

**AMERICAN UNIVERSITY OF BEIRUT**  
**FACULTY OF ENGINEERING AND ARCHITECTURE**  
**ELECTRICAL AND COMPUTER ENGINEERING DEPARTMENT**  
**EECE 340 – Signals & Systems**  
**QUIZ # 1**  
**Open book exam**  
**TIME: 1 HOUR and 30 MINUTES**  
**March 28, 2009**  
**INSTRUCTOR: Dr. JEAN J. SAADE**

NAME : \_\_\_\_\_

ID # : \_\_\_\_\_

**INSTRUCTIONS**

- Write your Name and ID # on this sheet, the computer card and the scratch booklet in the provided spaces.
- Provide your answer on the computer card and solution of each problem on the scratch booklet.
- Return the computer card, this question sheet and the scratch booklet when you finish the test.
- All questions are equally weighted in grading.

Classify the signals given in Problems #1 and # 2 into energy-type, power-type and signals that are neither energy nor power-type. For each energy or power-type signal, find the energy or average power of the signal.

**PROBLEM # 1**

The following continuous-time signal is considered:

$$f(t) = 4e^{j\omega_0 t} + 3e^{j2\omega_0 t} \text{ Volts}$$

- (a) Neither energy nor power signal.
- (b) Energy signal with energy equal 25 Joules.
- (c) Power signal with average power equal to 25 Watts.
- (d) Periodic signal, but neither energy nor power signal.
- (e) None of the above.

**PROBLEM # 2**

The following continuous-time signal is considered:

$$f(t) = e^{-2|t|} \text{ Volts}$$

- (a) Neither energy nor power signal.
- (b) Energy signal with energy equal 0.5 Joules.
- (c) Power signal with average power equal to 0.5 Watts.
- (d) Periodic signal but neither energy nor power signal.
- (e) None of the above.

**PROBLEM # 3**

In this problem,  $x(t)$  denotes the input and  $y(t)=T[x(t)]$  denotes the output of a continuous time system. Let the system be given by

$$y(t)=T[x(t)]=x(t) \cos [x(t)] \times \sin [x(t)]$$

Select from what is given below only one property that is **not** satisfied by the system.

- (a) Non-linearity
- (b) Time-invariance
- (c) Time-varying
- (d) Causality

**PROBLEM # 4**

In this problem,  $x(t)$  denotes the input and  $y(t)=T[x(t)]$  denotes the output of a continuous time system. Let the system be given by

$$y(t)=T[x(t)]=x(t-1)+2$$

Select from what is given below only one property that is **not** satisfied by the system.

- (a) Linearity
- (b) Non-linearity
- (c) Time-invariance
- (d) Causality

**PROBLEM # 5**

In this problem,  $x(t)$  denotes the input and  $y(t)=T[x(t)]$  denotes the output of a continuous time system. Let the system be given by

$$y(t)=T[x(t)]=x(2t-2)$$

Determine the time instants at which the system output depends on future values of the system input.

- (a)  $-3 < t < 3$  secs.
- (b)  $t < 3$  secs.
- (c)  $-2 < t < 2$  secs.
- (d)  $-\infty < t < \infty$
- (e)  $t > 2$  secs.

**PROBLEM # 6**

Consider the following system with  $f(t)$  being the system input and  $g(t)$  the system output:

$$g(t) = T[f(t)] = \int_{-\infty}^t f(\tau) d\tau + \int_{-\infty}^{\infty} e^{(t-\tau)} f(\tau) d\tau, \quad -\infty < t < \infty$$

Determine the impulse response,  $h(t)$ , of the system.

- (a)  $h(t) = \begin{cases} e^{-t}, & t < 0 \\ e^{-t} + 1, & t \geq 0 \end{cases}$       (b)  $h(t) = \begin{cases} e^{-t}, & t < 0 \\ e^t + 1, & t \geq 0 \end{cases}$       (c)  $h(t) = \begin{cases} e^t, & t < 0 \\ e^{-t} + 1, & t \geq 0 \end{cases}$
- (d)  $h(t) = \begin{cases} e^t, & t < 0 \\ e^t + 1, & t \geq 0 \end{cases}$       (e)  $h(t) = \begin{cases} e^t + 1, & t < 0 \\ e^t, & t \geq 0 \end{cases}$

**PROBLEM # 7**

Consider again the system in Problem # 6. Determine the system output when the input is the unit step signal,  $u(t)$ .

