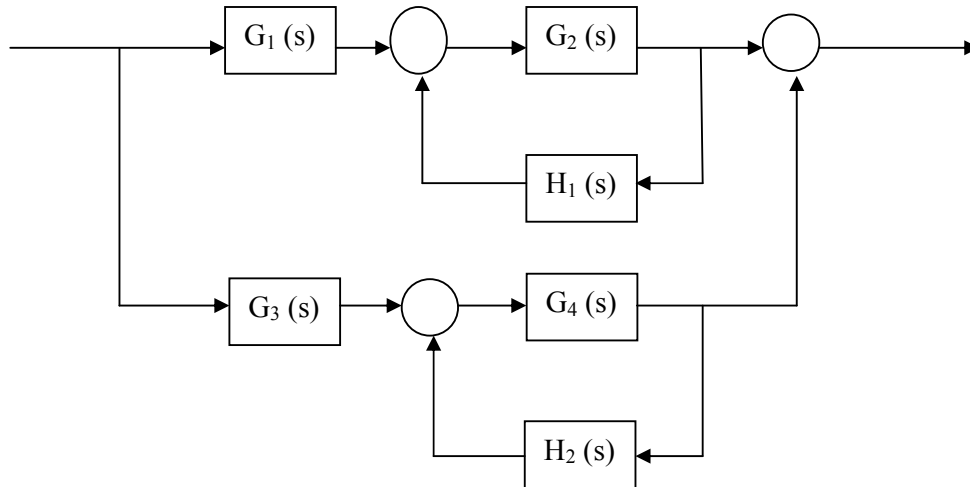
$sF(s)$  has poles at  $s=-3$  and  $s=-2$ . Final value theorem can be applied. (3 pts)

$$\lim_{t \rightarrow \infty} f(t) = \lim_{s \rightarrow 0} sF(s) = 0 \quad (1 \text{ pt})$$

$$t \rightarrow \infty \quad s \rightarrow 0$$

**Problem 7 (8 pts)**

A linear time-invariant control system is represented by the block diagram shown below:



Determine the transfer function of the system.

$$\frac{C(s)}{R(s)} = \frac{G_1(s)G_2(s)[1 + G_4(s)H_2(s)] + G_3(s)G_4(s)[1 + G_2(s)H_1(s)]}{1 + G_4(s)H_2(s) + G_2(s)H_1(s) + G_4(s)H_2(s)G_2(s)H_1(s)}$$

**(-2 for every missing term)**

**Problem 8 (12 pts)**

The open-loop transfer function of a unity feedback control system is given by:

$$G(s) = \frac{3s + 10}{s^3 + 2s^2 + s}$$

- Determine the Differential equation representing this system.

System transfer function

$$\frac{C(s)}{R(s)} = \frac{3s + 10}{s^3 + 2s^2 + 4s + 10} \quad \text{(4 pts)}$$

Differential equation

$$\frac{d^3c(t)}{dt^3} + 2\frac{d^2c(t)}{dt^2} + 4\frac{dc(t)}{dt} + 10c(t) = 3\frac{dr(t)}{dt} + 10r(t) \quad (2 \text{ pts})$$

2. Represent this system in a state variable form by determining the state equations and the output equation in matrix format.

Using the same procedures explained in class

$$\begin{bmatrix} X_1'(t) \\ X_2'(t) \\ X_3'(t) \end{bmatrix} = \begin{bmatrix} -2 & 1 & 0 \\ -4 & 0 & 1 \\ -10 & 0 & 0 \end{bmatrix} \begin{bmatrix} X_1(t) \\ X_2(t) \\ X_3(t) \end{bmatrix} + \begin{bmatrix} 0 \\ 3 \\ 10 \end{bmatrix} r(t) \quad (4 \text{ pts})$$

$$y(t) = \begin{bmatrix} 1 & 0 & 0 \end{bmatrix} \begin{bmatrix} X_1(t) \\ X_2(t) \\ X_3(t) \end{bmatrix} + \begin{bmatrix} 0 \end{bmatrix} r(t) \quad (2 \text{ pts})$$

**Problem 9 (7 pts)**

The transfer function of a linear time-invariant system, whose input is  $x(t)$  and output  $y(t)$ , is given by

$$\frac{Y(s)}{X(s)} = \frac{3s + 10}{2s^4 + s^3 + 3s^2 + 5s + 10}$$

Check the stability of the system by determining the number of poles of the system in each half of the complex-plane.

$$\begin{array}{l|lll} S^4 & 2 & 3 & 10 \\ S^3 & 1 & 5 & \\ S^2 & -7 & 10 & \\ S^1 & 45/7 & & \\ S^0 & 10 & & \end{array}$$

2 changes of sign implies 2 roots in the rhp. No complex roots. Root distributions: (2,0,2). The system is unstable.

- 2 for every wrong entry in the table
- 2 for root locations
- 2 for stability conclusion.

**Problem 10 (8 pts)**

The transfer function of a linear time-invariant system, whose input is  $x(t)$  and output  $y(t)$ , is given by

$$\frac{Y(s)}{X(s)} = \frac{3s + 10}{s^4 + s^3 - s - 1}$$

Check the stability of the system by determining the number of poles of the system in each half of the complex-plane.

$S^4$	1	0	-1	
$S^3$	1	-1		
$S^2$	1	-1		$A(S) = S^2 - 1$
$S^1$	0			$A'(S) = 2S$
$S^0$	2			
$S^0$	-1			

1 change of sign implies 1 root in the rhp. No complex roots because the roots of  $A(S)$  are -1 and +1. Root distributions: (3,0,1). The system is unstable.

