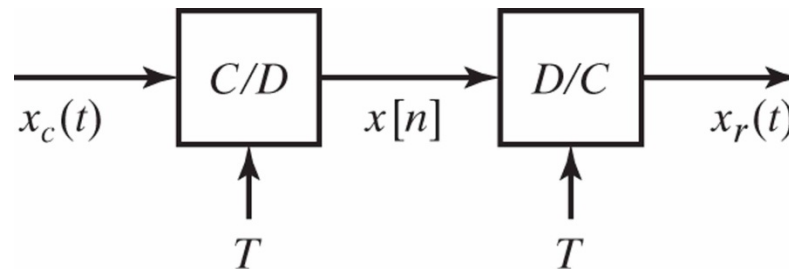
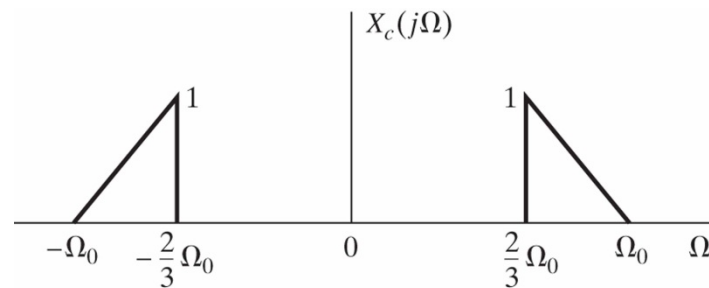

EECE 491: Discrete-time Signal Processing

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American University of Beirut

Practice Problems on
Sampling

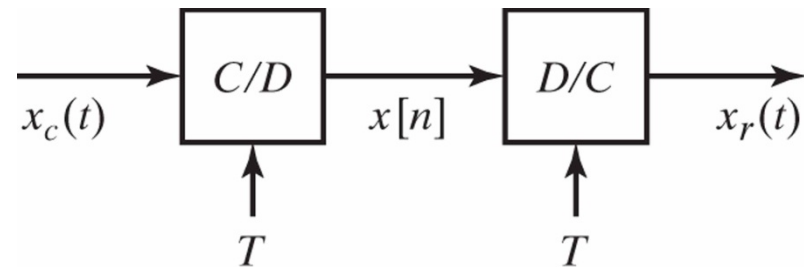
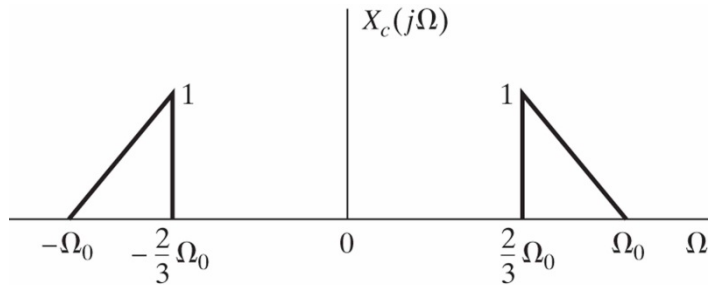
Problem 1

- $x_c(t)$ below is passed through the system on the bottom. Determine the range of values of T for which $x_r(t) = x_c(t)$.



Problem 1 Solution

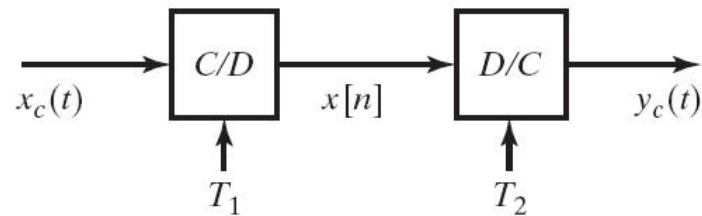
- $x_c(t)$ passed through the system on the right. Determine the range of values of T for which $x_r(t) = x_c(t)$.



- **Solution:** $\Omega_0 T \leq \pi$

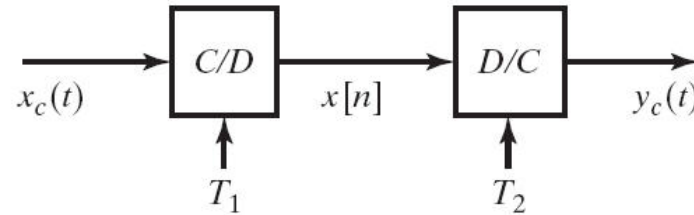
Problem 2

- For the system below, assume $X_c(j\Omega) = 0, |\Omega| \geq \pi/T_1$. Express $y_c(t)$ in terms of $x_c(t)$. Is the relationship different for $T_1 > T_2$ and $T_1 < T_2$?



Problem 2 Solution

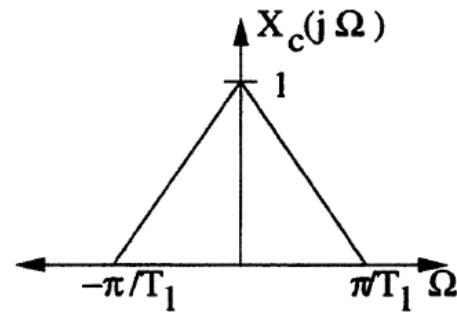
- For the system below, assume $X_c(j\Omega) = 0, |\Omega| \geq \pi/T_1$. Express $y_c(t)$ in terms of $x_c(t)$. Is the relationship different for $T_1 > T_2$ and $T_1 < T_2$?



