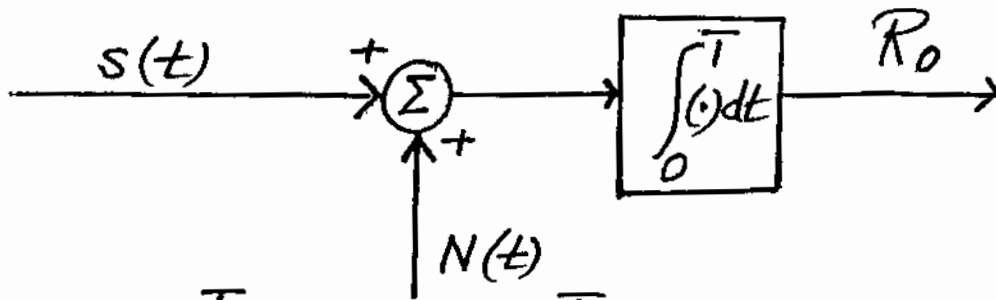


Problem #1

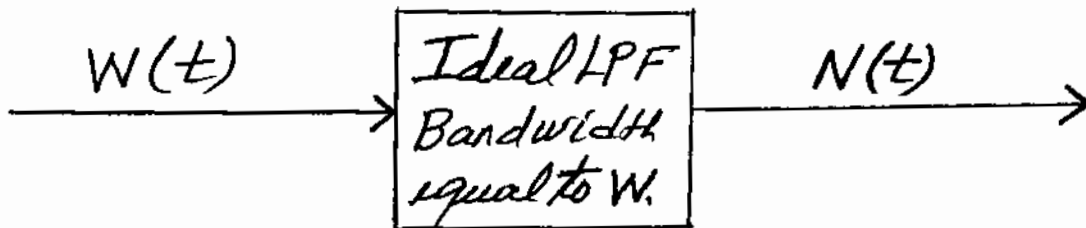


$$R_0 = \int_0^T s(t) dt + \int_0^T N(t) dt$$

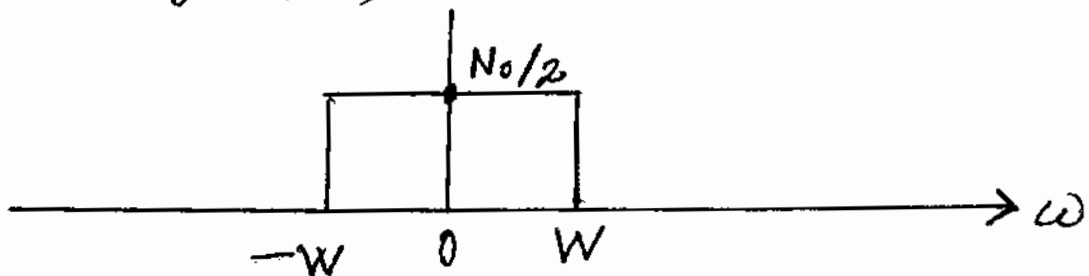
$$E(R_0) = \int_0^T s(t) dt + \int_0^T E[N(t)] dt = \int_0^T s(t) dt.$$

$$\begin{aligned} \sigma_{R_0}^2 &= E[(R_0 - \bar{R}_0)^2] = E\left[\left(\int_0^T N(t) dt\right)^2\right] \\ &= E\left[\int_0^T N(t) dt \int_0^T N(u) du\right] = \int_0^T dt \int_0^T E[N(t)N(u)] du \\ &= \int_0^T dt \int_0^T \frac{N_0}{2} \delta(t-u) du = \frac{N_0}{2} \int_0^T dt = \frac{N_0}{2} T. \end{aligned}$$

Problem #2



The PSD of  $N(t)$  is:



$$R_N(0) = E[N^2(t)] = \frac{1}{2\pi} \int_{-W}^W \frac{N_0}{2} d\omega = \frac{N_0 W}{2\pi}$$

Hence, from among the given auto correlation functions,  $R_N(\tau) = \frac{N_0 W}{2\pi} e^{-2|\tau|}$  can be the auto correlation function of  $N(t)$ . Also, knowing that  $R_N(\tau)$  is even; i.e.,  $R_N(\tau) = R_N(-\tau)$ , then  $R_N(\tau) = \frac{N_0 W}{2\pi} e^{-2\tau}$  cannot be the auto correlation function of  $N(t)$ .