- 2 for every wrong entry in the table**
- 4 for wrong table or not using RH table**
- 2 for  $A(S)$  and its solution**
- 2 for root locations**
- 2 for stability conclusion.**

**Problem 11 (5 pts)**

What is the Fourier Transform of

$$y(t) = e^{-a|t|} \sin(\omega_0 t)$$

Let  $f(t) = e^{-a|t|}$ .  $F(\omega) = \frac{2a}{a^2 + \omega^2}$  **(2 pts)**

Shifting property of the FT

$$f(t) \leftrightarrow F(\omega)$$

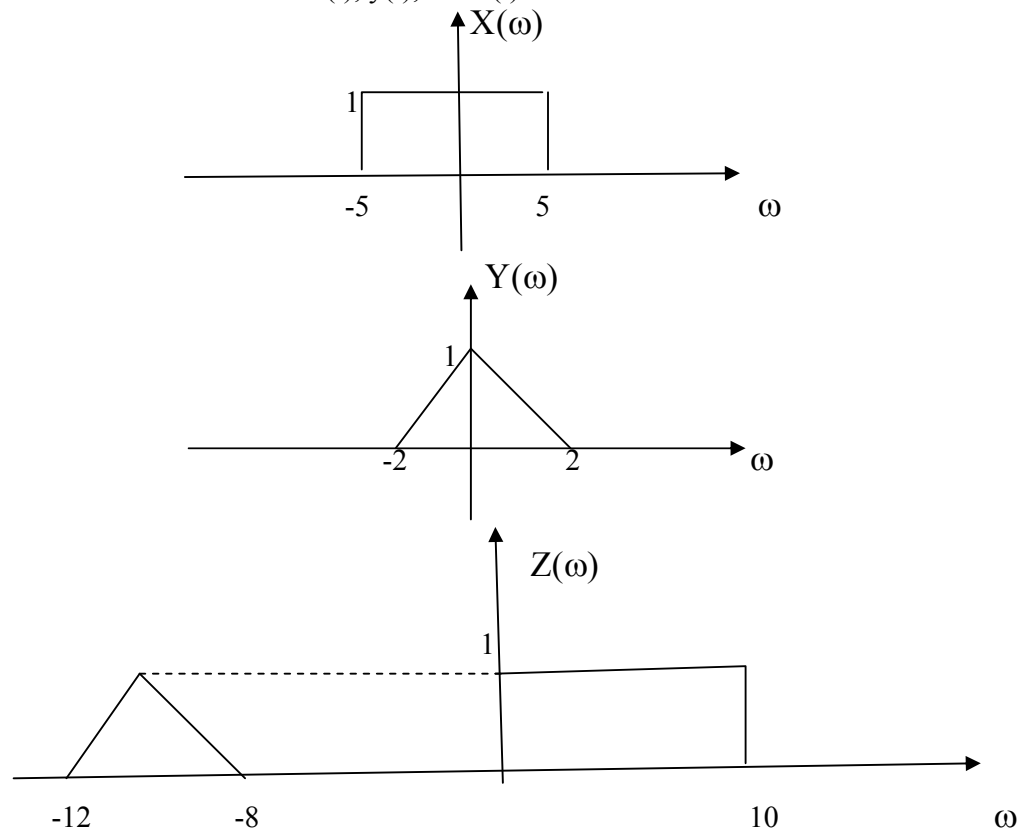
$$f(t) \sin(\omega_0 t) \leftrightarrow \frac{1}{2j} [F(\omega - \omega_0) - F(\omega + \omega_0)]$$

Accordingly,

$$Y(\omega) = \frac{1}{2j} \left[ \frac{2a}{a^2 + (\omega - \omega_0)^2} - \frac{2a}{a^2 + (\omega + \omega_0)^2} \right] \quad \text{(3 pts)}$$

**Problem 12 (10 pts)**

The Fourier transforms of  $x(t)$ ,  $y(t)$ , and  $z(t)$  are shown below.

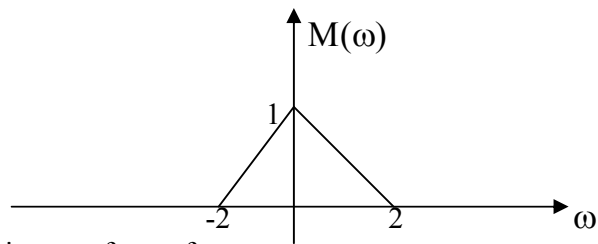


- What is the bandwidth of  $x(t)$ ?  
5 rad/s **(2 pts)**
- What is the bandwidth of  $y(t)$ ?  
2 rad/s **(2 pts)**
- Is  $z(t)$  a real-valued signal? Explain.  
Not real as there is no symmetry about the origin (even symmetry) **(2 pts)**
- Write  $z(t)$  as a function of  $x(t)$  and  $y(t)$ .  
 $z(t) = x(t)e^{i5t} + y(t)e^{-j10t}$  **(4 pts)**



**Problem 13 (5 pts)**

The Fourier transforms of  $m(t)$  is shown below.



Sketch the Fourier transform of

$$s(t) = 10[1 + 0.8m(t)]\cos(20t)$$

$$s(t) = 10\cos 20t + 8m(t)\cos(20t)$$

$$S(\omega) = 10\pi[\delta(\omega - 20) + \delta(\omega + 20)] + 4[M(\omega - 20) + M(\omega + 20)]$$