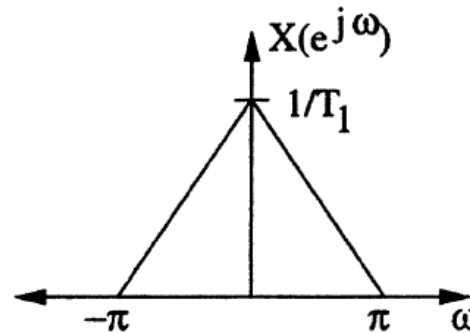
- Solution:**

$$X_c(j\Omega) = 0, |\Omega| \geq \pi/T_1$$

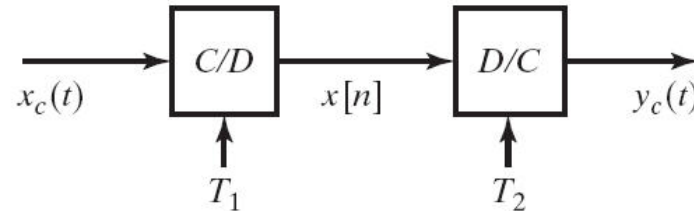
$$x[n] = x_c(nT_1)$$



$$X(e^{j\omega}) = \frac{1}{T_1} \sum_{k=-\infty}^{\infty} X_c\left(j\left(\frac{\omega}{T_1} - \frac{2\pi k}{T_1}\right)\right)$$



Problem 2 Solution (cont'd)



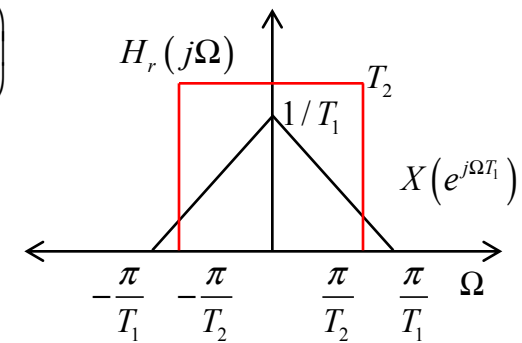
- $y_c(t)$ is a bandlimited interpolation of $x[n]$ at a different period than T_1 :

$$Y_c(j\Omega) = H_r(j\Omega) X(e^{j\Omega T_2})$$

$$= \begin{cases} T_2 X(e^{j\Omega T_2}) & |\Omega| < \pi/T_2 \\ 0 & \text{otherwise} \end{cases}$$

$$= \begin{cases} \frac{T_2}{T_1} \sum_{k=-\infty}^{\infty} X_c \left(j \left(\frac{\Omega T_2}{T_1} - \frac{2\pi k}{T_1} \right) \right) & |\Omega| < \pi/T_2 \\ 0 & \text{otherwise} \end{cases}$$

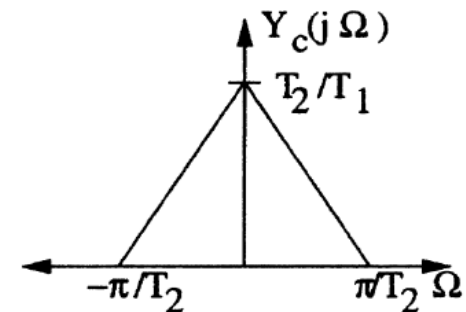
$$X(e^{j\omega}) = \frac{1}{T_1} \sum_{k=-\infty}^{\infty} X_c \left(j \left(\frac{\omega}{T_1} - \frac{2\pi k}{T_1} \right) \right)$$



$$y_c(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} Y_c(j\Omega) e^{j\Omega t} d\Omega$$

$$= \frac{1}{2\pi} \frac{T_2}{T_1} \int_{-\pi/T_2}^{\pi/T_2} \sum_{k=-\infty}^{\infty} X_c \left(j \left(\frac{\Omega T_2}{T_1} - \frac{2\pi k}{T_1} \right) \right) e^{j\Omega t} d\Omega$$

$$y_c(t) = \frac{T_2}{T_1} x_c \left(\frac{T_2}{T_1} t \right)$$



Problem 3

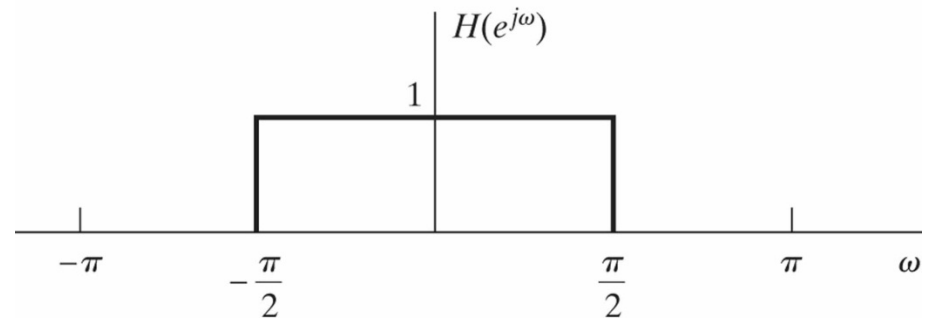
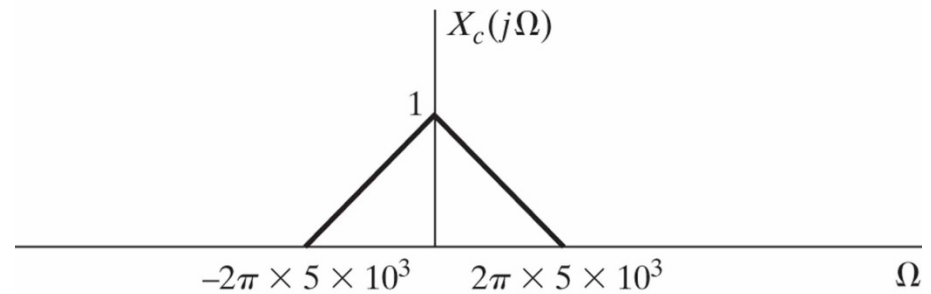
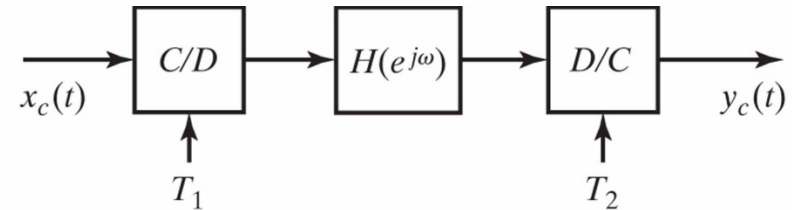
- Consider the following system.
- Sketch and label $Y_c(j\Omega)$ when:

