

AMERICAN UNIVERSITY OF BEIRUT
FACULTY OF ENGINEERING AND ARCHITECTURE
ELECTRICAL AND COMPUTER ENGINEERING DEPARTMENT

EECE 440 – Signals and Systems

FINAL EXAM

Closed book exam

**NINE SHEETS OF FORMULAS WITH NO PROBLEM SOLUTIONS ARE
ALLOWED**

TIME: 2 hours

Saturday, June 11, 2005

INSTRUCTOR: Dr. JEAN J. SAADE

Student Name: _____

ID #: _____

PROBLEM # 1

An analog linear and time-invariant system has the frequency response given by

$$H(\omega) = \frac{1}{2} \left[\frac{1}{a + j(\omega - 2)} + \frac{1}{a + j(\omega + 2)} \right]$$

where a is real. The system is assumed to be causal and stable.

- (a) Determine the transfer function $H(s)$ of the system. Locate the poles of $H(s)$ in the s -plane.
- (b) Determine the system impulse response, $h(t)$.
- (c) The system impulse response $h(t)$ is sampled at $t=nT$ to give the unit sample response, $h(n)$, of a discrete time system. Let $T=1\text{sec.}$ and determine $h(n)$ as well as $H(z)$. Locate the poles of $H(z)$ in the z -plane and state whether or not the system is stable. Justify your answer.

PROBLEM # 2

Consider the difference equation given below.

$$y(n) = \frac{5}{2} y(n-1) - y(n-2) + x(n) - x(n-1).$$

- (a) Determine the system transfer function $H(z)$.
- (b) Determine $h(n)$ for the causal system.
- (c) Let the input to the causal system be $x(n) = (1/2)^n u(n)$. Determine the output $y(n)$ using the inverse Z-transform.
- (d) Implement $H(z)$ in direct form I and direct form II.

PROBLEM # 3

Consider the discrete time sequences

$$h(n) = e^{-bn} u(n),$$

$$x(n) = e^{-a|n|}, \text{ all } n$$

with $a > b > 0$. $h(n)$ represents the unit sample response of a discrete time, linear and shift-invariant system.

- (a) Determine the output $y(n)$ of the system if $x(n)$ is the input. Use discrete convolution.
- (b) Determine whether $h(n)$ is stable. Validate your answer using time-domain analysis.
- (c) Determine the frequency response $H(\omega)$ of $h(n)$. Also, determine and plot the magnitude frequency response $|H(\omega)|$.

PROBLEM # 4

Consider the sequence

$$x(n) = \begin{cases} \left(\frac{1}{2}\right)^n, & 0 \leq n \leq 5, \\ 0, & \text{elsewhere.} \end{cases}$$

- (a) Determine the Z-transform of $x(n)$.
- (b) Determine the expression of the sampled version, $\tilde{X}(k)$, of $X(z)$ for $z = e^{j\frac{2\pi}{8}k} = W_8^{-k}$ and $k=0,1,\dots,7$. Plot the periodic extension of $x(n)$ which admits $\tilde{X}(k)$ as the Fourier series coefficients.
- (c) Determine the DFT, $X(k)$, of $x(n)$ considered to be of length 8. Evaluate $X(k)$ for $k=0,1$ only.
- (d) Use the DFT in Part (c) and plot the finite length sequence $x_1(n)$ whose DFT is given by $X_1(k) = W_8^{-4k} X(k)$.

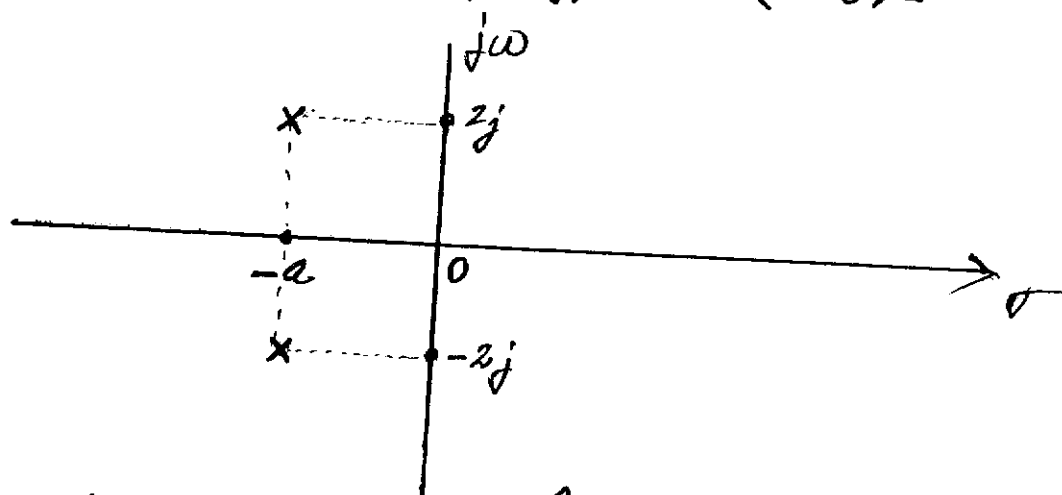
Problem #1

$$H(\omega) = \frac{1}{2} \left[\frac{1}{a + j(\omega - 2)} + \frac{1}{a + j(\omega + 2)} \right]$$

(a) Since the system is causal and stable, then the region of convergence of $H(s)$ is a right half s -plane that contains the imaginary axis.

Hence, $H(s)$ can be obtained from $H(\omega)$ or vice-versa by replacing $j\omega$ by s .

$$\Rightarrow H(s) = \frac{1}{2} \left[\frac{1}{s + (a - 2j)} + \frac{1}{s + (a + 2j)} \right]$$



The poles of $H(s)$ are as located above and they are in the left-half s -plane.

$$\begin{aligned} (b) \quad h(t) &= \mathcal{L}^{-1}[H(s)] = \frac{1}{2} \left[e^{(-a+2j)t} + e^{(-a-2j)t} \right] \\ &= e^{-at} \left[e^{2jt} + e^{-2jt} \right] \quad -at \end{aligned}$$

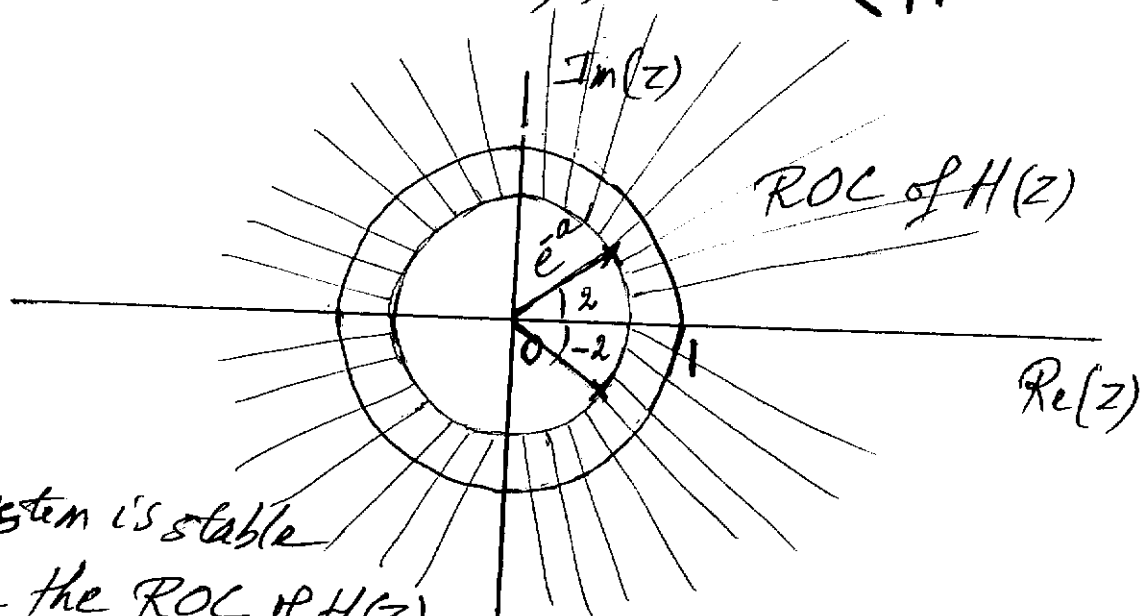
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$$(c) h(n) = e^{-an} \cos 2n u(n)$$

