

Midterm Med 435 Spring 2012

1) a) system is type one in open loop \rightarrow no ss error

b) open loop system:
$$G(s) = \frac{k_p k_v k_m}{\tau_m s + k_v k_m + 1}$$

$$k_{vel} = \lim_{s \rightarrow 0} G(s) = \frac{k_p k_v k_m}{k_v k_m + 1}$$

e_{ss} for ramp input should be 0,01

$$\Rightarrow e_{ss} = \frac{A}{k_v} \quad A=1 \text{ because unit ramp input}$$

$$\Rightarrow 0,01 = \frac{k_p k_v k_m}{k_v k_m + 1} \quad k_m = 1 \text{ given}$$

$$\Rightarrow 0,01 k_v + 0,01 = k_p k_v$$

$$\Rightarrow k_v (k_p - 0,01) = 0,01 \quad \leftarrow \text{relationship between } k_p \text{ and } k_v$$

c) T_s of ϕ required to be 2 seconds in addition to above requirements.

$$G_{cl} = \frac{k_p k_v k_m / \tau_m}{s^2 + \frac{1 + k_v k_m}{\tau_m} s + \frac{1 + k_p k_v k_m}{\tau_m}}$$

$$T_s = 2 \Leftrightarrow \frac{4}{\sigma} = 2 \Leftrightarrow \sigma = 2$$

$$\Rightarrow \frac{1 + k_v k_m}{2 \tau_m} = 2 \quad \text{with } k_m = 1, \tau_m = 0,5$$

$$\Rightarrow 1 + k_v = 2$$

$$\Rightarrow k_v = 2$$

with problem b) we get $k_p = 0,02$

d) if k_v is negative, no clear statement, whether system is still stable or not. For $1 + k_v k_m < 0$, the system will be unstable

Problem 2

Root locus

$$G(s) = \frac{s^2 - s + 1}{s(s+1)}$$

a) locate poles/zeros

$$\text{zeros: } s^2 - s + 1 = 0$$

$$\Leftrightarrow (s - 0,5)^2 = -0,75$$

$$\Rightarrow z_{1,2} = +0,5 \pm j\sqrt{0,75}$$

$$\text{poles: } p_1 = 0$$

$$p_2 = -1$$

b) RL on real axis

c) Asymptotes: none

d) breakaway point expected between -1 and 0

$$\text{Found by } \frac{dK}{ds} = 0 \Leftrightarrow D(s)N'(s) - N(s)D'(s) = 0 \quad \text{with } D(s) = s(s+1)$$

$$N(s) = s^2 - s + 1$$

$$\Leftrightarrow (2s+1)(s^2-s+1) - (2s-1)(s^2+s) = 0$$

$$\Leftrightarrow 2s^3 + s^2 - 2s^2 - s + 2s + 1 - [2s^3 - s^2 + 2s^2 - s] = 0$$

$$\Leftrightarrow 0 + s^2(1-2+1-2) + s(-1+2+1) + 1 = 0$$

$$\Leftrightarrow -2s^2 + 2s + 1 = 0$$

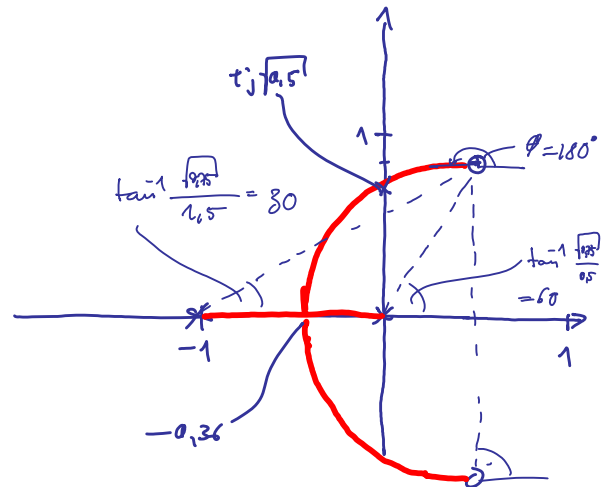
$$\Leftrightarrow s^2 - s - 0,5 = 0$$

$$\Leftrightarrow (s - 0,5)^2 = 0,5 + 0,25$$

$$\Leftrightarrow s_{1,2} = 0,5 \pm \sqrt{0,75}$$

$$s_1 = 1,36$$

$$s_2 = -0,36$$



e) Angle of arrival in zeros:

Apply angle criterion in point infinitely near zero $s \approx z_1$

$$\Rightarrow \underbrace{\frac{s-z_1}{s-z_1}}_{\phi?} + \underbrace{\frac{s-z_2}{s-z_2}}_{90} - \underbrace{\frac{s-p_1}{s-p_1}}_{60} - \underbrace{\frac{s-p_2}{s-p_2}}_{30} = 180^\circ \pm k360^\circ$$

$$\Rightarrow \phi = 180 \pm k360 + 30 + 60 - 90 = 180$$

f) K for marginal stability with Routh:

$$1 + kG(s) = 0$$
$$\Leftrightarrow 1 + k \frac{s^2 - s + 1}{s(s+1)} = 0$$

$$\Leftrightarrow s^2 + 1 + ks^2 - ks + k = 0$$

$$\Leftrightarrow s^2(1+k) + s(1-k) + k = 0$$

$$\begin{array}{ccc} s^2 & 1+k & k \\ s & 1-k & \\ 1 & k & \end{array}$$

\Rightarrow marginal stable for $k=1$

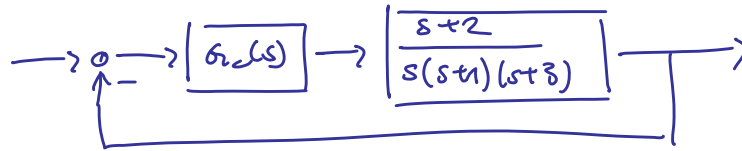
g) \Rightarrow auxiliary polynomial to find pair of critically stable poles:
 $k=1$