a) $\frac{1}{T_1} = \frac{1}{T_2} = 10^4$

b) $\frac{1}{T_1} = \frac{1}{T_2} = 2 \times 10^4$

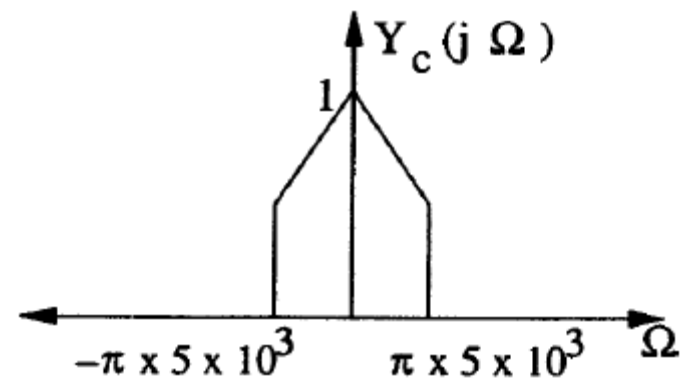
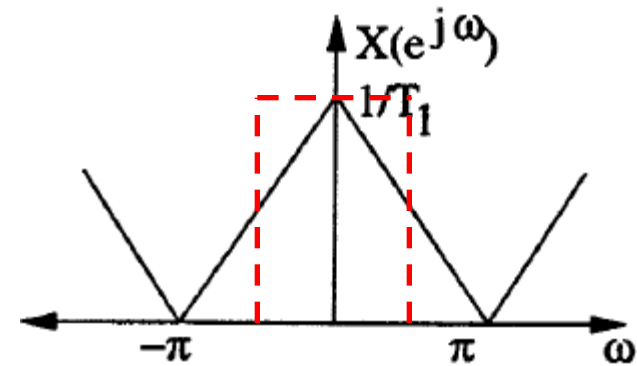
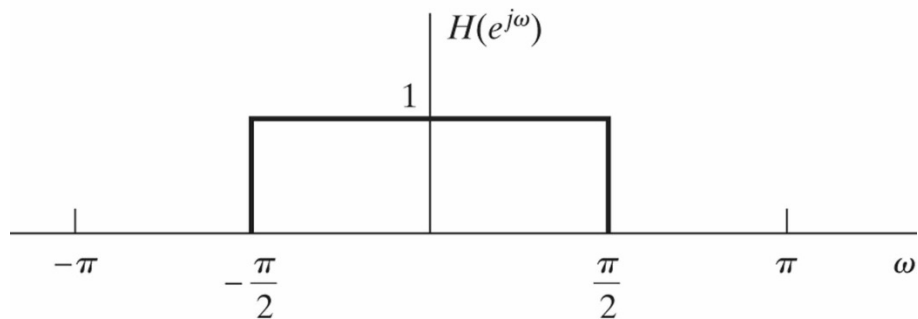
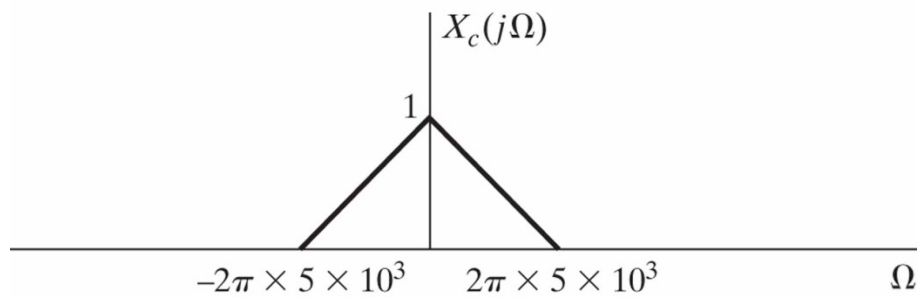
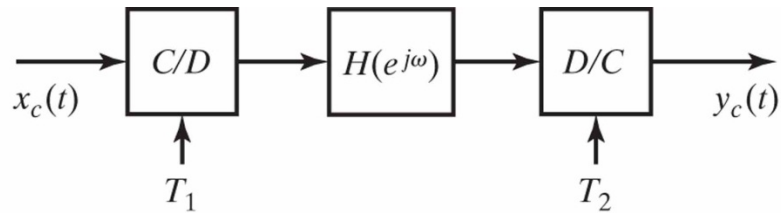
c) $\frac{1}{T_1} = 2 \times 10^4$ $\frac{1}{T_2} = 10^4$