$$\begin{aligned} H(z) &= \sum_{n=0}^{\infty} h(n) z^{-n} = \sum_{n=0}^{\infty} e^{-an} \cos 2n z^{-n} \\ &= \sum_{n=0}^{\infty} \frac{e^{-an} e^{j2n}}{2} z^{-n} + \sum_{n=0}^{\infty} \frac{e^{-an} e^{-j2n}}{2} z^{-n} \\ &= \sum_{n=0}^{\infty} \frac{1}{2} \left(e^{-a+j2} z^{-1} \right)^n + \sum_{n=0}^{\infty} \frac{1}{2} \left(e^{-a-j2} z^{-1} \right)^n \\ &= \frac{1}{2} \cdot \frac{1}{1 - e^{-a+j2} z^{-1}} + \frac{1}{2} \cdot \frac{1}{1 - e^{-a-j2} z^{-1}} \end{aligned}$$

With $|e^{-a} z^{-1}| < 1$ or $|z| > e^{-a}$.

Since $a > 0$ (see (a)), then $e^{-a} < 1$.



The system is stable

Since the ROC of $H(z)$ contains the unit circle z -plane

Problem #2

$$y(n) = \frac{5}{2} y(n-1) - y(n-2) + x(n) - x(n-1).$$

$$(a) \quad Y(z) - \frac{5}{2} z^{-1} Y(z) + z^{-2} Y(z) = X(z) - z^{-1} X(z)$$

$$\Rightarrow Y(z) \left[1 - \frac{5}{2} z^{-1} + z^{-2} \right] = X(z) [1 - z^{-1}]$$

$$H(z) = \frac{Y(z)}{X(z)} = \frac{1 - z^{-1}}{1 - \frac{5}{2} z^{-1} + z^{-2}}$$

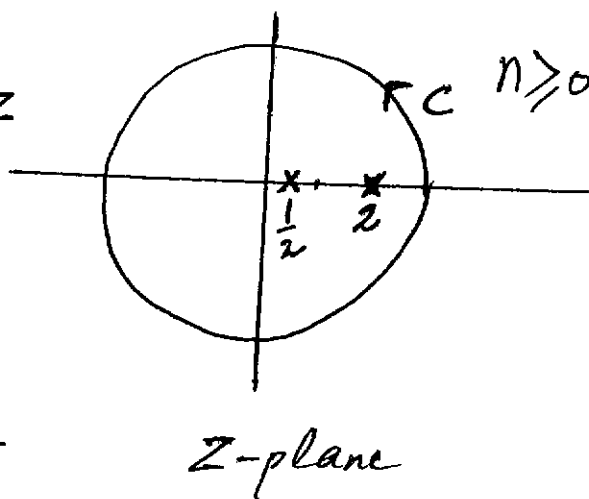
$$= \frac{z^2 - z}{z^2 - \frac{5}{2} z + 1} = \frac{z(z-1)}{(z-2)(z-\frac{1}{2})}$$

$$(b) \quad h(n) = \frac{1}{2\pi j} \oint_C H(z) z^{n-1} dz$$

$$= \left. \frac{z(z-1)}{(z-2)} z^{n-1} \right|_{z=\frac{1}{2}}$$

$$+ \left. \frac{z(z-1)}{(z-\frac{1}{2})} z^{n-1} \right|_{z=2}$$

$$= \frac{(\frac{1}{2}-1)}{(\frac{1}{2}-2)} \left(\frac{1}{2}\right)^n + \frac{(2-1)}{(2-\frac{1}{2})} 2^n$$



$$(c) \quad x(n) = \left(\frac{1}{2}\right)^n u(n)$$

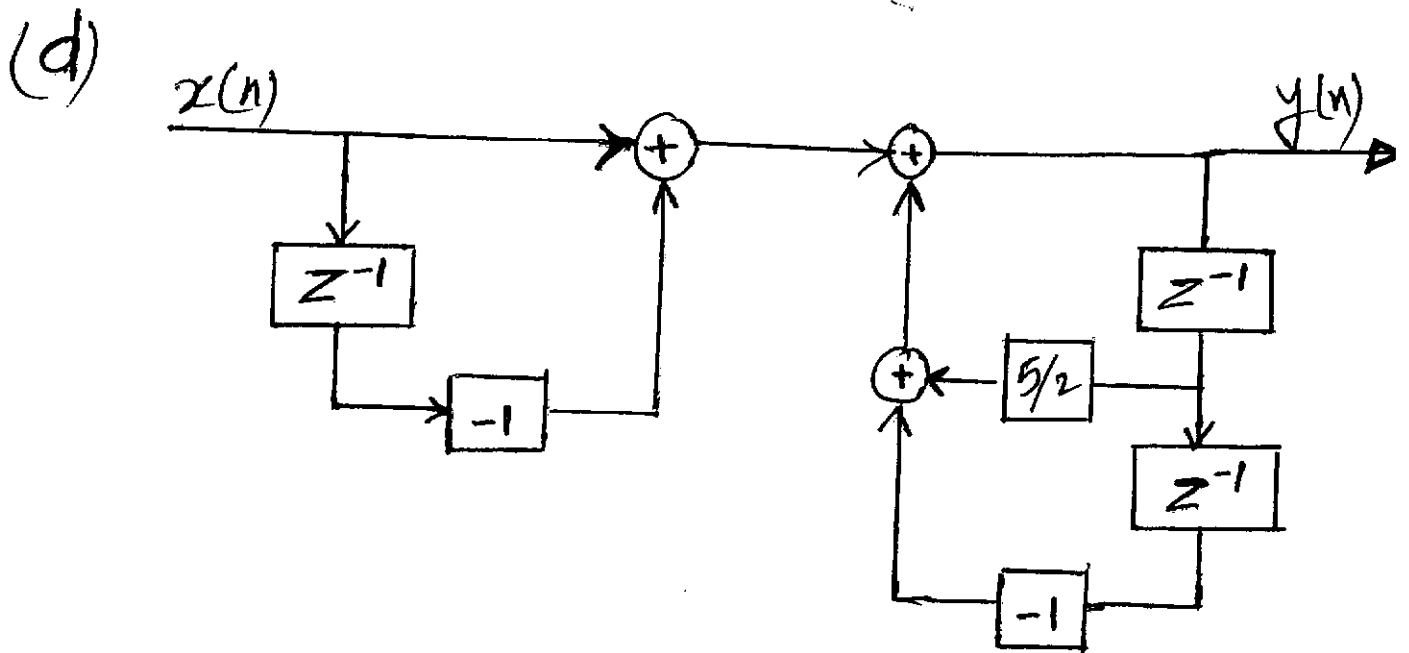
$$X(z) = \sum_{n=0}^{\infty} \left(\frac{1}{2} z^{-1}\right)^n = \frac{1}{1 - \frac{1}{2} z^{-1}}$$