### Problem #3

$E[N(t)] = E[W(t)] \times H(0)$ , where  $H(0)$  is the value of the LPF frequency response at  $\omega = 0$ .

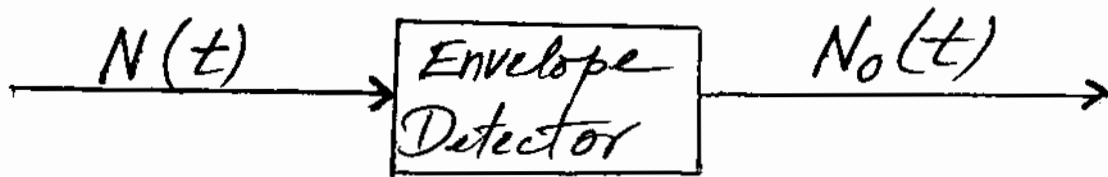
Hence,  $E[N(t)] = 0$  since  $E[W(t)] = 0$ .

$$\begin{aligned} \sigma_N^2 &= E[(N(t) - \bar{N}(t))^2] = E[N^2(t)] = R_N(0) \\ &= \frac{N_0 W}{2\pi} \end{aligned}$$

# Problem # 4

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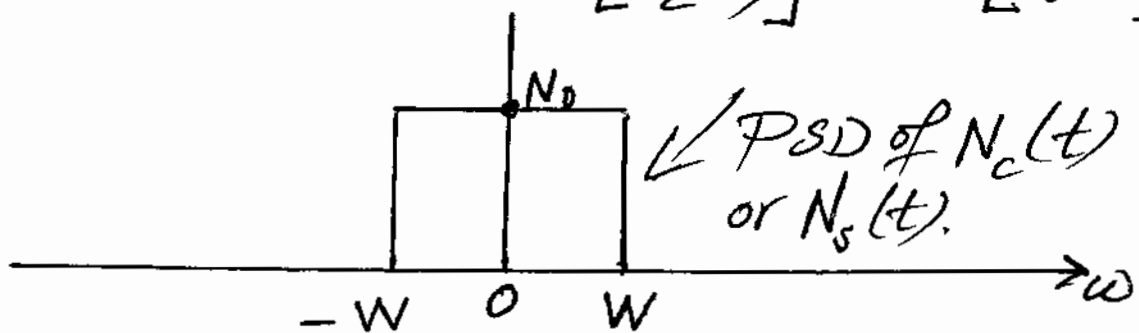
$$\begin{aligned} N(t) &= N_c(t) \cos(\omega_c t) - N_s(t) \sin(\omega_c t) \\ &= R(t) \cos[\omega_c t + \psi(t)] \end{aligned}$$



Average power of  $N_0(t)$  is equal to  $E[N_0^2(t)]$ .

$$\text{But, } N_0(t) = [N_c^2(t) + N_s^2(t)]^{1/2}$$

$$\begin{aligned} \text{Hence, } E[N_0^2(t)] &= E[N_c^2(t) + N_s^2(t)] \\ &= 2E[N_c^2(t)] = 2E[N_s^2(t)]. \end{aligned}$$



$$\text{Hence, } E[N_c^2(t)] = E[N_s^2(t)] = \frac{N_0 W}{\pi}$$

$$\Rightarrow E[N_0^2(t)] = \frac{2N_0 W}{\pi}$$

## Problem #5

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$$\begin{aligned} N(t) &= N_c(t) \cos(\omega_c t) - N_s(t) \sin(\omega_c t) \\ &= R(t) \cos[\omega_c t + \psi(t)] \\ &= R(t) \cos \psi(t) \cos(\omega_c t) - R(t) \sin \psi(t) \sin(\omega_c t). \end{aligned}$$