d) $\frac{1}{T_1} = 10^4$ $\frac{1}{T_2} = 2 \times 10^4$



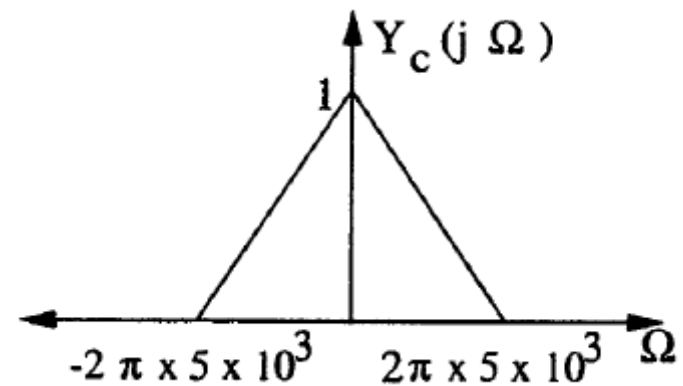
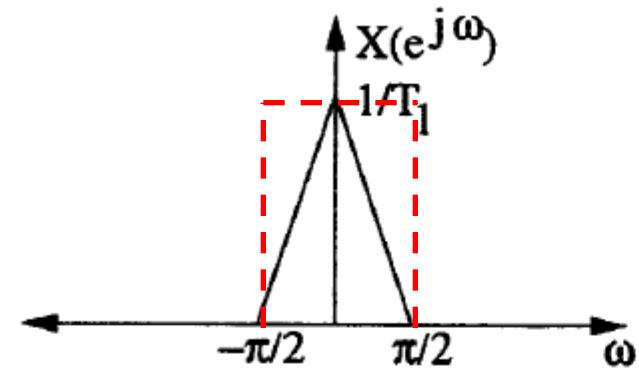
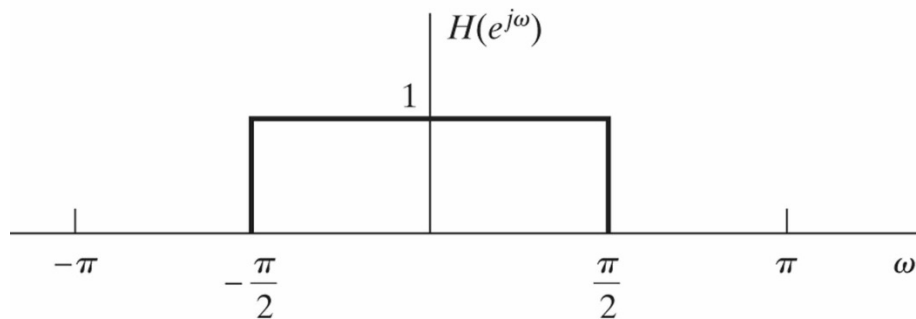
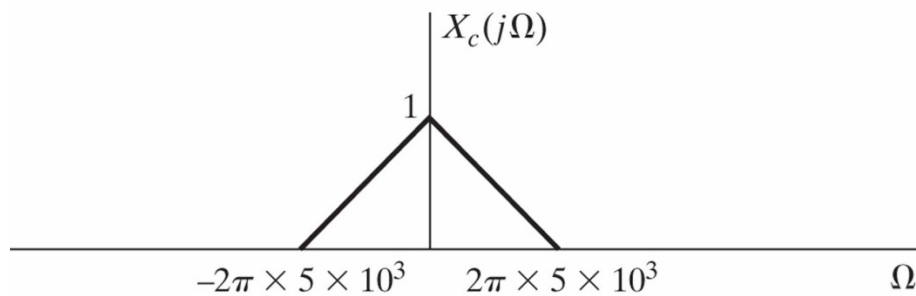
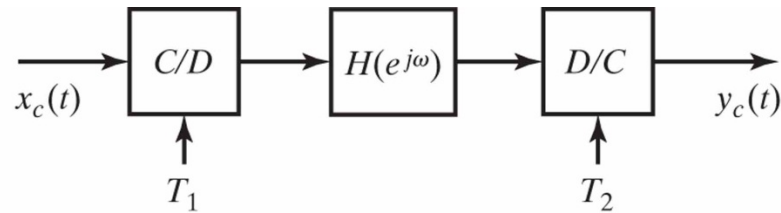
Problem 3 Solution