$$= \frac{z}{z - \frac{1}{2}}; \quad |z| > \frac{1}{2}.$$

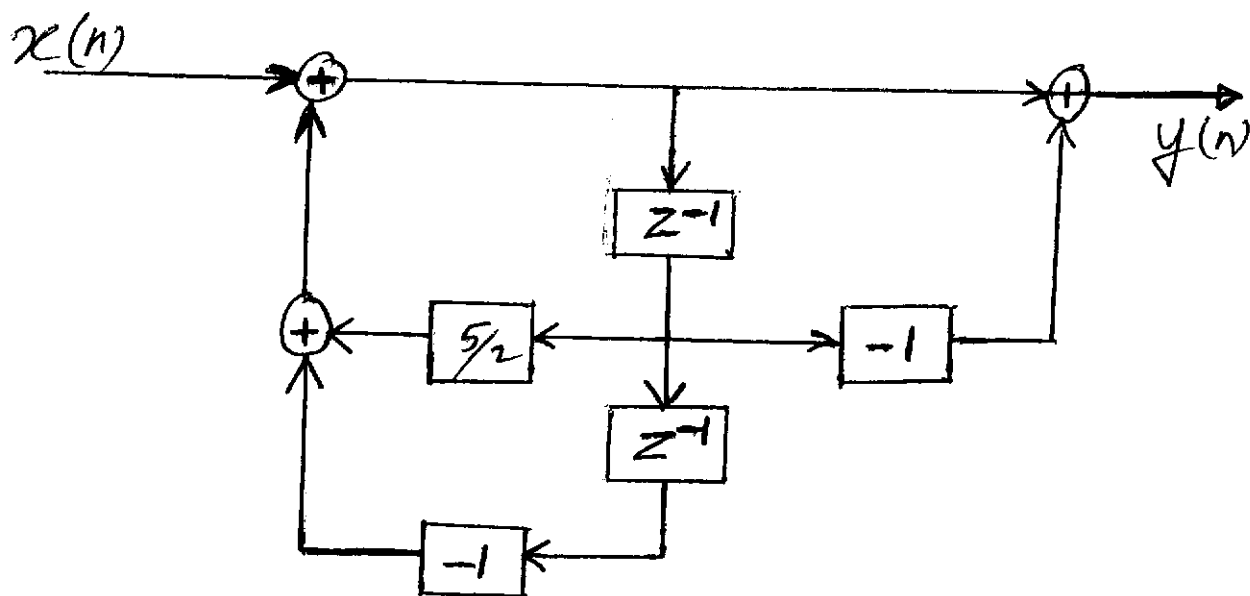
$$Y(z) = H(z)X(z) = \frac{z^2(z-1)}{(z-2)\left(z-\frac{1}{2}\right)^2}; \quad |z| > 2.$$

$$y(n) = \left. \frac{z^2(z-1)}{\left(z-\frac{1}{2}\right)^2} z^{n-1} \right|_{z=2} + \frac{1}{1!} \left[\frac{d}{dz} \frac{z^2(z-1)z^{n-1}}{\left(z-\frac{1}{2}\right)^2} \right]_{z=\frac{1}{2}}$$

$$= \frac{1}{9} \left[2^{n+3} + \left(\frac{1}{2}\right)^n (3n+1) \right] u(n)$$



Direct form I

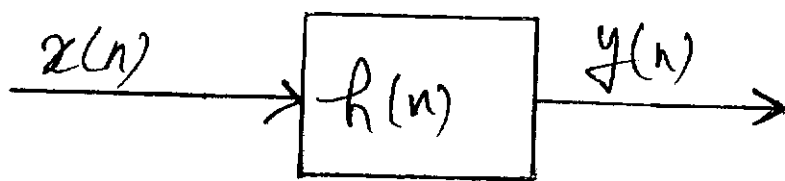


Direct form II

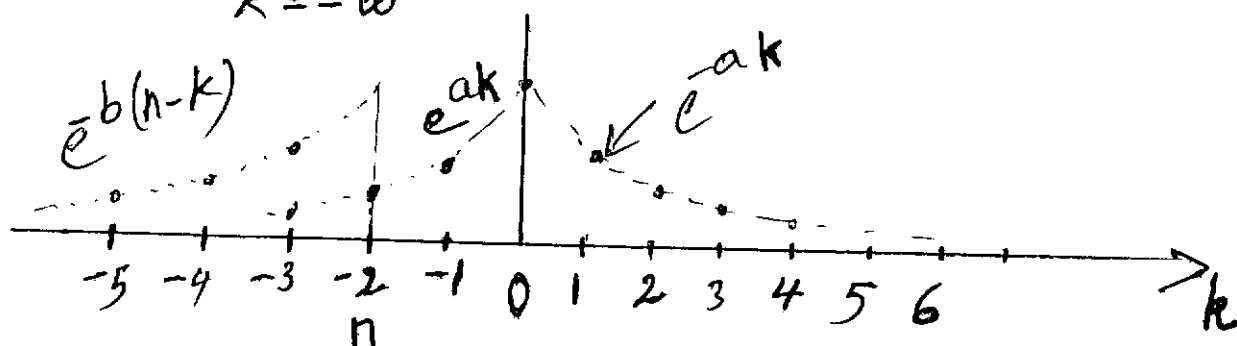
Problem #3

$$h(n) = e^{-bn} u(n); \quad x(n) = e^{-a|n|}, \text{ all } n.$$

$a > b > 0.$



$$(a) \quad y(n) = \sum_{k=-\infty}^{\infty} x(k) h(n-k)$$



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For $n \leq 0$,

$$\begin{aligned}
y(n) &= \sum_{k=-\infty}^n e^{-b(n-k)} e^{ak} = e^{-bn} \sum_{k=-\infty}^n e^{(a+b)k} \\
&= e^{-bn} \left[\sum_{k=-\infty}^0 e^{(a+b)k} - \sum_{k=n+1}^0 e^{(a+b)k} \right] \\
&= e^{-bn} \left[\sum_{k=0}^{\infty} e^{-(a+b)k} - \sum_{k=0}^{n-1} e^{-(a+b)k} \right] \\
&= e^{-bn} \left[\frac{1}{1 - e^{-(a+b)}} - \frac{1 - (e^{-(a+b)})^n}{1 - e^{-(a+b)}} \right] \\
&= e^{-bn} \times \frac{e^{-(a+b)n}}{1 - e^{-(a+b)}} \\
&= \frac{e^{an}}{1 - e^{-(a+b)}}
\end{aligned}$$

For $n \geq 0$,

$$\begin{aligned}
y(n) &= \sum_{k=-\infty}^0 e^{-b(n-k)} e^{ak} + \sum_{k=1}^n e^{-b(n-k)} e^{-ak} \\
&= e^{-bn} \sum_{k=-\infty}^0 e^{(a+b)k} + e^{-bn} \sum_{k=1}^n e^{-(a-b)k} \\
&= e^{-bn} \left[\sum_{k=0}^{\infty} e^{-(a+b)k} + \sum_{k=1}^n e^{-(a-b)k} \right]
\end{aligned}$$