- (a)  $h(t) = \begin{cases} e^t, & t < 0 \\ e^t + t, & t \geq 0 \end{cases}$       (b)  $h(t) = \begin{cases} e^{-t}, & t < 0 \\ e^t + t, & t \geq 0 \end{cases}$       (c)  $h(t) = \begin{cases} e^t, & t < 0 \\ e^{-t} + t, & t \geq 0 \end{cases}$
- (d)  $h(t) = \begin{cases} e^{-t}, & t < 0 \\ e^{-t} + t, & t \geq 0 \end{cases}$       (e)  $h(t) = \begin{cases} e^t + t, & t < 0 \\ e^t, & t \geq 0 \end{cases}$

**PROBLEM # 8**

Consider an LTI continuous-time system whose response to the input  $f(t) = [e^{-t} + e^{-3t}]u(t)$  is  $g(t) = [2e^{-t} - 2e^{-4t}]u(t)$ . Determine the impulse response,  $h(t)$ , of the system assumed causal.

- (a)  $h(t) = 1.5(e^{-2t} + e^{-t})u(t)$       (b)  $h(t) = 1.5(e^{-t} + e^{-4t})u(t)$
- (c)  $h(t) = 1.5(e^{-2t} + e^{-4t})u(t)$       (d)  $h(t) = 1.5(e^{-3t} + e^{-5t})u(t)$
- (e)  $h(t) = 1.5(e^{-6t} + e^{-5t})u(t)$

**PROBLEM # 9**

Consider again Problem # 8 and determine the differential equation relating the output,  $g(t)$ , to the input,  $f(t)$ , of the system.

$$(a) \frac{d^2 g(t)}{dt^2} + \frac{dg(t)}{dt} + 8g(t) = 3 \frac{df(t)}{dt} + 9f(t)$$

$$(b) \frac{d^2 g(t)}{dt^2} + 6 \frac{dg(t)}{dt} + 4g(t) = 3 \frac{df(t)}{dt} + 9f(t)$$

$$(c) \frac{d^2 g(t)}{dt^2} + 6 \frac{dg(t)}{dt} + 8g(t) = \frac{df(t)}{dt} + 9f(t)$$

$$(d) \frac{d^2 g(t)}{dt^2} + 6 \frac{dg(t)}{dt} + 8g(t) = 3 \frac{df(t)}{dt} + 9f(t)$$

$$(e) \frac{d^2 g(t)}{dt^2} + 6 \frac{dg(t)}{dt} + g(t) = 2 \frac{df(t)}{dt} + f(t)$$

**PROBLEM # 10**

Consider an LTI system with transfer function given by

$$H(s) = \frac{2e^{-2s}}{(s+4)(s+2)}$$

Determine the impulse response,  $h(t)$ , of the system considered to be causal.

$$(a) h(t) = (e^{-2(t-2)} - e^{-4(t-2)})u(t-2) \quad (b) h(t) = (e^{-2(t-2)} - e^{-4(t-2)})u(t)$$

$$(c) h(t) = (e^{-4t} - e^{-2t})u(t-2) \quad (d) h(t) = (e^{-2t} - e^{-4t})u(t)$$

$$(e) h(t) = (e^{-4(t+2)} - e^{-2(t+2)})u(t+2)$$

**PROBLEM # 11**

Consider the asynchronous detection of the following received DSB-SC signal:

$$s_r(t) = A_r f(t) \cos(\omega_c t - \pi/2)$$

The received signal is multiplied by  $c(t) = A_c \cos[(\omega_c + \pi/2)t]$  and then the output of the multiplier is inputted to an ideal low-pass filter. Determine the positive time instants at which the signal at the output of the LPF is null.