a) $\frac{1}{T_1} = \frac{1}{T_2} = 10^4$



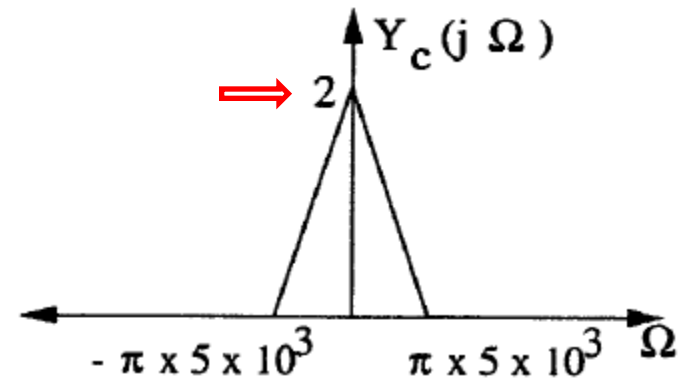
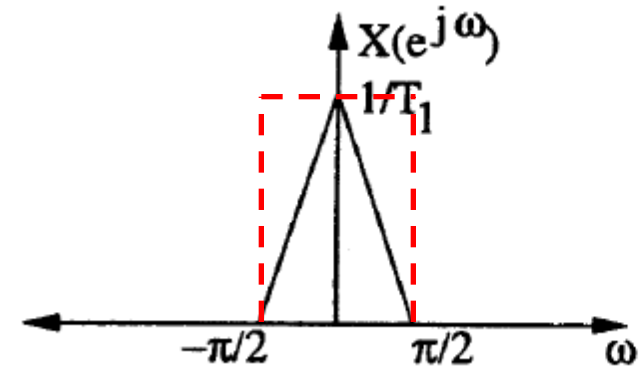
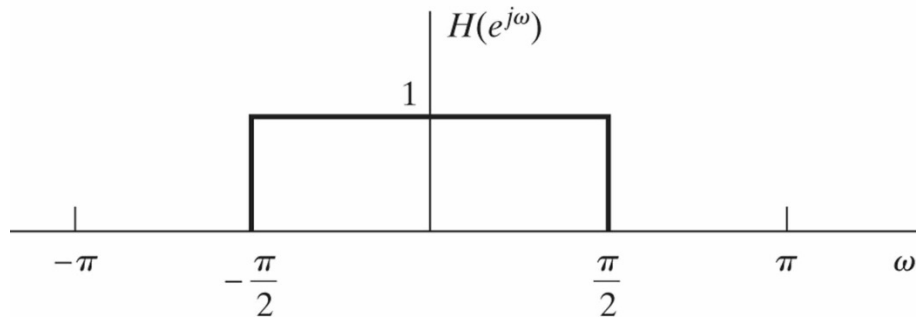
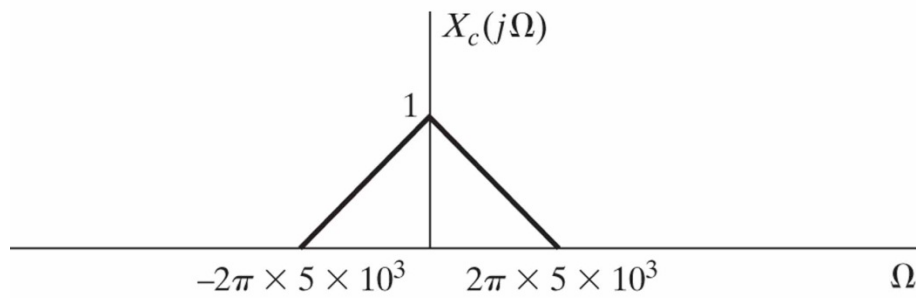
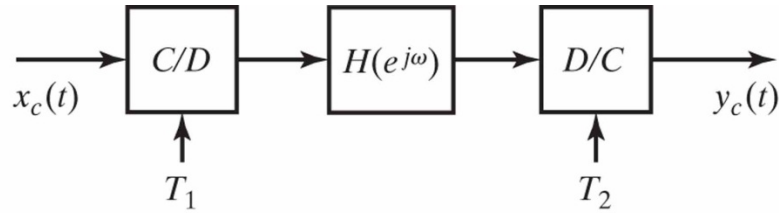
Problem 3 Solution (cont'd)

b) $\frac{1}{T_1} = \frac{1}{T_2} = 2 \times 10^4$



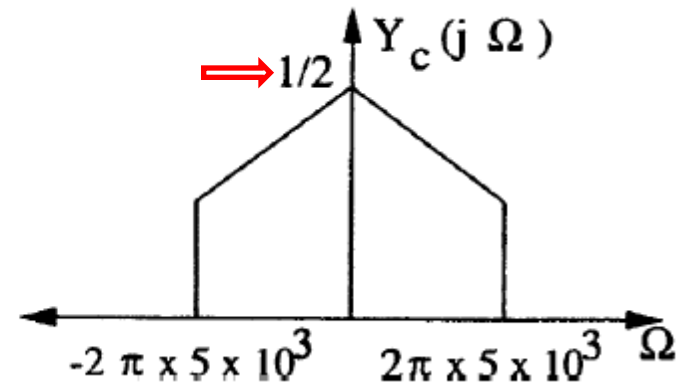
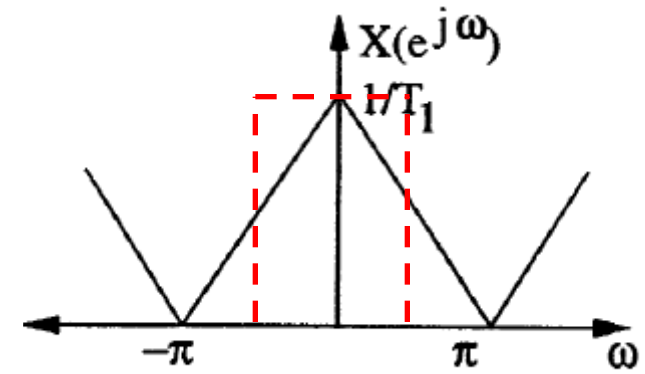
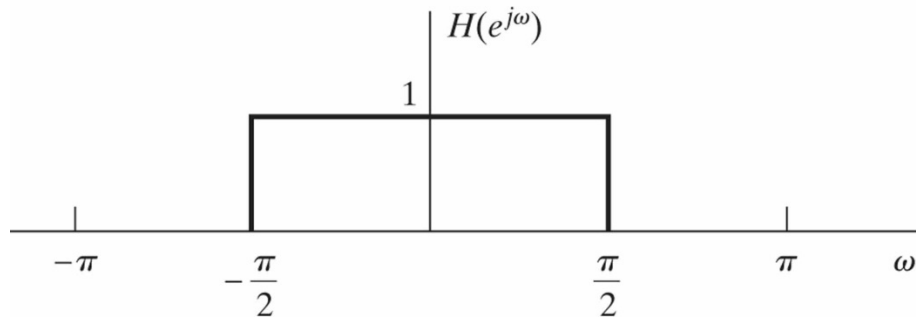
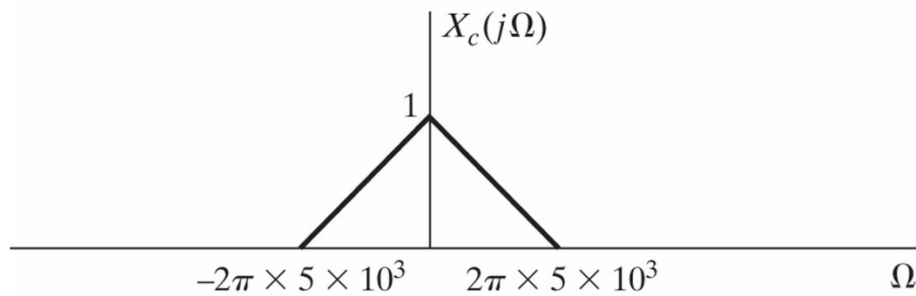
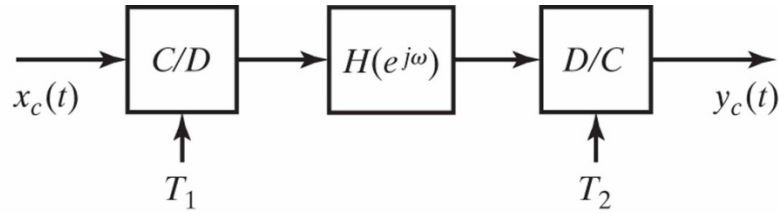
Problem 3 Solution (cont'd)