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(7/10)

$$\begin{aligned}
 y(n) &= e^{-bn} \left[\frac{1}{1 - e^{-(a+b)}} + \frac{1 - e^{-(a-b)(n+1)}}{1 - e^{-(a-b)}} - 1 \right] \\
 &= e^{-bn} \left[\frac{1}{1 - e^{-(a+b)}} + \frac{e^{-(a-b)} - e^{-(a-b)(n+1)}}{1 - e^{-(a-b)}} \right] \\
 &= e^{-bn} \left[\frac{1}{1 - e^{-(a+b)}} + \frac{e^{-(a-b)} [1 - e^{-(a-b)n}]}{1 - e^{-(a-b)}} \right] \\
 &= e^{-bn} \left[\frac{1}{1 - e^{-(a+b)}} - \frac{1 - e^{-(a-b)n}}{1 - e^{-(a-b)}} \right] \\
 &= \frac{e^{-bn}}{1 - e^{-(a+b)}} + \frac{e^{-an} - e^{-bn}}{1 - e^{-(a-b)}}
 \end{aligned}$$

(b) $h(n) = e^{-bn} u(n)$

$$\sum_{n=0}^{\infty} |h(n)| = \sum_{n=0}^{\infty} e^{-bn} = \frac{1}{1 - e^{-b}} < \infty.$$

\Rightarrow The system $h(n)$ is stable.

(c) $H(\omega) = \sum_{n=0}^{\infty} e^{-bn} e^{-j\omega n} = \sum_{n=0}^{\infty} e^{-(b+j\omega)n}$

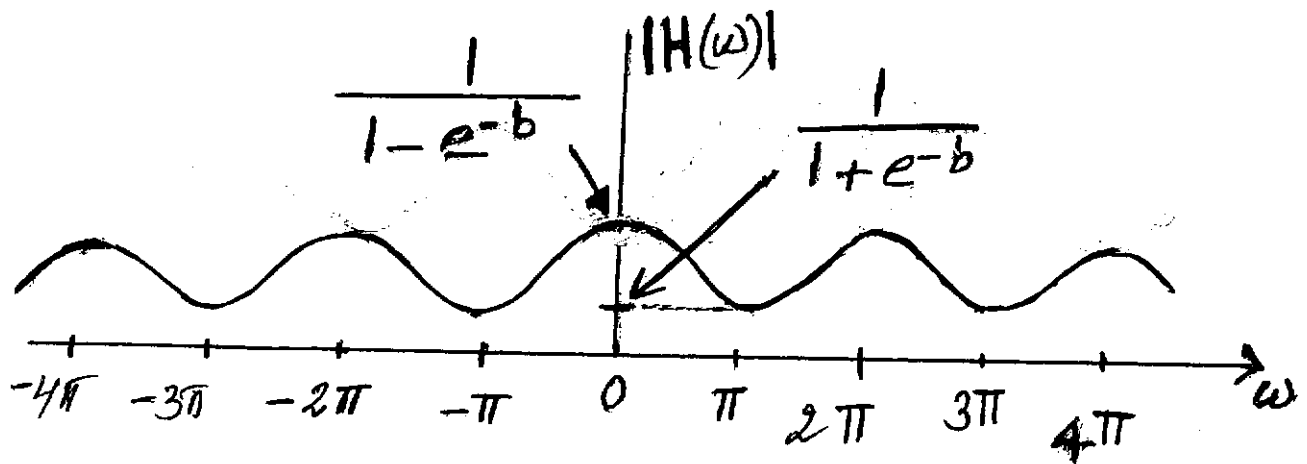
$$= \frac{1}{1 - e^{-b-j\omega}}$$

$$|H(\omega)| = \frac{1}{|1 - e^{-b}(\cos\omega - j\sin\omega)|}$$

$$= \frac{1}{[(1 - e^{-b}\cos\omega)^2 + (e^{-b}\sin\omega)^2]^{1/2}}$$

$$= \frac{1}{[1 - 2e^{-b}\cos(\omega) + e^{-2b}\cos^2\omega + e^{-2b}\sin^2\omega]^{1/2}}$$

$$= \frac{1}{[1 - 2e^{-b}\cos\omega + e^{-2b}]^{1/2}}$$



Problem # 4

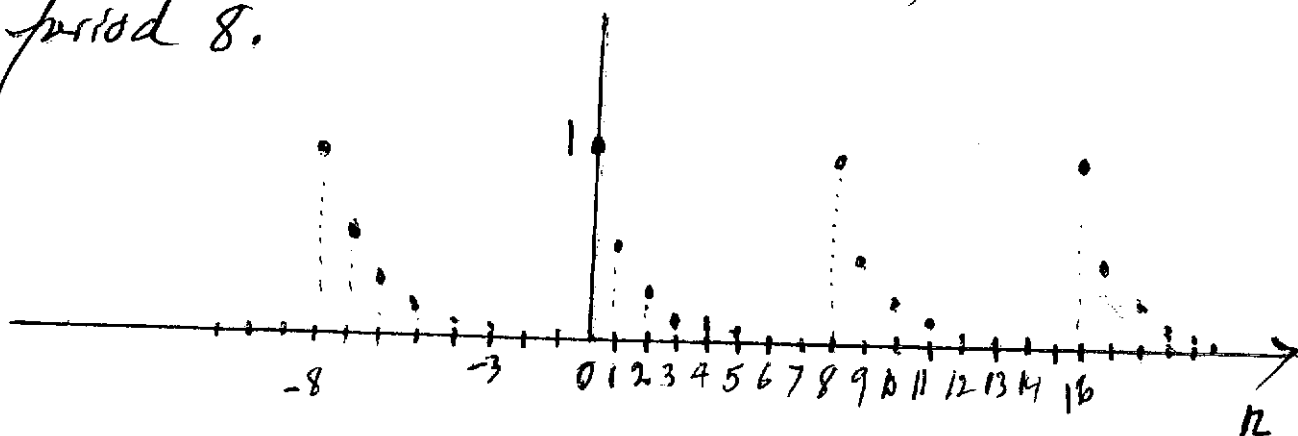
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$$x(n) = \begin{cases} \left(\frac{1}{2}\right)^n, & 0 \leq n \leq 5 \\ 0, & \text{elsewhere.} \end{cases}$$

$$\begin{aligned} (a) X(z) &= \sum_{n=0}^5 \left(\frac{1}{2}\right)^n z^{-n} = \sum_{n=0}^5 \left(\frac{1}{2} z^{-1}\right)^n \\ &= \frac{1 - \left(\frac{1}{2} z^{-1}\right)^6}{1 - \frac{1}{2} z^{-1}} = \frac{1 - \frac{1}{64} z^{-6}}{1 - \frac{1}{2} z^{-1}} \end{aligned}$$

$$\begin{aligned} (b) \tilde{X}(k) &= X(z) \Big|_{z = e^{j\frac{2\pi}{8}k}} \\ &= \frac{1 - \frac{1}{64} e^{-j\frac{12\pi}{8}k}}{1 - \frac{1}{2} e^{-j\frac{2\pi}{8}k}} \end{aligned}$$

$\tilde{X}(k)$ as expressed above is the Fourier Series coefficients of the periodic extension of $x(n)$ with period 8.



(c) $X(k)$ of $x(n)$ considered to be of length 8 is:

$$X(k) = \sum_{n=0}^8 x(n) W_8^{kn} = X(z) \Big|_{z=W_8^{-k}}$$

$\Rightarrow X(k) = \tilde{X}(k)$ as determined in (b).

