

**Problem Set: Frequency Analysis of LTI Systems**

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**Problem 1:** A simple causal LTI system characterized by the following unit sample response:

$$\begin{aligned} h[0] &= 2 \\ h[1] &= 3 \\ h[2] &= 2 \\ h[n] &= 0, \quad \forall n > 2 \end{aligned}$$

- What is the frequency response,  $H(e^{j\omega})$ , of this LTI system?
  - What is the magnitude of  $H(e^{j\omega})$  at  $\omega = 0, \pi/2, \pi$ ?
  - If this LTI system is used as a filter, what is the set of frequencies that are removed?
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**Problem 2:** Let  $x[n]$  be a causal,  $N$ -point sequence that is zero outside the range  $0 \leq n \leq N-1$ . When  $x[n]$  is the input to the causal LTI system represented by the difference equation:

$$y[n] - \frac{1}{4}y[n-2] = x[n-2] - \frac{1}{4}x[n]$$

the output is  $y[n]$ , also a causal,  $N$ -point sequence.

- Show that the causal LTI system described by this difference equation represents an all-pass filter.
- Given that

$$\sum_{n=0}^{N-1} |x[n]|^2 = 5$$

determine the value of

$$\sum_{n=0}^{N-1} |y[n]|^2$$


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**Problem 3:** Suppose a causal linear time invariant (LTI) system with frequency response  $H(e^{j\omega})$  is described by the following difference equation relating input  $x[n]$  to output  $y[n]$ :

$$y[n] = x[n] + \alpha x[n-1] + \beta x[n-2] + \gamma x[n-3]$$

Here,  $\alpha$ ,  $\beta$ , and  $\gamma$  are constants independent of  $\omega$ .

- Determine the values of  $\alpha$ ,  $\beta$  and  $\gamma$  so that the frequency response of system  $H$  is  $H(e^{j\omega}) = 1 - 0.5e^{-j2\omega} \cos \omega$ .
- Suppose that  $y[\cdot]$ , the output of the LTI system with frequency response  $H(e^{j\omega})$ , is used as the input to a second causal LTI system with frequency response  $G(e^{j\omega})$ , producing  $W(e^{j\omega})$ . If  $H(e^{j\omega}) = 1 - 0.5e^{-j2\omega} \cos \omega$ , what should the frequency response,  $G(e^{j\omega})$ , be so that  $w[n] = x[n]$  for all  $n$ ? Here  $W(e^{j\omega}) = \text{DTFT}\{w[n]\}$ .
- How would you choose the coefficients  $\alpha$ ,  $\beta$ , and  $\gamma$  so that  $H(e^{j\omega})$  has linear phase?

**Problem 4:** An LTI system with impulse response  $h_1[n]$  is an ideal lowpass filter with cutoff frequency  $\omega_c = \pi/2$ . The frequency response of the system is  $H_1(e^{j\omega})$ . Suppose a new LTI system with impulse response  $h_2[n]$  is obtained from  $h_1[n]$  by  $h_2[n] = (-1)^n h_1[n]$ . Sketch the frequency response  $H_2(e^{j\omega})$ .

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**Problem 5:** A sequence  $x[n]$  is the output of a linear time-invariant system whose input is  $s[n]$ . This system is described by the difference equation

$$x[n] = s[n] - e^{-8\alpha} s[n-8], \quad (1.1)$$

where  $\alpha > 0$ .

- a) Find the system function

$$H(z) = \frac{X(z)}{S(z)}$$

and plot its poles and zeros in the z-plane. Indicate the region of convergence.

- b) We wish to recover  $s[n]$  from  $x[n]$  with a linear time-invariant system. Find the system function

$$H_2(z) = \frac{Y(z)}{X(z)}$$

such that  $y[n] = s[n]$ . Find all possible regions of convergence for  $H_2(z)$ , and for each, tell whether or not the system is causal and/or stable.

- c) Find all possible choices for the impulse response  $h_2[n]$  such that

$$y[n] = h_2[n] * x[n] = s[n]$$

- d) For all choices determined in Part (c), demonstrate, by explicitly evaluating the convolution in Eq.(1.1), that when  $s[n] = \delta[n]$ , then  $y[n] = \delta[n]$ .
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**Problem 6:** Consider an LTI system with input  $x[n]$  and output  $y[n]$ . When the input to the system is

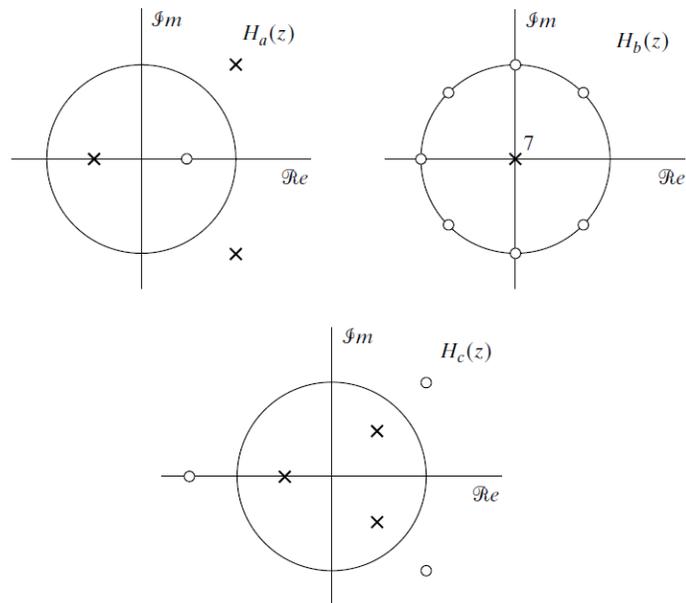
$$x[n] = 5 \frac{\sin(0.4\pi n)}{\pi n} + 10 \cos(0.5\pi n)$$

the corresponding output is

$$y[n] = 10 \frac{\sin(0.3\pi(n-10))}{\pi(n-10)}$$

Determine the frequency response  $H(e^{j\omega})$  and the impulse response  $h[n]$  for the LTI system.

**Problem 7:** Figure 1 shows the pole-zero plots for three different causal LTI systems with real impulse responses. Indicate which of the following properties apply to each of the systems pictured: stable, IIR, FIR, minimum phase, all-pass, linear phase, positive group delay at all  $\omega$  :



**Figure 1: Pole-Zero plots**