c) $\frac{1}{T_1} = 2 \times 10^4$ $\frac{1}{T_2} = 10^4$



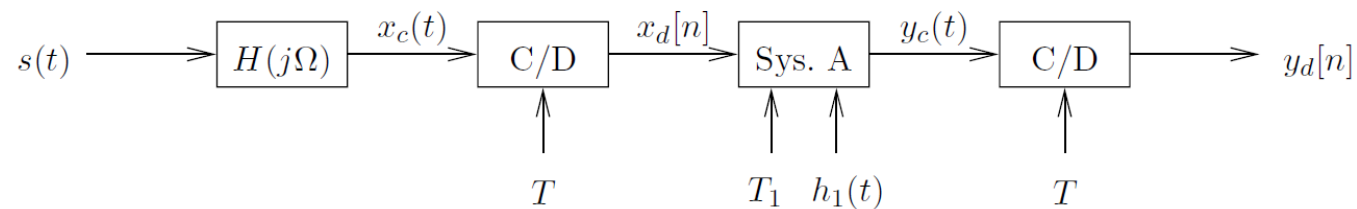
Problem 3 Solution (cont'd)

d) $\frac{1}{T_1} = 10^4$ $\frac{1}{T_2} = 2 \times 10^4$



Problem 4

- Consider the following system



$$H(j\Omega): \quad H(j\Omega) = \begin{cases} 1, & |\Omega| < \pi \cdot 10^{-3} \text{ rad/sec} \\ 0, & |\Omega| > \pi \cdot 10^{-3} \text{ rad/sec} \end{cases}$$

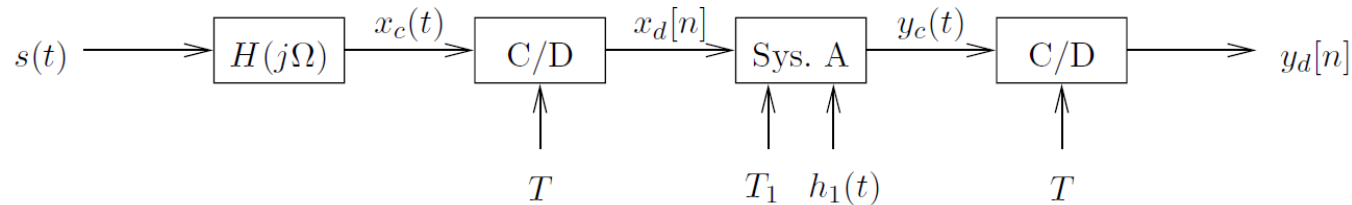
$$\text{First C/D: } x_d[n] = x_c(nT)$$

$$\text{System A: } y_c(t) = \sum_{k=-\infty}^{\infty} x_d[k] h_1(t - kT_1)$$

$$\text{Second C/D: } y_d[n] = y_c(nT)$$

Problem 4 (cont'd)

- Consider the following system



$$H(j\Omega): \quad H(j\Omega) = \begin{cases} 1, & |\Omega| < \pi \cdot 10^{-3} \text{ rad/sec} \\ 0, & |\Omega| > \pi \cdot 10^{-3} \text{ rad/sec} \end{cases}$$

$$\text{First C/D: } x_d[n] = x_c(nT)$$

$$\text{System A: } y_c(t) = \sum_{k=-\infty}^{\infty} x_d[k] h_1(t - kT_1) \quad \leftarrow$$

$$\text{Second C/D: } y_d[n] = y_c(nT)$$

Part-1: What does System A do?