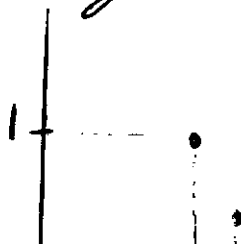
$$\Rightarrow X(k) = \frac{1 - \frac{1}{64} e^{-j \frac{12\pi}{8} k}}{1 - \frac{1}{2} e^{-j \frac{2\pi}{8} k}}$$

$$X(0) = 1 - \frac{1}{64} / 1 - \frac{1}{2} = 2 \times \frac{63}{64} = \frac{63}{32}$$

$$X(1) = \frac{1 - \frac{1}{64} \left(\cos \frac{12\pi}{8} - j \sin \frac{12\pi}{8} \right)}{1 - \frac{1}{2} \left(\cos \frac{2\pi}{8} - j \sin \frac{2\pi}{8} \right)}$$

=

(d) The sequence $x_1(n)$ such that $X_1(k) = W_8^{-4k} X(k)$ is obtained from $x(n)$ by a cyclic shift of amount 4. $x(n)$ is considered of length 8.



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EECE 440 – SIGNALS and SYSTEMS

Final

SPRING 2005-2006

June 6, 2006

TIME: 2 Hours

CLOSED BOOK EXAM

TWELVE SHEETS OF FORMULAS ARE ALLOWED

INSTRUCTOR: Dr. JEAN SAADE

NAME: _____ **ID #:** _____

INSTRUCTIONS

- WRITE YOUR ID # AND NAME ON THE COMPUTER CARD AND ON THIS SHEET IN THE PROVIDED SPACES.
- PROVIDE YOUR ANSWER ON THE COMPUTER CARD and solution of each problem on the scratch booklet
- Random checking will be done to find out about any inconsistency between the problem solutions and the provided answers on the computer card.
- RETURN THE COMPUTER CARD ATTACHED ON TOP OF THE QUESTION SHEET AND SCRATCH BOOKLET.
- USE PENCIL FOR MARKING YOUR ANSWERS AND ID # ON THE COMPUTER CARD.
- ONLY YOUR ANSWER PROVIDED ON THE COMPUTER CARD WILL BE CONSIDERED IN GRADING.
- All QUESTIONS ARE EQUALLY WEIGHTED IN GRADING.

Problem # 1

Consider a linear shift-invariant discrete time system with impulse response given by

$$h(n) = a^n u(n).$$

Let $x(n) = u(n)$ be the input to the system. Determine the output $y(n)$ of the system and then identify in this output the transient response, $y_t(n)$. Note that the transient response has the form of the system impulse response. Also, determine the condition that makes the transient response decay to zero when n tends to infinity.

$$(a) \quad y_t(n) = \frac{a^{n+1}}{a-1} u(n), \quad |a| > 1.$$

$$\textcircled{(b)} \quad y_t(n) = \frac{a^{n+1}}{a-1} u(n), \quad |a| < 1.$$

$$(c) \quad y_t(n) = \frac{a^n}{a-1} u(n), \quad |a| < 1.$$

$$(d) \quad y_t(n) = \frac{a^n}{a-1} u(n), \quad |a| > 1.$$

$$(e) \quad y_t(n) = a^n u(n), \quad |a| < 1.$$

Problem # 2

Consider the linear and time invariant (LTI) finite impulse response (FIR) system whose impulse response is given by

$$h(n) = a^n u(n) - a^n u(n-3)$$

with a being a finite real number larger than 1. Determine the transfer function, $H(z)$, of the system and specify whether the system is or is not stable.

$$(a) \quad H(z) = \frac{z^3 - a^3}{z^2}, \quad \text{and the system is stable.}$$

$$(b) \quad H(z) = \frac{z^3 - (1/a)^3}{z^2[z - (1/a)]}, \quad \text{and the system is stable.}$$

$$(c) \quad H(z) = \frac{z^3 - a^3}{z^2(z-a)}, \quad \text{and the system is not stable.}$$

$$\textcircled{(d)} \quad H(z) = \frac{z^3 - a^3}{z^2(z-a)}, \quad \text{and the system is stable.}$$

$$(e) \quad H(z) = \frac{z^3 - a^3}{z^2}, \quad \text{and the system is not stable.}$$

Problem # 3

Consider again the FIR system given in Problem # 2 and determine the difference equation for this system.

- (a) $y(n) + ay(n-1) = x(n) + a^2x(n-2)$
- (b) $y(n) - y(n-1) = x(n) + a^2x(n-2)$
- ☒ (c) $y(n) = x(n) + ax(n-1) + a^2x(n-2)$
- (d) $y(n) = x(n) + ax(n-1)$
- (e) $y(n) - ay(n-1) = x(n) + a^2x(n-2)$

Problem # 4

Consider the following two discrete time sequences:

$$x(n) = a^n u(n) \text{ and } h(n) = b^{-n} u(n).$$

Let the sequence $y(n)$ be given by the multiplication of $x(n)$ and $h(n)$. That is,

$$y(n) = x(n)h(n).$$

Determine the Fourier transform of $y(n)$; i.e., $Y(\omega)$, and the condition under which this transform exists.

- (a) $Y(\omega) = (1 - \frac{a}{b} e^{-j\omega})^{-1}, \left| \frac{a}{b} \right| > 1$
- (b) $Y(\omega) = (1 - ab e^{j\omega})^{-1}, |ab| < 1$
- (c) $Y(\omega) = (1 - ab e^{-j\omega})^{-1}, |ab| < 1$
- (d) $Y(\omega) = (1 - \frac{a}{b} e^{j\omega})^{-1}, \left| \frac{a}{b} \right| < 1$
- ☒ (e) $Y(\omega) = (1 - \frac{a}{b} e^{-j\omega})^{-1}, \left| \frac{a}{b} \right| < 1$

Problem # 5

Consider the following Z-transform, $X(z)$, of a discrete time sequence, $x(n)$:

$$X(z) = \frac{z^4 + 2z^3 - z + 2}{z^2}, \quad 0 < |z| < \infty.$$

Determine the sequence $x(n)$.

$$\begin{aligned}
 (a) \ x(n) &= \begin{cases} 1, n = -2 \\ 2, n = -1 \\ -1, n = 1 \\ 2, n = 2 \\ 0, \text{ elsewhere} \end{cases} & (b) \ x(n) &= \begin{cases} -1, n = -2 \\ -2, n = -1 \\ 1, n = 1 \\ -2, n = 2 \\ 0, \text{ elsewhere} \end{cases} & (c) \ x(n) &= \begin{cases} 1, n = -2 \\ 2, n = 0 \\ -1, n = 1 \\ 2, n = 2 \\ 0, \text{ elsewhere} \end{cases} \\
 (d) \ x(n) &= \begin{cases} -1, n = -2 \\ -2, n = 0 \\ 1, n = 1 \\ -2, n = 2 \\ 0, \text{ elsewhere} \end{cases} & (e) \ x(n) &= \begin{cases} -1, n = -1 \\ -2, n = 0 \\ 1, n = 1 \\ -2, n = 2 \\ 0, \text{ elsewhere} \end{cases}
 \end{aligned}$$

Problem # 6

Consider a linear and time-invariant system whose transfer function, $H(z)$, is given by

$$H(z) = \frac{z}{(z-a)(z-b)}, \text{ with } |a| > |b| > 1.$$

Determine the impulse response, $h(n)$, of the system assuming that the system is causal.