$$(a) 3, 5, 7, 9, \dots \text{ secs}$$

$$(b) 1, 2, 3, 4, \dots \text{ secs}$$

$$(c) 1, 3, 5, 7, \dots \text{ secs}$$

$$(d) 0, 2, 4, 6, \dots \text{ secs}$$

$$(e) 4, 6, 8, 10, \dots \text{ secs}$$

**PROBLEM # 12**

Consider an LTI system with impulse response given by

$$h(t) = 8e^{-2t} \sin(\omega_c t) u(t)$$

Determine the transfer function,  $H(s)$ , of this system and then conclude if this system is stable or not.

$$(a) H(s) = \frac{8\pi\omega_c}{s^2 + 2s + 2 + \omega_c^2}, \text{ stable} \quad (b) H(s) = \frac{8\omega_c}{s^2 + 2s + 4 + \omega_c^2}, \text{ unstable}$$

$$(c) H(s) = \frac{8\omega_c}{s^2 + 4s + 2 + \omega_c^2}, \text{ unstable} \quad (d) H(s) = \frac{8\omega_c}{s^2 + 4s + 4 + \omega_c^2}, \text{ stable}$$

$$(e) H(s) = \frac{16\pi\omega_c}{4s^2 + 4s + 4 + \omega_c^2}, \text{ unstable}$$

Problem #1

$$f(t) = 4e^{j\omega_0 t} + 3e^{j2\omega_0 t} \text{ Volts.}$$

The signal  $f(t)$  is periodic with period equal to  $\frac{2\pi}{\omega_0}$ . It is also a power signal with finite average power given by:

$$\begin{aligned} P &= \frac{\omega_0}{2\pi} \int_{-\frac{\pi}{\omega_0}}^{\frac{\pi}{\omega_0}} |f(t)|^2 dt = \frac{\omega_0}{2\pi} \int_{-\frac{\pi}{\omega_0}}^{\frac{\pi}{\omega_0}} [25 + 24\cos(\omega_0 t)] dt \\ &= \frac{\omega_0}{2\pi} \int_{-\frac{\pi}{\omega_0}}^{\frac{\pi}{\omega_0}} 25 dt = 25 \text{ Watts.} \end{aligned}$$

In the above we have used

$$|f(t)|^2 = [\operatorname{Re} f(t)]^2 + [\operatorname{Im} f(t)]^2 = 25 + 24\cos(\omega_0 t)$$

$$\text{and } \int_{-\frac{\pi}{\omega_0}}^{\frac{\pi}{\omega_0}} \cos(\omega_0 t) dt = 0.$$

Problem #2  $f(t) = e^{-2|t|}$  Volts.

$f(t)$  is an energy signal with finite energy given by:

$$\begin{aligned} E &= \int_{-\infty}^{\infty} f^2(t) dt = \int_{-\infty}^{\infty} e^{-4|t|} dt = 2 \int_0^{\infty} e^{-4t} dt \\ &= \frac{1}{2} \text{ Joule.} \end{aligned}$$

### Problem #3

2/10

$y(t) = T[x(t)] = x(t) \cos[x(t)] \sin[x(t)]$   
is the transformation defining a continuous-time system with input  $x(t)$  and output  $y(t)$ .

\* Linearity:

$$T[a_1 x_1(t) + a_2 x_2(t)] = [a_1 x_1(t) + a_2 x_2(t)] \times \\ \cos[a_1 x_1(t) + a_2 x_2(t)] \times \\ \sin[a_1 x_1(t) + a_2 x_2(t)]$$

$$= ?? a_1 T[x_1(t)] + a_2 T[x_2(t)]$$

$$= a_1 x_1(t) \cos[x_1(t)] \sin[x_1(t)]$$

$$+ a_2 x_2(t) \cos[x_2(t)] \sin[x_2(t)] \rightarrow \text{NO.}$$

$\Rightarrow$  The system is non-linear.

\* Time invariance

$$\text{Is } y(t-t_0) = T[x(t-t_0)] ??$$

$$\text{Yes, since } T[x(t-t_0)] = x(t-t_0) \cos[x(t-t_0)] \\ \times \sin[x(t-t_0)] \\ = y(t-t_0).$$

$\Rightarrow$  The system is time-invariant.