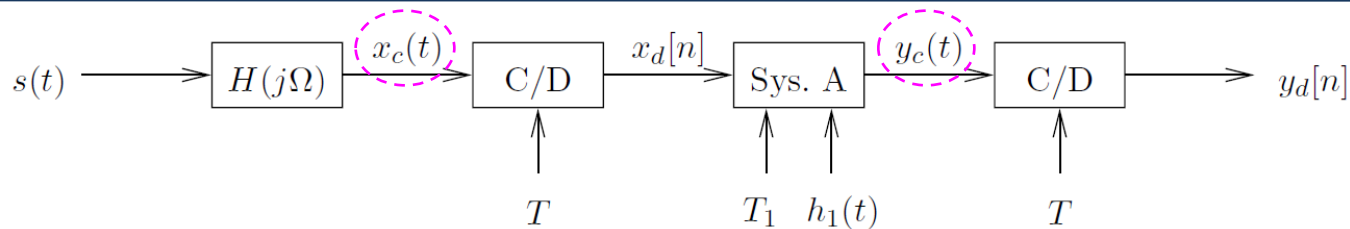
Solution: It converts the DT signal $x_d[n]$ into a CT signal $y_c(t)$ filtered by $h_1(t)$.

$$x_d[k] \quad \rightarrow \quad x_s(t) = \sum_{k=-\infty}^{\infty} x_d[k] \delta(t - kT_1) \quad \rightarrow \quad y_c(t) = x_s(t) * h_1(t)$$

(convert samples to impulses
placed at multiples of T_1)

(convolve $x_s(t)$ with C.T. filter $h_1(t)$)

Problem 4 (cont'd)



$$H(j\Omega): \quad H(j\Omega) = \begin{cases} 1, & |\Omega| < \pi \cdot 10^{-3} \text{ rad/sec} \\ 0, & |\Omega| > \pi \cdot 10^{-3} \text{ rad/sec} \end{cases}$$

$$\text{First C/D: } x_d[n] = x_c(nT)$$

$$\text{System A: } y_c(t) = \sum_{k=-\infty}^{\infty} x_d[k] h_1(t - kT_1) \quad \leftarrow$$

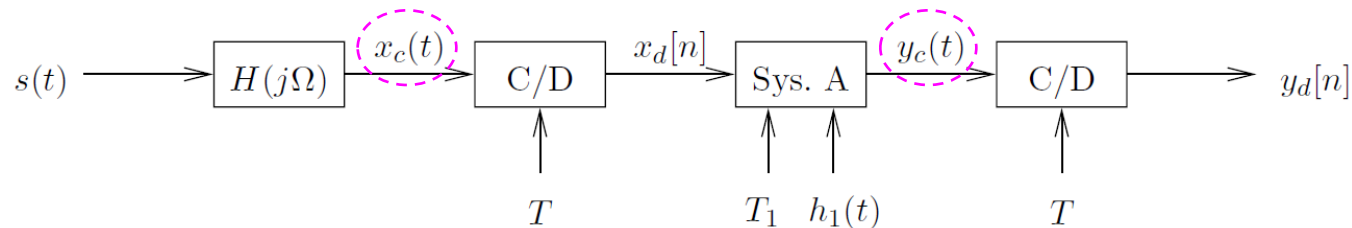
$$\text{Second C/D: } y_d[n] = y_c(nT)$$

Part-2: Specify a choice for T , T_1 and $h_1(t)$ so that $y_c(t)$ and $x_c(t)$ are guaranteed to be equal for any choice of the input signal $s(t)$.

Solution

- $x_c(t)$ is bandlimited to $\Omega_c = \pi \cdot 10^{-3} \text{ rad/s}$ because $s(t)$ was filtered by $H(j\Omega)$ with that cutoff frequency
- To guarantee equality of $x_c(t)$ and $y_c(t)$, T is chosen to be sufficiently small to avoid aliasing from the first C/D converter, and System A can be an ideal D/C converter with the same sampling period.
- We have no aliasing when $\Omega_c T < \pi \Rightarrow T < 1000s$
- System A is an ideal D/C converter when $h_1(t)$ is an appropriate sinc function $h_1(t) = \frac{\sin(\pi t / T)}{\pi t / T}$
- $T_1 = T$

Problem 4 (cont'd)



$$H(j\Omega): H(j\Omega) = \begin{cases} 1, & |\Omega| < \pi \cdot 10^{-3} \text{ rad/sec} \\ 0, & |\Omega| > \pi \cdot 10^{-3} \text{ rad/sec} \end{cases}$$

$$\text{System A: } y_c(t) = \sum_{k=-\infty}^{\infty} x_d[k] h_1(t - kT_1) \quad \leftarrow$$

$$\text{First C/D: } x_d[n] = x_c(nT)$$

$$\text{Second C/D: } y_d[n] = y_c(nT)$$

Part-3: Is the choice in Part-2 unique for T , T_1 and $h_1(t)$ for $y_c(t)$ and $x_c(t)$ to be equal for any choice of $s(t)$, or are there other choices?