$$\begin{aligned}
 (a) \ h(n) &= \frac{a^n - b^n}{a - b}, n \geq 2 & (b) \ h(n) &= \frac{a^n + b^n}{a - b}, n \geq 0 & (c) \ h(n) &= \frac{a^n - b^n}{a - b}, n \geq 0 \\
 (d) \ h(n) &= \frac{a^n + b^n}{a - b}, n \geq 2 & (e) \ h(n) &= \frac{a^n b^n}{a - b}, n \geq 0
 \end{aligned}$$

Problem # 7

Consider again the given in Problem # 6 and determine the impulse response, $h(n)$, of the system assuming that the system is stable.

$$\begin{aligned}
 (a) \ h(n) &= \frac{a^n b^n}{b - a}, n \leq 0 & (b) \ h(n) &= \frac{a^n + b^n}{b - a}, n \leq 0 & (c) \ h(n) &= \frac{a^n - b^n}{a - b}, n \leq 0 \\
 (d) \ h(n) &= \frac{a^n - b^n}{b - a}, n \leq 0 & (e) \ h(n) &= \frac{a^n + b^n}{a - b}, n \leq 0
 \end{aligned}$$

Problem # 8

Consider again the system whose transfer function is given in Problem # 6. Let this system be implemented using the block diagram form obtained from the system difference equation. The implemented system would turn out to be:

- ☒ (a) Causal but not stable
- (b) Causal and stable
- (c) Stable but not causal
- (d) Causal and stable but not time invariant
- (e) Causal and stable and time invariant

Problem # 9

Consider the cascaded connection of two causal, linear and time-invariant analog systems with transfer functions given by:

$$H_1(s) = \frac{s+1}{(s-3)(s+5)} \quad \text{and} \quad H_2(s) = \frac{P(s)}{(s+2)(s+3)}.$$

$P(s)$ is a polynomial in s of degree less than or equal 2. Select from what is given below the polynomial $P(s)$ that makes the system that is equivalent to the cascaded connection of $H_1(s)$ and $H_2(s)$ stable and having the transient response that decays the fastest possible to zero as time goes to infinity.

- (a) $P(s) = (s-3)$
- ☒ (b) $P(s) = (s-3)(s+2)$
- (c) $P(s) = (s-3)(s+5)$
- (d) $P(s) = (s-3)(s+3)$
- (e) $P(s) = (s+2)(s+3).$

Problem # 10

Consider the cascaded connection of two causal, linear and time-invariant discrete systems with transfer functions given by:

$$H_1(z) = \frac{z+1}{\left(z - \frac{1}{2}\right)(z-2)} \quad \text{and} \quad H_2(z) = \frac{P(z)}{\left(z - \frac{1}{4}\right)\left(z - \frac{3}{4}\right)}.$$

$P(z)$ is a polynomial in z of degree less than or equal 2. Select from what is given below the polynomial $P(z)$ that makes the system that is equivalent to the cascaded connection of $H_1(z)$ and $H_2(z)$ stable and having the transient response that decays the fastest possible to zero as time goes to infinity.

$$(a) P(z) = (z - 2)$$

$$(b) P(z) = (z - 2) \left(z - \frac{1}{2} \right)$$

$$(c) P(z) = (z - 2) \left(z - \frac{1}{4} \right)$$

$$(d) P(z) = \left(z - \frac{1}{2} \right) \left(z - \frac{3}{4} \right)$$

$$(e) P(z) = (z - 2) \left(z - \frac{3}{4} \right)$$

Problem # 11

Consider a linear and time-invariant analog system with impulse response given by:

$$h(t) = 2e^{-t}u(t) + e^{2t}u(-t)$$

Determine the ROC of the system transfer function, $H(s)$, and specify if the system is or is not stable.

(a) $-1 < \sigma < 2$ and the system is *not* stable.

☒ (b) $-1 < \sigma < 2$ and the system is stable.

(c) $-2 < \sigma < 1$ and the system is *not* stable.

(d) $-2 < \sigma < 1$ and the system is stable.

(e) $-1 < \sigma < 1$ and the system is stable.

Problem # 12

Consider the linear and time-invariant discrete system whose impulse response is obtained by sampling the impulse response of the analog system given in Problem # 11. Determine the ROC of the discrete system transfer function, $H(z)$, and specify if the system is or is not stable.

(a) $e^{-2} < |z| < e$ and the system is *not* stable.

(b) $e^{-2} < |z| < e$ and the system is stable.

(c) $e^{-1} < |z| < e^2$ and the system is *not* stable.

☒ (d) $e^{-1} < |z| < e^2$ and the system is stable.

(e) $e^{-1} < |z| < e$ and the system is stable.

Problem # 13

Consider the following finite duration discrete time sequence:

$$x(n) = \begin{cases} 2^n, & 0 \leq n \leq 2 \\ 0, & \text{elsewhere} \end{cases}$$

Use the DFT, $X(k)$, of $x(n)$, that represents the coefficients of the DFS representation of the periodic repetition of $x(n)$ with period equal to 3 to determine $X(0)$ and $X(1)$.

- (a) $X(0) = 7, X(1) = -3 + 2.732j$
- ☒ (b) $X(0) = 7, X(1) = -2 + 1.732j$
- (c) $X(0) = 7, X(1) = 2 - 1.732j$
- (d) $X(0) = 7, X(1) = 3 - 2.732j$
- (e) $X(0) = 7, X(1) = -4 + 2.732j$

Problem # 14

Consider the same discrete time sequence, $x(n)$, given in Problem # 13. Use the DFT, $X(k)$, that represents samples from the Fourier transform of $x(n)$ at $\omega = (2\pi/10)k, k=0, 1, \dots, 9$ and determine $X(0)$ and $X(1)$.

- (a) $X(0) = 7, X(1) = -3.854 + 4.979j$
- ☒ (b) $X(0) = 7, X(1) = 3.854 - 4.979j$
- (c) $X(0) = 7, X(1) = 4.854 - 5.979j$
- (d) $X(0) = 7, X(1) = -4.854 + 5.979j$
- (e) $X(0) = 7, X(1) = -5.854 + 6.979j$

Problem # 15

The comparison of the results; i.e., $X(0)$ and $X(1)$, in Problems # 13 and 14 shows that $X(0)$ is the same but $X(1)$ differs. Determine the reason for this difference in the values of $X(1)$.

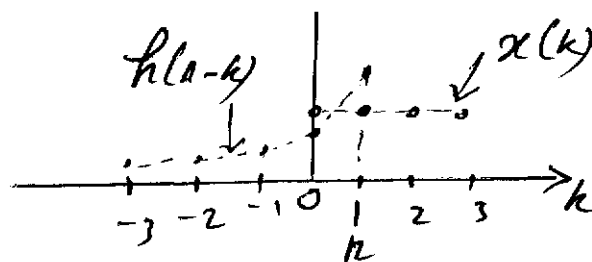
- (a) $X(k)$ in Problem # 13 represents samples from $X(\omega)$ at $\omega = (2\pi/4)k, k=0, 1, 2, 3$.
- (b) $X(k)$ in Problem # 13 represents samples from $X(\omega)$ at $\omega = (2\pi/6)k, k=0, 1, 2, 3, 4, 5$.
- ☒ (c) $X(k)$ in Problem # 13 represents samples from $X(\omega)$ at $\omega = (2\pi/3)k, k=0, 1, 2$.
- (d) $X(k)$ in Problem # 13 represents samples from $X(\omega)$ at $\omega = (2\pi/5)k, k=0, 1, 2, 3, 4$.
- (e) $X(k)$ in Problem # 13 represents samples from $X(\omega)$ at $\omega = (2\pi/7)k, k=0, 1, 2, 3, 4, 5, 6$.