\* Causality  $y(t_0) = x(t_0) \cos[x(t_0)] \sin[x(t_0)]$

So, the output at time  $t_0$  depends on value of the input at time  $t_0$ . The system is causal.

## Problem # 4

(3/10)

$y(t) = T[x(t)] = x(t-1) + 2$   
is the transformation defining a continuous-time system.

### \* Linearity

$$T[a_1 x_1(t) + a_2 x_2(t)] = a_1 x_1(t-1) + a_2 x_2(t-1) + 2.$$

$$= ?? a_1 T[x_1(t)] + a_2 T[x_2(t)]$$

$$= a_1 \{x_1(t-1) + 2\} + a_2 \{x_2(t-1) + 2\}$$

$$= a_1 x_1(t-1) + 2a_1 + a_2 x_2(t-1) + 2a_2$$

$$= a_1 x_1(t-1) + a_2 x_2(t-1) + 2(a_1 + a_2) \rightarrow \text{No.}$$

$\Rightarrow$  The system is non-linear.

### \* Time invariance

$$\text{Is } y(t-t_0) = T[x(t-t_0)] ??$$

$$T[x(t-t_0)] = x(t-t_0-1) + 2$$

$$= y(t-t_0) \rightarrow \text{Yes.}$$

$\Rightarrow$  The system is time-invariant.

### \* Causality $y(t_0) = x(t_0-1) + 2$

The output at time  $t_0$  depends on the input value at time  $(t_0-1)$ .  $\Rightarrow$  The system is causal.



## Problem # 5

(4/10)

$$y(t) = T[x(t)] = x(2t-2)$$

The time instants,  $t$ , at which the system output  $y(t)$  depends on future values of the system input are such that  $t < 2t-2$ . Hence,  $t > 2$  secs.

The system, therefore, cannot be realized for any  $t > 2$  secs. Take for example  $t = 3$  secs.

The input  $x(3)$  produces the output  $y(3) = x(4)$ , which is not available at  $t = 3$  secs.

The system, however, can be realized if the input signals,  $x(t)$ , are only restricted to  $t \leq 2$  secs.

For example,  $y(2) = T[x(2)] = x(2)$

$$y(1) = T[x(1)] = x(0)$$

$$y(0) = T[x(0)] = x(-2)$$

$$y(-1) = T[x(-1)] = x(-4)$$

The output is obtained from the past values of the system input. (present)

As a whole; i.e., for input signals defined for all  $t \in (-\infty, +\infty)$ , the system is non-realizable. It is, therefore, non-causal.

Problem # 6

(5/10)

$$g(t) = T[f(t)] = \int_{-\infty}^t f(\tau) d\tau + \int_{-\infty}^{\infty} e^{(t-\tau)} f(\tau) d\tau,$$

$$\begin{aligned} h(t) &= T[\delta(t)] = \int_{-\infty}^t \delta(\tau) d\tau + \int_{-\infty}^{\infty} e^{(t-\tau)} \delta(\tau) d\tau \\ &= \int_{-\infty}^t \delta(\tau) d\tau + e^t \int_{-\infty}^{\infty} \delta(\tau) d\tau \\ &= \begin{cases} e^t, & t < 0 \\ e^t + 1, & t \geq 0 \end{cases} \end{aligned}$$

Problem # 7

$$\begin{aligned} g(t) &= T[ut(t)] = \int_{-\infty}^t u(\tau) d\tau + e^t \int_{-\infty}^{\infty} e^{-\tau} u(\tau) d\tau \\ &= \int_0^t d\tau + e^t \int_0^{\infty} e^{-\tau} d\tau \\ &= t + e^t, \quad t \geq 0 \end{aligned}$$

$$\text{For } t < 0, \quad g(t) = e^t \int_0^{\infty} e^{-\tau} d\tau = e^t$$

$$\Rightarrow g(t) = \begin{cases} e^t, & t < 0 \\ t + e^t, & t \geq 0 \end{cases}$$

(6/10)