Solution

- Any $T < 1000\text{s}$ works
- T_1 must be equal to T
- What about the choice of $h_1(t)$? If we choose $T < 1000\text{s}$, then:

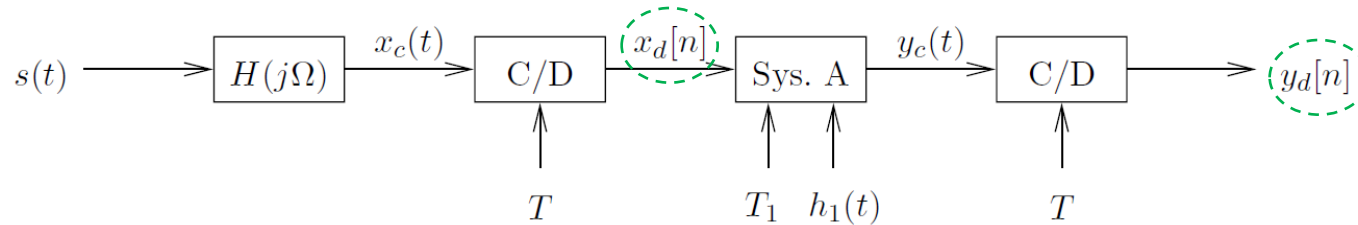
$$X_d(e^{j\omega}), \text{ DTFT of } x_d[n], \text{ is zero for } \frac{\pi T}{1000} < |\omega| < \pi$$

$$\Rightarrow X_s(j\Omega), \text{ CTFT of } x_s(t), \text{ is zero for } \frac{\pi T}{1000T_1} = \frac{\pi}{1000} < |\Omega| < \frac{\pi}{T_1}$$

$$H_1(j\Omega), \text{ CTFT of } h_1(t), \text{ can be anything in that frequency range}$$

$$H_1(j\Omega) = \begin{cases} T & |\Omega| < \pi / 1000 \\ \text{anything} & \pi / 1000 < |\Omega| < \pi / T \\ 0 & |\Omega| > \pi / T \end{cases}$$

Problem 4 (cont'd)



$$H(j\Omega): \quad H(j\Omega) = \begin{cases} 1, & |\Omega| < \pi \cdot 10^{-3} \text{ rad/sec} \\ 0, & |\Omega| > \pi \cdot 10^{-3} \text{ rad/sec} \end{cases}$$

$$\text{System A: } y_c(t) = \sum_{k=-\infty}^{\infty} x_d[k] h_1(t - kT_1) \quad \leftarrow$$

$$\text{First C/D: } x_d[n] = x_c(nT)$$

$$\text{Second C/D: } y_d[n] = y_c(nT)$$

Part-4: Determine the most general conditions you can on T , T_1 and $h_1(t)$ so that $y_d[n] = x_d[n]$ (condition called **consistent resampling**). Consider what happens $x_d[n] = \delta[n - n_0]$ for an integer n_0 .

Solution: System A constructs a CT signal $y_c(t)$ from $x_d[n]$, which is then resampled to obtain $y_d[n]$. The resampling is referred to as consistent if $y_d[n] = x_d[n]$.

- System A takes each sample of $x_d[n]$ and replaces it with $h_1(t)$ delayed by nT_1 and scaled by $x_d[n]$ at that point.
- The C/D converter resamples the result

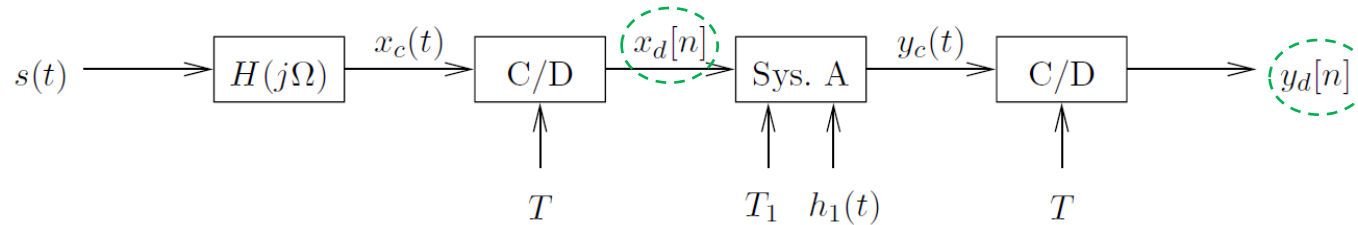
$$x_d[n] = \delta[n - n_0] \Rightarrow y_c(t) = h_1(t - n_0T_1)$$

$$y_d[n] = y_c(nT) = h_1(nT - n_0T_1)$$

$$\text{consistent resampling} \Rightarrow h_1(nT - n_0T_1) = \delta[n - n_0]$$

- Because of linearity between $x_d[n]$ and $y_d[n]$, this is the only condition that must be checked

Problem 4 (cont'd)



$$H(j\Omega): \quad H(j\Omega) = \begin{cases} 1, & |\Omega| < \pi \cdot 10^{-3} \text{ rad/sec} \\ 0, & |\Omega| > \pi \cdot 10^{-3} \text{ rad/sec} \end{cases}$$

$$\text{First C/D: } x_d[n] = x_c(nT)$$

$$\begin{aligned} \text{System A: } y_c(t) &= \sum_{k=-\infty}^{\infty} x_d[k] h_1(t - kT_1) \\ \text{Second C/D: } y_d[n] &= y_c(nT) \end{aligned} \quad \leftarrow$$

$$h_1(nT - n_0T_1) = \delta[n - n_0]$$

$$\Rightarrow \begin{aligned} \text{evaluating at } n = n_0: \quad & 1 = h_1(nT - nT_1) = h_1(n(T - T_1)) \\ \text{evaluating at } n \neq n_0: \quad & 0 = h_1(nT - n_0T_1) \end{aligned}$$

The case of practical significance is when $T = T_1$. Then

$$\Rightarrow \begin{aligned} \text{evaluating at } n = n_0: \quad & 1 = h_1(0) \\ \text{evaluating at } n \neq n_0: \quad & 0 = h_1(t), t = mT, m \neq 0 \end{aligned}$$

Therefore, $h_1(t)$ should satisfy an interpolating condition: $h_1(0) = 1$ and $h_1(t) = 0$ for all multiples of T .

It doesn't matter what $h_1(t)$ is for other values of t .