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Problem #1

$$h(n) = a^n u(n), \quad x(n) = u(n).$$

$$y(n) = \sum_{k=-\infty}^{\infty} x(k) h(n-k)$$



$$n < 0, y(n) = 0.$$

$$n \geq 0, y(n) = \sum_{k=0}^n a^{n-k}$$

$$= a^n \sum_{k=0}^n (a^{-1})^k = a^n \left[\frac{1 - (a^{-1})^{n+1}}{1 - a^{-1}} \right]$$

$$= \frac{a^{n+1} - 1}{a - 1}$$

$$= \frac{a^{n+1} u(n)}{a - 1} - \frac{u(n)}{a - 1} = y_t(n) + y_s(n).$$

\Rightarrow the transient response of the system is:

$$y_t(n) = \frac{a^{n+1}}{a - 1} u(n).$$

If $|a| < 1$, then $y_t(n) \rightarrow 0$ when $n \rightarrow \infty$.

Note that $y_t(n)$ has the same form of the system impulse response. Also, $|a| < 1$ is the condition

Problem #2

$$\begin{aligned}
 h(n) &= a^n u(n) - a^n u(n-3) \\
 &= \begin{cases} 1, & n=0 \\ a, & n=1 \\ a^2, & n=2, \text{ and } 0 \text{ elsewhere.} \end{cases}
 \end{aligned}$$

$$\begin{aligned}
 H(z) &= \sum_{n=0}^2 h(n) z^{-n} = 1 + az^{-1} + a^2 z^{-2} \\
 &= \frac{z^2 + az + a^2}{z^2} = \frac{(z-a)(z^2 + az + a^2)}{z^2(z-a)} \\
 &= \frac{z^3 - a^3}{z^2(z-a)}, \text{ for } 0 < |z| \leq \infty
 \end{aligned}$$

The region of convergence contains the unit circle.
Hence, the system is stable.

Problem #3

$$H(z) = \frac{Y(z)}{X(z)} = \frac{z^2 + az + a^2}{z^2} = 1 + az^{-1} + a^2 z^{-2}$$

$$\Rightarrow Y(z) = X(z) + az^{-1}X(z) + a^2 z^{-2}X(z)$$

$$\Rightarrow y(n) = x(n) + ax(n-1) + a^2 x(n-2)$$

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Problem # 4

$$x(n) = a^n u(n), \quad h(n) = b^{-n} u(n)$$

$$y(n) = x(n)h(n) \\ = (ab^{-1})^n u(n)$$

$$Y(\omega) = \sum_{n=0}^{\infty} (ab^{-1})^n e^{-j\omega n} = \sum_{n=0}^{\infty} \left[\frac{a}{b} e^{-j\omega} \right]^n$$

$$= \frac{1}{(1 - \frac{a}{b} e^{-j\omega})} \text{ with } \left| \frac{a}{b} \right| < 1.$$

Problem # 5

$$X(z) = \frac{z^4 + 2z^3 - z + 2}{z^2}, \quad 0 < |z| < \infty.$$

$$X(z) = z^2 + 2z - z^{-1} + 2z^{-2} = \sum_{n=-\infty}^{\infty} x(n) z^{-n}$$

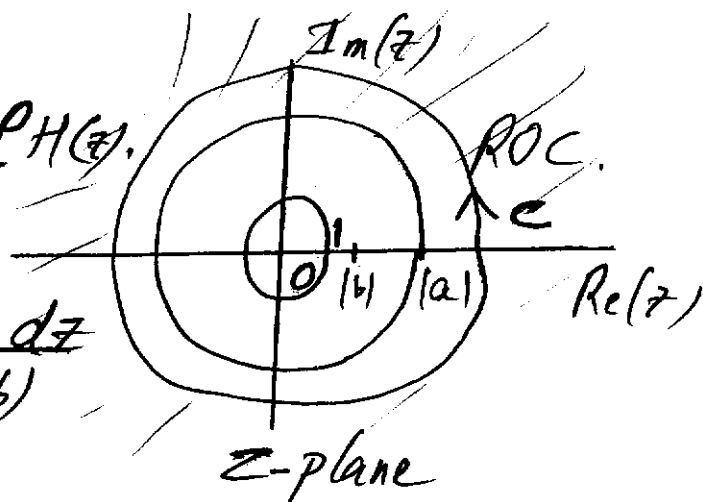
$$\Rightarrow x(n) = \begin{cases} 1, & n = -2 \\ 2, & n = -1 \\ -1, & n = 1 \\ 2, & n = 2 \\ 0, & \text{elsewhere.} \end{cases}$$

Problem # 6

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$$H(z) = \frac{z}{(z-a)(z-b)}, \text{ with } |a| > |b| > 1$$

For the causal system,
 $|z| > |a|$ is the ROC of $H(z)$.



$$h(n) = \frac{1}{2\pi j} \oint_C \frac{z z^{n-1} dz}{(z-a)(z-b)}$$

$$= \frac{1}{2\pi j} \oint_C \frac{z^n dz}{(z-a)(z-b)} = [\text{Res. at } z=a] + [\text{Res. at } z=b]$$

$$= \frac{z^n}{(z-b)} \Big|_{z=a} + \frac{z^n}{(z-a)} \Big|_{z=b}$$

$$= \frac{a^n}{a-b} + \frac{b^n}{b-a} = \frac{a^n - b^n}{a-b}, n \geq 0$$

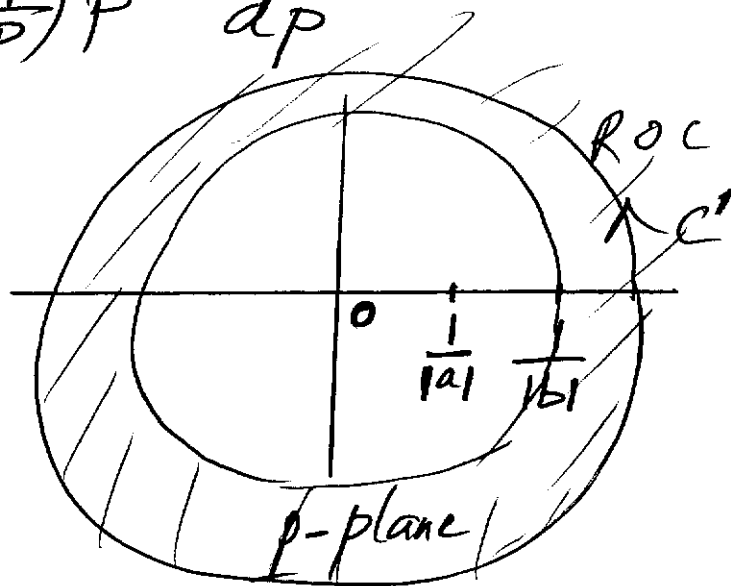
Problem # 7 The stable system has the ROC such that $|z| < |b|$.