Problem # 8

The system input is  $f(t) = [e^{-t} + e^{-3t}]u(t)$

The system output is:  $g(t) = [2e^{-t} - 2e^{-4t}]u(t)$

Since the system is LTI, then the system transfer function can be written as:

$$H(s) = \frac{G(s)}{F(s)}$$

$$F(s) = \mathcal{L}[f(t)] = \frac{1}{s+1} + \frac{1}{s+3} \\ = \frac{2s+4}{(s+1)(s+3)}, \sigma > -1$$

$$G(s) = \frac{2}{s+1} - \frac{2}{s+4} = \frac{6}{(s+1)(s+4)}, \sigma > -1$$

$$\Rightarrow H(s) = \frac{3(s+3)}{(s+2)(s+4)}, \sigma > -2 \text{ (causal system)}$$

$$h(t) = \frac{1}{2\pi j} \int_{Br} H(s) e^{st} ds = \text{Residue at } -2 + \text{Residue at } -4 \\ = \frac{3(s+3)e^{st}}{(s+4)} \Big|_{s=-2} + \frac{3(s+3)e^{st}}{(s+2)} \Big|_{s=-4}$$

$$= \frac{3}{2} e^{-2t} + \frac{3}{2} e^{-4t}, t > 0$$

$$= 0, t < 0$$

$$\Rightarrow h(t) = 1.5(e^{-2t} + e^{-4t})u(t)$$

Problem # 9

$$H(s) = \frac{3s+9}{(s+2)(s+4)} = \frac{3s+9}{s^2+6s+8}$$

$$= \frac{G(s)}{F(s)}$$

$$\Rightarrow s^2 G(s) + 6s G(s) + 8 G(s) = 3s F(s) + 9 F(s)$$

$$\Rightarrow \frac{d^2 g(t)}{dt^2} + 6 \frac{dg(t)}{dt} + 8 g(t) = 3 \frac{df(t)}{dt} + 9 f(t)$$

is the system differential equation.