$$\Rightarrow s^2 - 2 + 1 = 0$$

$$\Leftrightarrow s^2 = -0,5$$

$$\Leftrightarrow s_{1,2} = \pm j\sqrt{0,5}$$

Problem 3



- a) Design control such that $T_s = 2s$ and ξ of closed loop pole pair is $\xi = \frac{\sqrt{2}}{2}$

$$1) \cos^{-1} \frac{\sqrt{2}}{2} = 45^\circ$$

$$2) T_s = 2 = \frac{4}{\omega} \Rightarrow \omega = 2$$

\rightarrow poles are not on RL!

- b) Lead design:

$$G_c = K_c \alpha \frac{T_s + 1}{aT_s + 1}$$

$$= K_c \frac{s + \frac{1}{T}}{s + \frac{1}{aT}}$$

3 parameters to determine. First determine angle contribution for RL to go through s_{des} .

$$\angle(G_c(s)) + \angle_{s_{des} Z_1} - \angle_{s_{des} P_1} - \angle_{s_{des} P_2} - \angle_{s_{des} P_3} = 180 \pm k360^\circ$$

$$\angle(G_c(s)) = 180 \pm k360^\circ - 90^\circ + 117^\circ + 135^\circ + 63^\circ$$

$$= 45^\circ$$

Next we have to place the pole and zero of the compensator

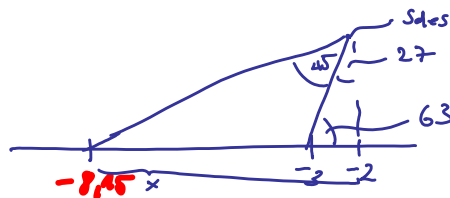
I) Bisector method

II) pole cancellation

\rightarrow try pole cancellation \Rightarrow zero on pole at -3

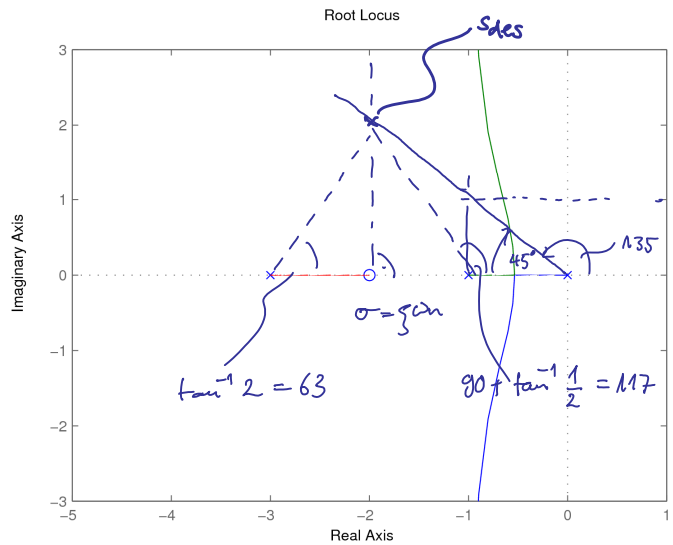
$$\Rightarrow \frac{1}{T} = 3 \Rightarrow T = \frac{1}{3}$$

For angle contribution to be 45° we chose pole at:



$$\tan(45 + 27) = \frac{x}{2}$$

$$\Rightarrow 2 \cdot \tan(45 + 27) = x = 6,45$$



$$\Rightarrow \frac{1}{\zeta T} = +8,15$$

$$\Rightarrow \zeta = \frac{1}{\frac{1}{3} \cdot 8,15} = 0,368$$

At last, we determine k_c using the angle criterion in s_{des} .

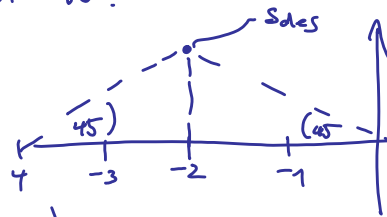
$$k_c \frac{|(s_{des} + 2)| \cdot |(s_{des} + 3)|}{|s_{des}| |s_{des} + 1| |s_{des} + 3| |s_{des} + 8,15|} = 1$$

$$\Leftrightarrow k_c = \frac{\sqrt{8} \cdot \sqrt{5} \cdot \sqrt{10,15}}{2} = 10,25 \Rightarrow G_c = 10,25 \cdot \frac{s+3}{s+6,15}$$

Design PD control:

Angle contribution of PD has to be 45° .

We can simply draw a line from s_{des} which intersects the Real axis at 45° .



$$\Rightarrow G_c = k_p (s - \alpha) \quad \alpha = -4$$

We get k_p through the mag. criterion:

$$k_p \frac{|(s_{des} + 2)| |s_{des} + 4|}{|s_{des}| |s_{des} - 3| |s_{des} + 1|} = 1$$

$$\Leftrightarrow k_p = \frac{\sqrt{8 \cdot 5 \cdot 10,15}}{2 \cdot \sqrt{8}} = 3,56 \Rightarrow G_c = 3,56 \cdot (s + 4)$$