$\Rightarrow h(n)$ is a left-sided sequence.

$$H(p^{-1}) = \frac{\frac{1}{p}}{(\frac{1}{p}-a)(\frac{1}{p}-b)} = \frac{p^{-1}}{(1-ap)(1-bp)p^{-2}}$$

$$h(n) = \frac{1}{2\pi j} \oint_{C'} H\left(\frac{1}{p}\right) p^{-n-1} dp$$

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$$h(n) = \frac{1}{2\pi j} \oint_{C'} \frac{p p^{-n-1}}{ab(p - \frac{1}{a})(p - \frac{1}{b})} dp$$

$$= \frac{1}{ab} \frac{1}{2\pi j} \oint_{C'} \frac{p^{-n}}{(p - \frac{1}{a})(p - \frac{1}{b})} dp$$

$$= \frac{1}{ab} \left[\frac{p^{-n}}{p - \frac{1}{b}} \Big|_{p=\frac{1}{a}} + \frac{p^{-n}}{p - \frac{1}{a}} \Big|_{p=\frac{1}{b}} \right]$$

$$= \frac{a^n}{b-a} + \frac{b^n}{a-b} = \frac{a^n - b^n}{b-a}, \quad n \leq 0.$$

Problem #8

The implemented system is the causal system.

But, the causal system is not stable.

The system is linear and time invariant.

Problem # 9

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The cascaded system has a transfer function given by:

$$H(s) = H_1(s) H_2(s) \\ = \frac{(s+1)P(s)}{(s-3)(s+5)(s+2)(s+3)}$$

The polynomial $P(s)$ is of degree ≤ 2 .

$P(s) = (s-3)(s+2)$ cancels $(s-3)$ from the denominator of $H(s)$. This makes the system stable, by deleting the pole at $s=3$ which is located in the right half s -plane. Also, $P(s) = (s-3)(s+2)$ cancels $(s+2)$ from the denominator which provides the slowest decaying component in $h(t)$ or the transient response compared to $(s+3)$ and $(s+5)$.

$s = -2$ is the pole that is the closest to the imaginary axis.

Problem # 10

$$H(z) = H_1(z) H_2(z) \\ = \frac{(z+1)P(z)}{(z-\frac{1}{2})(z-2)(z-\frac{1}{4})(z-\frac{3}{4})}$$

$$P(z) = (z-2)(z-3)$$

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$P(z)$ as selected, deletes the pole at $z=2$, which is outside the unit circle. This makes $H(z)$ stable. It also deletes $z=\frac{3}{4}$ which is the pole the closest to the unit circle. Hence, the transient response component that decays the slowest possible to 0 when n increases is also deleted.

Problem #11

$$h(t) = 2e^{-t}u(t) + e^{2t}u(-t)$$

$$H(s) = \int_{-\infty}^0 e^{2t} e^{-st} dt + \int_0^{\infty} 2e^{-t} e^{-st} dt$$

$$= \int_{-\infty}^0 e^{-(s-2)t} dt + \int_0^{\infty} 2e^{-(s+1)t} dt$$

$$= -\frac{1}{s-2} e^{-(s-2)t} \Big|_{-\infty}^0 + 2 \times \frac{-1}{s+1} e^{-(s+1)t} \Big|_0^{\infty}$$

$$= \frac{-1}{s-2} + \frac{2}{s+1} = \frac{-s+1+2s-4}{(s+1)(s-2)}$$

$$= \frac{s-5}{(s+1)(s-2)}, \quad -1 < \sigma < 2.$$

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Problem #12

$$h(n) = 2e^{-n}u(n) + e^{2n}u(-n).$$

$$\begin{aligned} H(z) &= \sum_{n=-\infty}^0 e^{2n} z^{-n} + 2 \sum_{n=0}^{\infty} e^{-n} z^{-n} \\ &= \sum_{n=-\infty}^0 (e^2 z^{-1})^n + 2 \sum_{n=0}^{\infty} (e^{-1} z^{-1})^n \\ &= \sum_{n=0}^{\infty} (e^{-2} z)^n + 2 \sum_{n=0}^{\infty} (e^{-1} z^{-1})^n \\ &= \frac{1}{1 - e^{-2} z} + \frac{2}{1 - e^{-1} z^{-1}} \end{aligned}$$

with $|e^{-2} z| < 1 \Rightarrow |z| < e^2$

and $|e^{-1} z^{-1}| < 1 \Rightarrow |z| > e^{-1}$

\Rightarrow The ROC is such that, $e^{-1} < |z| < e^2$

The ROC contains the unit circle and the system is stable.

Problem #13

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$$x(n) = \begin{cases} 2^n, & 0 \leq n \leq 2 \\ 0, & \text{elsewhere} \end{cases}$$

$$X(k) = \sum_{n=0}^2 2^n W_3^{kn}, \quad k=0,1,2$$

$$= 1 + 2W_3^k + 4W_3^{2k}$$

$$= 1 + 2e^{-j\frac{2\pi}{3}k} + 4e^{-j\frac{2\pi}{3}2k}$$

$$X(0) = 1 + 2 + 4 = 7$$

$$\begin{aligned} X(1) &= 1 + 2\left(\cos\left(\frac{2\pi}{3}\right) - j\sin\left(\frac{2\pi}{3}\right)\right) + 4\left(\cos\left(\frac{4\pi}{3}\right) - j\sin\left(\frac{4\pi}{3}\right)\right) \\ &= \left[1 + 2\cos\left(\frac{2\pi}{3}\right) + 4\cos\left(\frac{4\pi}{3}\right)\right] - j\left[2\sin\left(\frac{2\pi}{3}\right) + 4\sin\left(\frac{4\pi}{3}\right)\right] \\ &= -2 + 1.732j. \end{aligned}$$

Problem #14

$$\begin{aligned} X(k) &= \sum_{n=0}^2 2^n W_{10}^{kn} = \sum_{n=0}^2 2^n e^{-j\frac{2\pi}{10}kn} \\ &= \sum_{n=0}^2 2^n e^{-j\omega n} \bigg|_{\omega = \frac{2\pi}{10}k}, \quad k=0,1,2,\dots,9. \\ &= 1 + 2e^{-j\frac{2\pi}{10}k} + 4e^{-j\frac{2\pi}{10}2k} \end{aligned}$$

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$$X(0) = 1 + 2 + 4 = 7$$

$$\begin{aligned} X(1) &= 1 + 2 \left(\cos\left(\frac{2\pi}{10}\right) - j \sin\left(\frac{2\pi}{10}\right) \right) \\ &\quad + 4 \left(\cos\left(\frac{4\pi}{10}\right) - j \sin\left(\frac{4\pi}{10}\right) \right) \\ &= 1 + 2 \cos\left(\frac{2\pi}{10}\right) + 4 \cos\left(\frac{4\pi}{10}\right) \\ &\quad - j \left(2 \sin\left(\frac{2\pi}{10}\right) + 4 \sin\left(\frac{4\pi}{10}\right) \right) \\ &= 3.854 - 4.979j. \end{aligned}$$

Problem #15

$X(1)$ in Problem #14 is a sample from $X(\omega)$ at $\omega = \frac{2\pi}{10}$.

In Problem #13:

$$\begin{aligned} X(k) &= \sum_{n=0}^2 2^n W_3^{kn} = \sum_{n=0}^2 2^n e^{-j\frac{2\pi}{3}kn} \\ &= \sum_{n=0}^2 2^n e^{-j\omega n} \Big|_{\omega = \frac{2\pi}{3}k, k=0,1,2} \end{aligned}$$

$\Rightarrow X(1)$ is a sample from $X(\omega)$ at $\omega = \frac{2\pi}{3}$.