Problem # 10

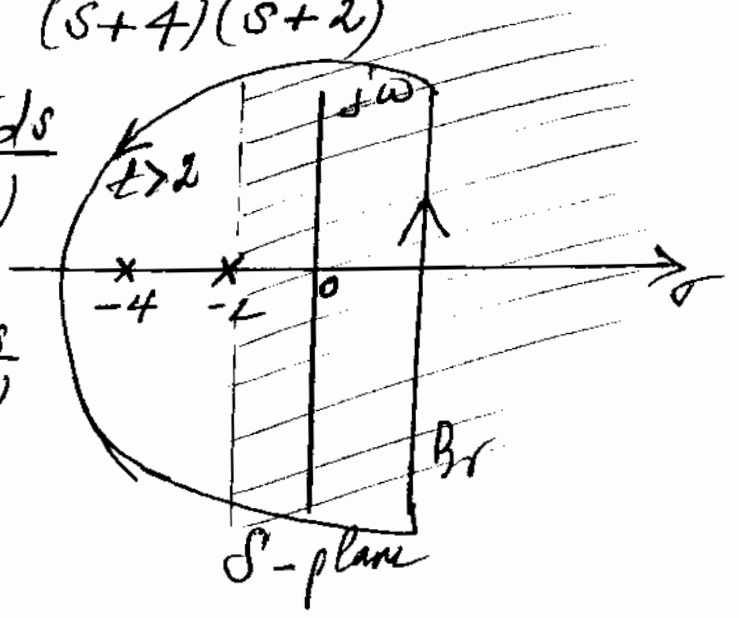
$$H(s) = \frac{2e^{-2s}}{(s+4)(s+2)}$$

$$h(t) = \frac{1}{2\pi j} \int_{Br} \frac{2e^{-2s} e^{st}}{(s+4)(s+2)} ds$$

$$= \frac{1}{2\pi j} \int_{Br} \frac{2e^{s(t-2)}}{(s+4)(s+2)} ds$$

$$= \frac{2e^{s(t-2)}}{(s+4)} \Big|_{s=-2}$$

$$+ \frac{2e^{s(t-2)}}{(s+2)} \Big|_{s=-4}$$



$$= \left( e^{-2(t-2)} - e^{-4(t-2)} \right) u(t-2)$$

Problem # 11

8/10

$$s_r(t) = A_r f(t) \cos\left(\omega_c t - \frac{\pi}{2}\right)$$

$$c(t) = A_c \cos\left[\left(\omega_c + \frac{\pi}{2}\right)t\right]$$

$$\begin{aligned} s_r(t)c(t) &= A_r A_c f(t) \cos\left(\omega_c t - \frac{\pi}{2}\right) \cos\left[\left(\omega_c + \frac{\pi}{2}\right)t\right] \\ &= \frac{1}{2} A_r A_c f(t) \left\{ \cos\left[\left(2\omega_c + \frac{\pi}{2}\right)t - \frac{\pi}{2}\right] \right. \\ &\quad \left. + \cos\left(\frac{\pi}{2}t + \frac{\pi}{2}\right) \right\} \end{aligned}$$

The low-pass filter output is:

$$LPF_0 = \frac{1}{2} A_r A_c f(t) \cos\left(\frac{\pi}{2}t + \frac{\pi}{2}\right)$$

$$LPF_0 = 0 \text{ when } \cos\left(\frac{\pi}{2}t + \frac{\pi}{2}\right) = 0 \text{ or}$$

$$\frac{\pi}{2}t + \frac{\pi}{2} = (2k+1)\frac{\pi}{2}, \quad k = 0, \pm 1, \pm 2, \dots$$

$$\Rightarrow t = 2k, \quad k = 0, \pm 1, \pm 2, \dots$$

Hence, the positive time instants at which the low-pass filter output is null are as above for  $k = 0, +1, +2, \dots$

$$\Rightarrow t = 0, 2, 4, 6, \dots \text{ secs.}$$

Problem #12

9/10

$$h(t) = 8 e^{-2t} \sin(\omega_c t) u(t)$$

is the impulse response of an LTI system.

Using  $\sin(\omega_c t) = \frac{e^{j\omega_c t} - e^{-j\omega_c t}}{2j}$ , then:

$$H(s) = \frac{8}{2j} \int_0^{\infty} e^{-2t} \frac{e^{j\omega_c t} - e^{-j\omega_c t}}{e^{-st}} dt = \frac{8}{2j} \int_0^{\infty} e^{-2t} e^{j\omega_c t} e^{-st} dt - \frac{8}{2j} \int_0^{\infty} e^{-2t} e^{-j\omega_c t} e^{-st} dt$$

$$= \frac{8}{2j} \left[ \int_0^{\infty} e^{-(s+2-j\omega_c)t} dt - \int_0^{\infty} e^{-(s+2+j\omega_c)t} dt \right]$$

$$\int_0^{\infty} e^{-(s+2-j\omega_c)t} dt = -\frac{1}{(s+2-j\omega_c)} e^{-(s+2-j\omega_c)t} \Big|_0^{\infty}$$

$$= \frac{1}{s+2-j\omega_c}, \quad \sigma > -2.$$

$$\int_0^{\infty} e^{-(s+2+j\omega_c)t} dt = \frac{1}{s+2+j\omega_c}, \quad \sigma > -2.$$

$$\Rightarrow H(s) = \frac{8}{2j} \left[ \frac{1}{s+2-j\omega_c} - \frac{1}{s+2+j\omega_c} \right]$$

$$= \frac{8}{2j} \left[ \frac{s+2+j\omega_c - s-2+j\omega_c}{(s+2-j\omega_c)(s+2+j\omega_c)} \right]$$

$$= \frac{8\omega_c}{s^2 + 4s + 4 + \omega_c^2} = \frac{8\omega_c}{(s+2)^2 + \omega_c^2}$$

$H(s)$  has poles at  $s = -2 \pm j\omega_c$ . (10/10)

These poles are in the left half  $s$ -plane and since the system is causal;  $h(t) = 0$  for  $t < 0$ , then the system is stable.

Note  $H(s)$  can also be obtained from the Laplace of  $\sin(\omega_c t) u(t) = f(t)$  by writing  $H(s) = \delta F(s+2)$ .

From the Laplace transform tables, we have:

$$F(s) = \frac{\omega_c}{s^2 + \omega_c^2}$$

$$\Rightarrow H(s) = \frac{\delta \omega_c}{(s+2)^2 + \omega_c^2}$$