Hence,  $N_c(t) = R(t) \cos \psi(t)$

$$N_s(t) = R(t) \sin \psi(t).$$

$$\Rightarrow R^2(t) = N_c^2(t) + N_s^2(t) = N_o^2(t)$$

$$\Rightarrow R(t) = [N_c^2(t) + N_s^2(t)]^{1/2}$$

Hence, the average power of  $R(t)$  is:

$$E[R^2(t)] = E[N_o^2(t)] = \frac{2N_o W}{\pi}$$

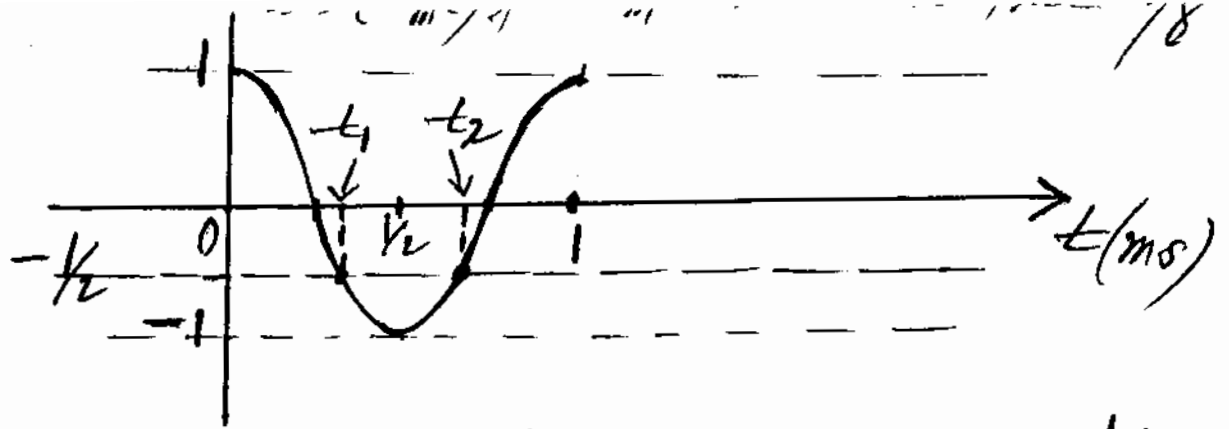
## Problem #6

$$s(t) = A_c [1 + 2 \cos(\omega_m t)] \cos(\omega_c t)$$

The output of the envelope detector is:

$$\begin{cases} -A_c [1 + 2 \cos(\omega_m t)] & \text{when } (1 + 2 \cos(\omega_m t)) < 0 \\ A_c [1 + 2 \cos(\omega_m t)] & \text{when } (1 + 2 \cos(\omega_m t)) > 0. \end{cases}$$

$$1 + 2 \cos(\omega_m t) < 0 \Rightarrow \cos(\omega_m t) < -\frac{1}{2}$$



$t_1$  can be obtained using  $\cos(\omega_m t_1) = -1/2$

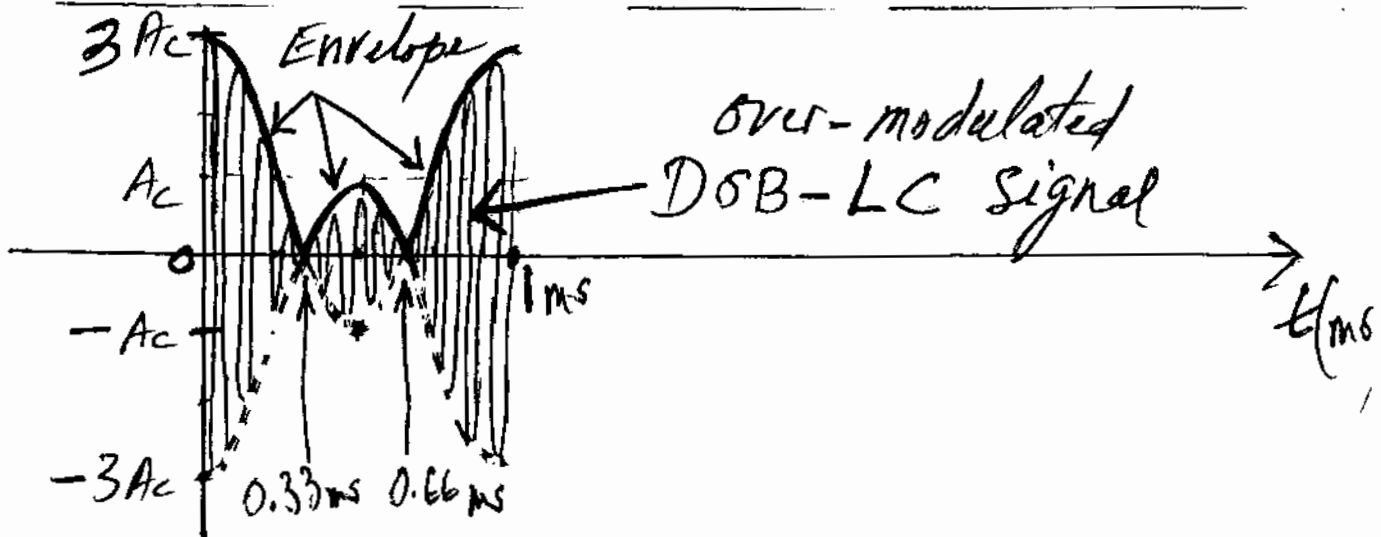
$$\Rightarrow t_1 = \frac{\cos^{-1}(-1/2)}{\omega_m} = \frac{2\pi}{3} \times \frac{1}{2\pi \times 10^3} = 0.33 \text{ ms.}$$

$$t_2 = 1 - \frac{1}{3} = 0.66 \text{ ms.}$$

Hence, the output of the envelope detector is:

$$\begin{cases} -A_c [1 + 2\cos(\omega_m t)], & 0.33 \text{ ms} \leq t \leq 0.66 \text{ ms} \\ A_c [1 + 2\cos(\omega_m t)], & \text{otherwise} \end{cases}$$

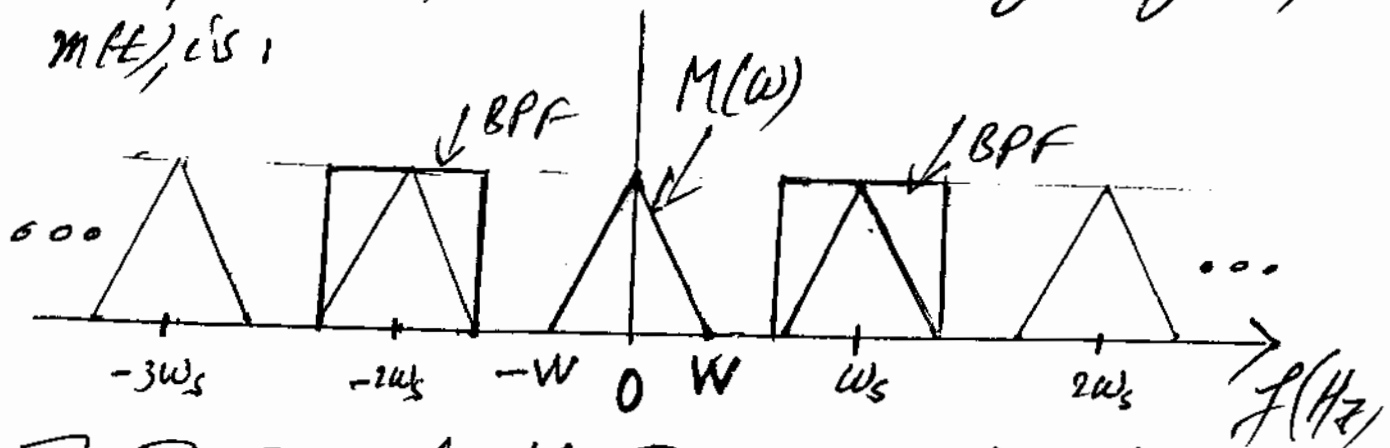
and this is for one period of the message signal or  $\cos(\omega_m t)$ .



## Problem # 7

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The spectrum of the sampled message signal,  $m(t)/T_s$ , is:



The BPF picks the DSB-SC signal with carrier frequency equal to  $\omega_s$ .

$$\text{But, } \omega_s = \frac{1}{T_s} = 500 \text{ kHz.}$$

$$\Rightarrow T_s = \frac{1}{5 \times 10^5} = 2 \times 10^{-6} \text{ sec or } 2 \mu\text{sec.}$$

The maximum value of  $W$  is such that:

$$\omega_s - W = W \text{ or } \omega_s = 2W$$

$$\Rightarrow W = \frac{\omega_s}{2} = \frac{500 \text{ kHz}}{2} = 250 \text{ kHz.}$$

## Problem # 8

By the mixing operation:

$$A_c m(t) \cos(\omega_c t) \cos(\omega_p t) =$$

$$\frac{A_c m(t)}{2} [\cos(\omega_c + \omega_p)t + \cos(\omega_c - \omega_p)t]$$

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Since  $f_{c_1} = 1200 \text{ KHz}$ ,  $f_{c_2} = 900 \text{ KHz}$ ,  
and  $f_{IF} = 500 \text{ KHz}$ , then the frequency  
translation is to be done using  $\frac{A_c}{2} m(t) \cos(\omega_c - \omega_p)t$ .

Hence,  $f_{c_1} - f_{p_1} = f_{IF}$  and  $f_{c_2} - f_{p_2} = f_{IF}$ .

$$\Rightarrow f_{p_1} = f_{c_1} - f_{IF} = 1200 - 500 = 700 \text{ KHz.}$$

$$f_{p_2} = f_{c_2} - f_{IF} = 900 - 500 = 400 \text{ KHz.}$$

### Problem # 9

$$S_{app}(t) = A_c \cos(\omega_c t) - \beta A_c \sin(\omega_m t) \sin(\omega_c t).$$

$$\begin{aligned} \text{Env}[S_{app}(t)] &= [A_c^2 + \beta^2 A_c^2 \sin^2(\omega_m t)]^{1/2} \\ &= A_c [1 + \beta^2 \sin^2(\omega_m t)]^{1/2} \end{aligned}$$

$$\text{Max Env}[S_{app}(t)] = A_c [1 + \beta^2]^{1/2}$$

Max Deviation of  $\text{Env}[S_{app}(t)]$  from  $A_c$  is:

$$\begin{aligned} A_c - A_c [1 + \beta^2]^{1/2} &= A_c [1 - (1 + \beta^2)^{1/2}] \\ &= A_c [1 - (1 + (0.1)^2)^{1/2}] = 0.005 A_c \end{aligned}$$

## Problem #10

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By Carson's rule:

$$B_T = 2(\Delta\omega + W)$$

$$\Rightarrow 200 \text{ KHz} = 2(\Delta\omega + 52)$$

$$\Rightarrow \Delta\omega = \frac{200 - 104}{2} = 48 \text{ KHz.}$$

But,  $\Delta\omega = k_f |\max m(t)|$

$$\Rightarrow |\max(m(t))| = \frac{\Delta\omega}{k_f} = \frac{48 \times 10^3}{4 \times 10^3} = 12 \text{ Volts.}$$